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TPL9201
MICROCONTROLLER POWER SUPPLY
AND MULTIPLE LOW-SIDE DRIVER
SLIS123A−JUNE 2006−REVISED JUNE 2006

FEATURES

- **Eight Low-Side Drivers With Internal Clamp for Inductive Loads and Current Limiting for Self Protection**
  - **Seven Outputs Rated at 150 mA and Controlled Through Serial Interface**
  - **One Output Rated at 150 mA and Controlled Through Serial Interface and Dedicated Enable Pin**
- **5-V ±5% Regulated Power Supply With 200-mA Load Capability at V_{IN} Max of 18 V**
- **Internal Voltage Supervisory for Regulated Output**
- **Serial Communications for Control of Eight Low-Side Drivers**
- **Enable/Disable Input for OUT1**
- **5-V or 3.3-V I/O Tolerant for Interface to Microcontroller**
- **Programmable Power-On Reset Delay Before $\overline{RST}$ Asserted High, Once 5 V Is Within Specified Range (6 ms Typ)**
- **Programmable Deglitch Timer Before $\overline{RST}$ Is Asserted Low (40 μs Typ)**
- **Zero-Voltage Detection Signal With Built-In Filter of 20 μs**
- **Thermal Shutdown for Self Protection**

APPLICATIONS

- **Electrical Appliances**
  - **Air Conditioning Units**
  - **Ranges**
  - **Dishwashers**
  - **Refrigerators**
  - **Microwaves**
  - **Washing Machines**
- **General-Purpose Interface Circuits, Allowing Microcontroller Interface to Relays, Electric Motors, LEDs, and Buzzers**

N OR PWP PACKAGE
(TOP VIEW)

ZVS	1	20	SYN
OUT1	2	19	V_{IN}
OUT2	3	18	$5V_{OUT}$
OUT3	4	17	SCLK
OUT4	5	16	NCS
OUT5	6	15	MOSI
OUT6	7	14	$\overline{RST}$
OUT7	8	13	R_{DELAY}
OUT8	9	12	EN1
GND	10	11	GND
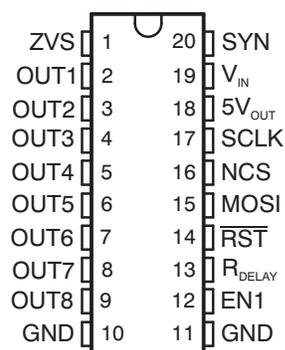
DESCRIPTION/ORDERING INFORMATION

The power supply provides regulated 5-V output to power the system microcontroller and drive eight low-side switches. The AC zero-detect circuitry is monitoring the crossover voltage of the mains AC supply. The resultant signal is a low-frequency clock output on the ZVS terminal, based on the AC-line cycle. This information allows the microcontroller to reduce in-rush current by powering loads on the AC-line peak voltage.

A serial communications interface controls the eight low-side outputs; each output has an internal snubber circuit to absorb the energy in the inductor at turn OFF. Alternatively, the system can use a fly-back diode to V_{IN} to help recirculate the energy in an inductive load at turn OFF.

ORDERING INFORMATION

T_A	PACKAGE		ORDERABLE PART NUMBER	TOP-SIDE MARKING
−40°C to 125°C	PDIP – N	Tube of 20	TPL9201N	TPL9201
	PowerPAD™ – PWP	Reel of 2000	TPL9201PWPR	IC9201
		Tube of 70	TPL9201PWP	

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PINOUT CONFIGURATION

NO.	NAME	I/O	DESCRIPTION
1	ZVS	O	Zero-voltage synchronization
2	OUT1	O	Low-side output 1
3	OUT2	O	Low-side output 2
4	OUT3	O	Low-side output 3
5	OUT4	O	Low-side output 4
6	OUT5	O	Low-side output 5
7	OUT6	O	Low-side output 6
8	OUT7	O	Low-side output 7
9	OUT8	O	Low-side output 8
10[1]	GND	I	Ground
11[1]	GND	I	Ground
12	EN1	I	Enable/disable for OUT1
13	R_{DELAY}	O	Power-up reset delay
14[2]	$\overline{RST}$	I/O	Power-on reset output (open drain, active low)
15	MOSI	I	Serial data input
16	NCS	I	Chip select
17	SCLK	I	Serial clock for data synchronization
18	$5V_{OUT}$	O	Regulated output
19	V_{IN}	I	Unregulated input voltage source
20	SYN	I	AC zero detect input

(1) Terminals 10 and 11 are fused internally in the lead frame for the 20-pin PDIP package.
(2) Terminal 14 can be used as an input or an output.

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FUNCTIONAL BLOCK DIAGRAM

Enable — EN1 — Enables OUT1 — OUT1 — OUT1 at 150 mA — OUT1

100 kΩ

6 V

100 kΩ

Gate Control for Outputs 1 Through 8

NCS — NCS

Parallel Register

SCLK — SCLK

Serial Register

100 kΩ

MOSI — MOSI

20 Ω (2 W)

10 V — V$_{IN}$

7–18 V

100 kΩ

OUT2 at 150 mA — OUT2
OUT3 at 150 mA — OUT3
OUT4 at 150 mA — OUT4
OUT5 at 150 mA — OUT5
OUT6 at 150 mA — OUT6
OUT7 at 150 mA — OUT7
OUT8 at 150 mA — OUT8

Optional, dependent on heat-management implementation (see Note A)

Bandgap Ref

Gate Drive and Control

PMOS

Comp

V$_{ref}$

5V$_{OUT}$ — 5 V

5 kΩ

GND — GND

I$_{const}$

V$_{ref}$ — Voltage Supervisor

$\overline{RST}$ — $\overline{RST}$

6 V

20-µs Filter

S Q
R

500 kΩ

R$_{DELAY}$ — R$_{DELAY}$

ZVS — ZVS

SYN — SYN

10 kΩ

25 kΩ

100 kΩ

A. The resistor and Zener diode are required if there is insufficient thermal-management allocation.

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DETAILED DESCRIPTION

The 5-V regulator is powered from V_{IN}, and the regulated output is within 5 V ±5% over the operating conditions. The open-drain power-on reset ($\overline{RST}$) pin remains low until the regulator exceeds the set threshold, and the timer value set by the capacitor on the reset delay (R_{DELAY}) pin expires. If both of these conditions are satisfied, $\overline{RST}$ is asserted high. This signifies to the microcontroller that serial communications can be initiated to the TPL9201.

The serial communications is an 8-bit format, with data transfer synchronized using a serial clock from the microcontroller. A single register controls all the outputs (one bit per output). The default value is zero (OFF). If an output requires pulse width modulation (PWM) function, the register must be updated at a rate faster than the desired PWM frequency. OUT1 can be controlled by serial input from the microcontroller or with the dedicated enable (EN1) pin. If EN1 is pulled low or left open, the serial input through the shift register controls OUT1. If EN1 is pulled high, OUT1 always is turned on, and the serial input for OUT1 is ignored.

The SYN input translates the image of the mains voltage through the secondary of the transformer. The SYN input has a resistor to protect from high currents into the IC. The zero-voltage synchronization output translates the AC-line cycle frequency into a low-frequency clock, which can be used for a timing reference and to help power loads on the AC-line peak voltage (to reduce in-rush currents).

If $\overline{RST}$ is asserted, all outputs are turned OFF internally, and the input register is reset to all zeroes. The microcontroller must write to the register to turn the outputs ON again.

Absolute Maximum Ratings[1]

			MIN	MAX	UNIT
$V_{I(unreg)}$	Unregulated input voltage[2][3]	V_{IN}		24	V
		SYN		24	
$V_{I(logic)}$	Logic input voltage[2][3]	EN1, MOSI, SCLK, and NCS		7	V
		$\overline{RST}$ and R_{DELAY}		7	
V_O	Low-side output voltage	OUT1−OUT8		16.5	V
I_{LIMIT}	Output current limit[4]	OUTn = ON and shorted to V_{IN} with low impedance		350	mA
θ_{JA}	Thermal impedance, junction to ambient[5]	N package		69	°C/W
		PWP package		33	
θ_{JC}	Thermal impedance, junction to case[5]	N package		54	°C/W
		PWP package		20	
θ_{JP}	Thermal impedance, junction to thermal pad[5]	PWP package		1.4	°C/W
P_D	Continuous power dissipation[6]	N package		1.8	W
		PWP package		3.7	
ESD	Electrostatic discharge[7]			2	kV
T_A	Operating ambient temperature range		−40	125	°C
T_{stg}	Storage temperature range		−65	125	°C
T_{lead}	Lead temperature	Soldering, 10 s		260	°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values are with respect to GND.
(3) Absolute negative voltage on these pins must not go below −0.5 V.
(4) Not more than one output should be shorted at a time, and duration of the short circuit should not exceed 1 ms.
(5) The thermal data is based on using 1-oz copper trace with JEDEC 51-5 test board for PWP and JEDEC 51-7 test board for N.
(6) The data is based on ambient temperature of 25°C max.
(7) The Human-Body Model is a 100-pF capacitor discharged through a 1.5-kΩ resistor into each pin.

Dissipation Ratings

PACKAGE	$T_A \leq 25°C$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25°C$	$T_A = 125°C$ POWER RATING
N	1812 mW	14.5 mW/°C	362 mW
PWP	3787 mW	30.3 mW/°C	757 mW

Recommended Operating Conditions

			MIN	MAX	UNIT
$V_{I(unreg)}$	Unregulated input voltage	V_{IN}	7	18	V
		SYN	0	18	
$V_{I(logic)}$	Logic input voltage	EN1, MOSI, SCLK, NCS, $\overline{RST}$, and R_{DELAY}	0	5.25	V
T_A	Operating ambient temperature		−40	125	°C

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Electrical Characteristics

T_A = –40°C to 125°C, V_{IN} = 7 V to 18 V (unless otherwise stated)

PARAMETER		TEST CONDITIONS	MIN	TYP[1]	MAX	UNIT
Supply Voltage and Current						
V_{IN}[2]	Input voltage		7		18	V
I_{VIN}	Input supply current	Enable = ON, OUT1–OUT8 = OFF			3	mA
		Enable = ON, OUT1–OUT8 = ON			5	
Logic Inputs (MOSI, NCS, SCLK, and EN1)						
V_{IL}	Logic input low level	I_{IL} = 100 µA			0.8	V
V_{IH}	Logic input high level	I_{IL} = 100 µA	2.4			
Reset ($\overline{RST}$)						
V_{OL}	Logic level output	I_{OL} = 1.6 mA			0.4	V
V_{OH}[3]	Logic level output	5-kΩ pullup to V_{CC}	V_{CC} – 0.8			V
V_H	Disabling reset threshold	5-V regulator ramps up		4.25	4.5	V
V_L	Enabling reset threshold	5-V regulator ramps down	3.3	3.75		V
V_{HYS}	Threshold hysteresis		0.12	0.5		V
Reset Delay (R_{DELAY})						
I_{OUT}	Output current		18	28	48	µA
T_{DW}	Reset delay timer	C = 47 nF	3	6		ms
T_{UP}	Reset capacitor to low level	C = 47 nF		45		µs
Output (OUT1–OUT8)						
V_{OL}	Output ON	I_{OUTn} = 150 mA		0.4	0.7	V
I_{OH}	Output leakage	V_{OH} = Max of 16.5 V			2	µA
Regulator Output ($5V_{OUT}$)						
$5V_{OUT}$	Output supply	I_{5VOUT} = 5 mA to 200 mA, V_{IN} = 7 V to 18 V, C_{5VOUT} = 1 µF	4.75	5	5.25	V
I_{5VOUT} limit	Output short-circuit current	$5V_{OUT}$ = 0 V	200			mA
Thermal Shutdown						
T_{SD}	Thermal shutdown			150		°C
T_{HYS}	Hysteresis			20		°C
Zero Voltage Synchronization (ZVS)						
V_{SYNTH}	Transition threshold		0.4	0.75	1.1	V
I_{SYN}	Input activating current	R_{ZV} = 10 kΩ, V_{SYN} = 24 V			2	mA
t_D	Transition filtering time	Rising and falling	10	30	70	µs

(1) All typical values are at T_A = 25°C.
(2) There are external high-frequency noise-suppression capacitors and filter capacitors on V_{IN}.
(3) V_{CC} is the pullup resistor voltage.

Output Control Register

MSB   LSB

IN8	IN7	IN6	IN5	IN4	IN3	IN2	IN1
0	0	0	0	0	0	0	0

INn = 0: Output OFF
INn = 1: Output ON

To operate the output in PWM mode, the output control register must be updated at a rate twice the desired PWM frequency of the output. Maximum PWM frequency is 5 kHz. The register is updated every 100 µs.

ENABLE TRUTH TABLE

EN1	SERIAL INPUT FOR OUT1	OUT1
Open	H	On
Open	L	Off
L	H	On
L	L	Off
H	H	On
H	L	On

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Serial Communications Interface

The serial communications is an 8-bit format, with data transfer synchronized using a serial clock from the microcontroller (see Figure 1). A single register controls all the outputs. The signal gives the instruction to control the output of TPL9201.

The NCS signal enables the SCLK and MOSI data when it is low. After NCS is set low for T_1, synchronization clock and data begin to transmit and, after the 8-bit data has been transmitted, NCS is set high again to disable SCLK and MOSI and transfer the serial data to the control register. SCLK must be held low when NCS is in the high state.
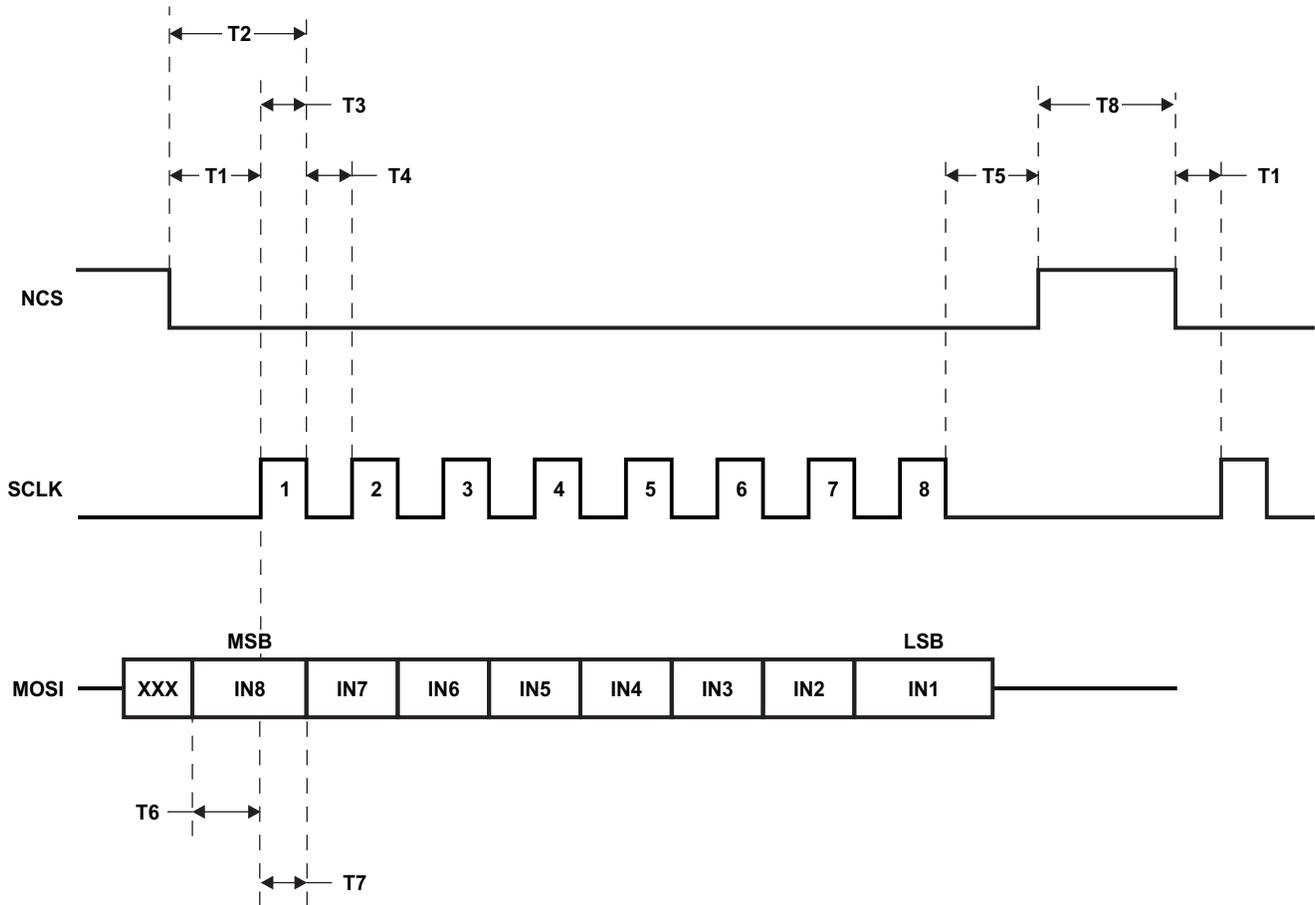
Figure 1. Serial Communications

Timing Requirements

T_A = −40°C to 125°C, V_{IN} = 7 V to 18 V (unless otherwise stated)

	PARAMETER	MIN	TYP	MAX	UNIT
f_{SPI}	SPI frequency			4	kHz
T1	Delay time, NCS falling edge to SCLK rising edge	10			ns
T2	Delay time, NCS falling edge to SCLK falling edge	80			ns
T3	Pulse duration, SCLK high	60			ns
T4	Pulse duration, SCLK low	60			ns
T5	Delay time, last SCLK falling edge to NCS rising edge	80			ns
T6	Setup time, MOSI valid before SCLK edge	10			ns
T7	Hold time, MOSI valid after SCLK edge	10			ns
T8	Time between two words for transmitting	170			ns

Reset Delay (R$_{DELAY}$)

The R_{DELAY} output provides a constant current source to charge an external capacitor to approximately 6.5 V. The external capacitor is selected to provide a delay time, based on the current equation for a capacitor, I = C(Δv/Δt) and a 28-μA typical output current.

Therefore, the user should select a 47-nF capacitor to provide a 6-ms delay at 3.55 V.

I = C(Δv/Δt)
28 μA = C × (3.55 V/6 ms)
C = 47 nF

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APPLICATION INFORMATION
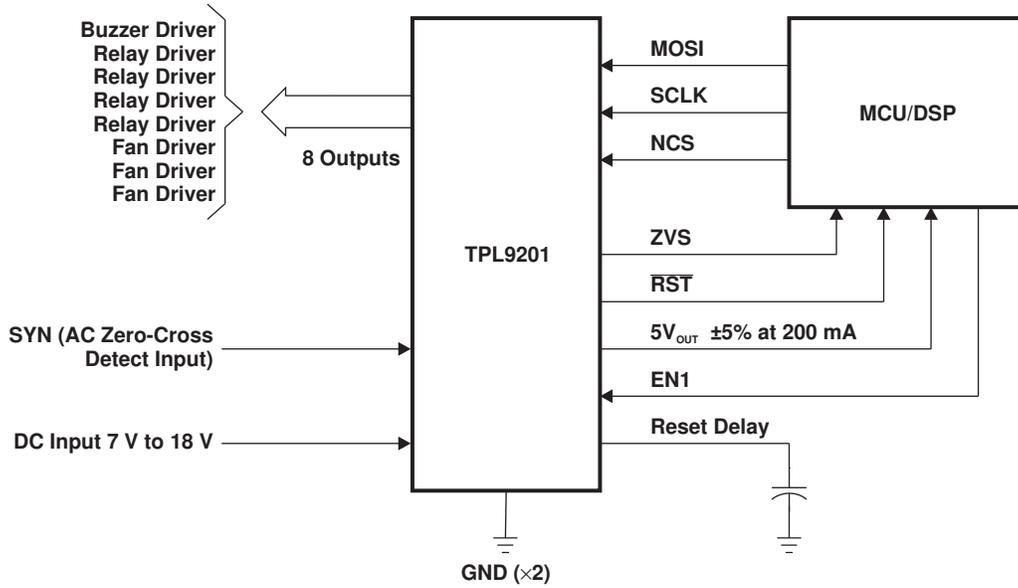
Figure 2. Typical Application
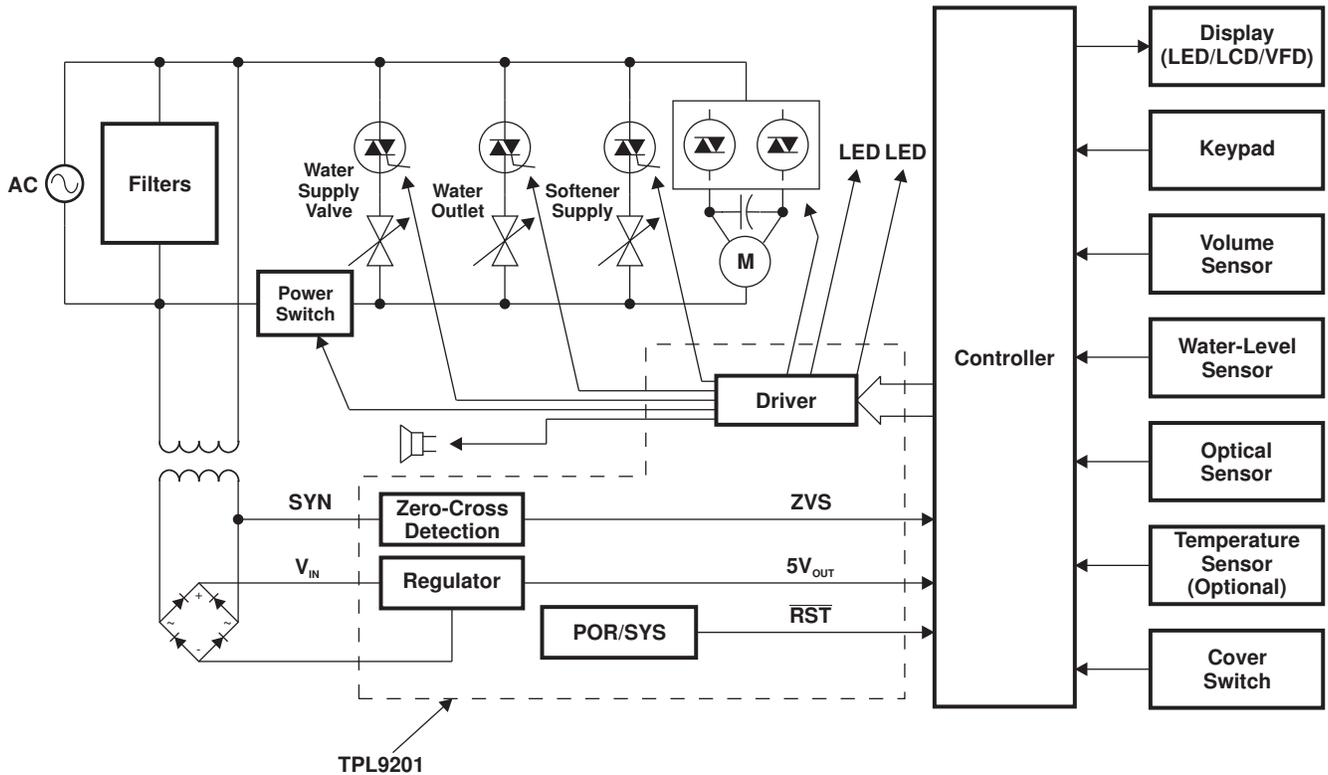
Figure 3. Washing-Machine Application

APPLICATION INFORMATION (continued)

PCB Layout

To maximize the efficiency of this package for application on a single-layer or multi-layer PCB, certain guidelines must be followed when laying out this part on the PCB.

The following information is to be used as a guideline only.

For further information, see the PowerPAD concept implementation document.

Application Using a Multilayer PCB

In a multilayer board application, the thermal vias are the primary method of heat transfer from the package thermal pad to the internal ground plane (see Figure 4 and Figure 5).

The efficiency of this method depends on several factors: die area, number of thermal vias, thickness of copper, etc. (see the *PowerPAD™ Thermally Enhanced Package Technical Brief*, literature number SLMA002).
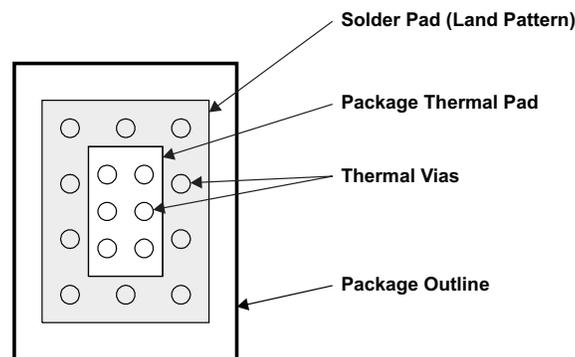
Figure 4. Package and PCB Land Configuration for a Multilayer PCB
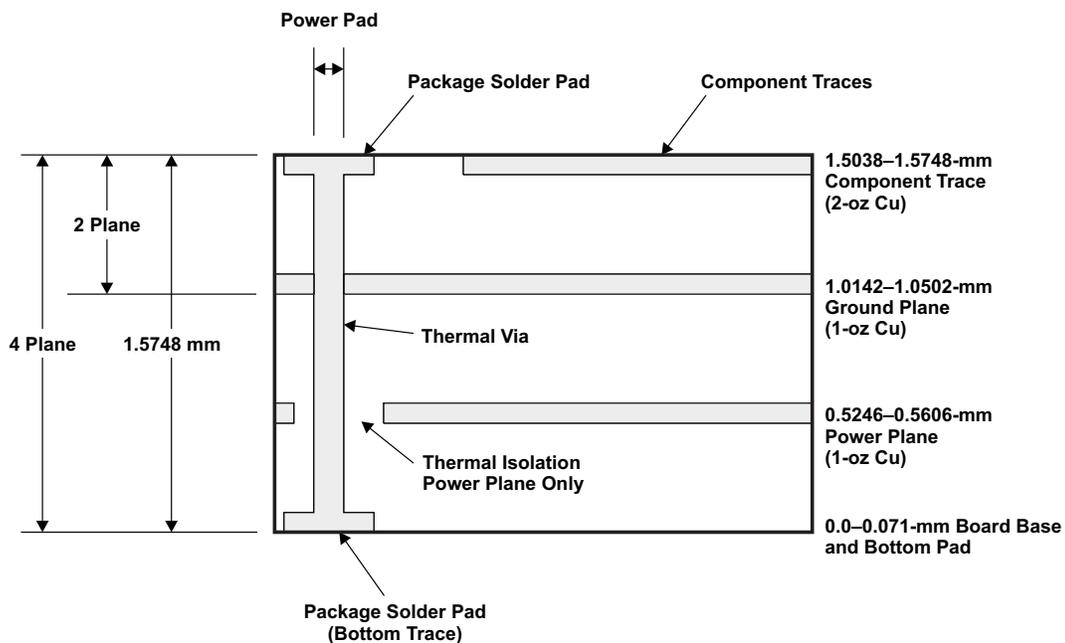
Figure 5. Multilayer Board (Side View)

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APPLICATION INFORMATION (continued)

Application Using a Single-Layer PCB

In a single-layer board application, the thermal pad is attached to a heat spreader (copper area) by using the low thermal-impedance attachment method (solder paste or thermal-conductive epoxy). With either method, it is advisable to use as much copper trace area as possible to dissipate the heat.

CAUTION:

If the attachment method is not implemented correctly, the functionality of the product cannot be assured. Power-dissipation capability is adversely affected if the device is incorrectly mounted onto the circuit board.

Use as Much Copper Area as Possible for Heat Spread

Package Thermal Pad

Package Outline
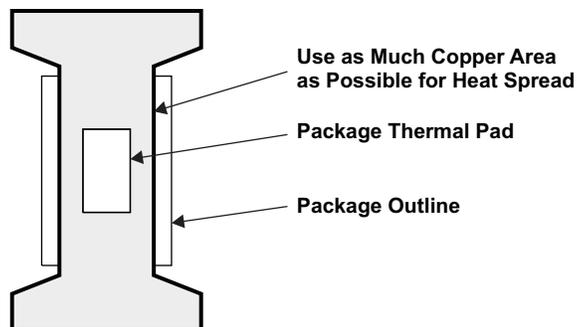
Figure 6. Layout Recommendations for a Single-Layer PCB

APPLICATION INFORMATION (continued)

Recommended Board Layout
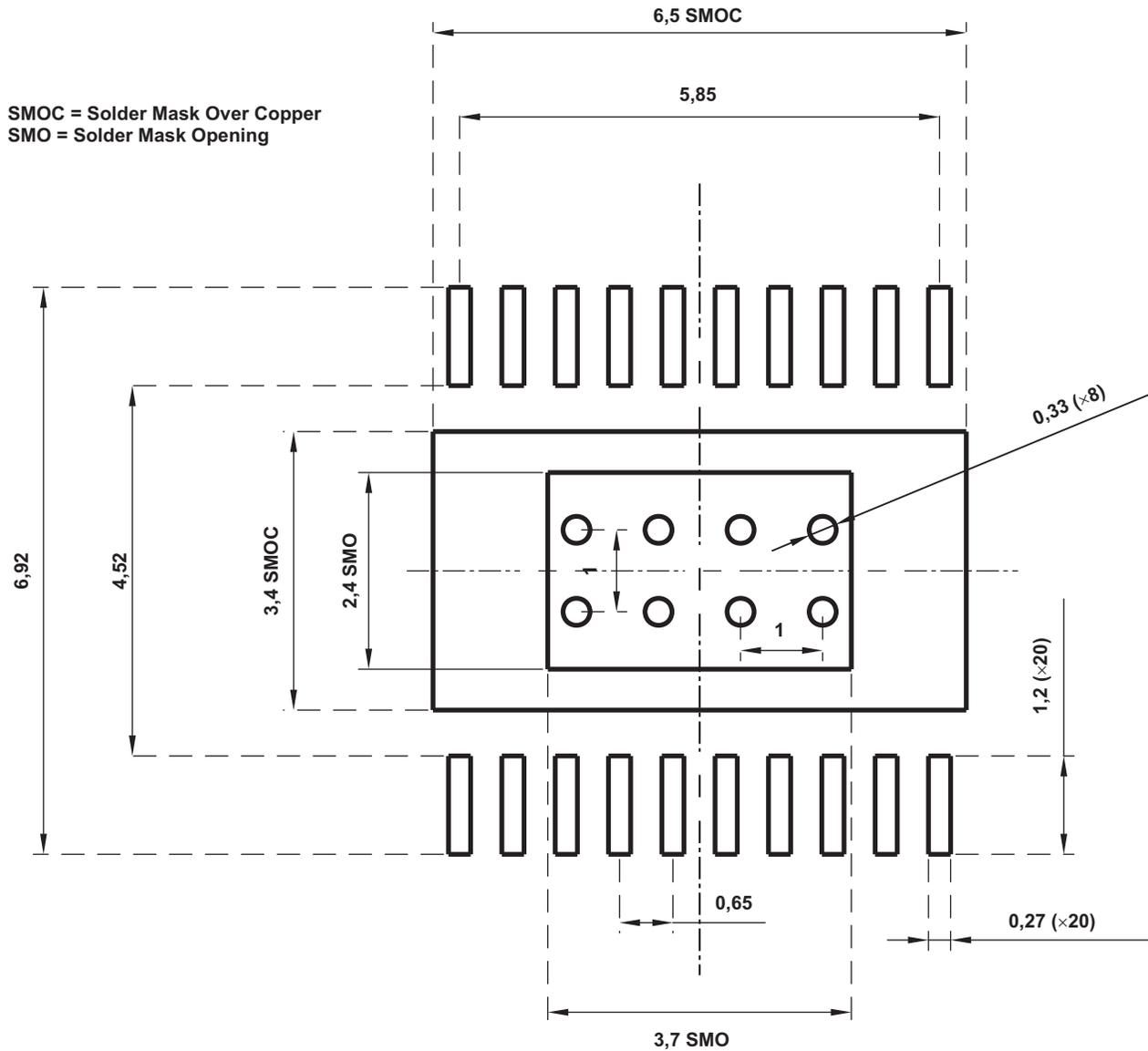
Figure 7. Recommended Board Layout for PWP

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Mailing Address: Texas Instruments
                     Post Office Box 655303 Dallas, Texas 75265