# ARM966E-S<sup>™</sup> 16/32-Bit Flash MCU with Ethernet, USB, CAN, AC motor control, 4 timers, ADC, RTC, DMA

#### ■ 16/32-bit 96 MHz ARM9E based MCU

- ARM966E-S RISC core: Harvard architecture, 5-stage pipeline, Tightly-Coupled Memories (SRAM and Flash)
- STR91xF implementation of core adds highspeed burst Flash memory interface, instruction prefetch queue, branch cache
- Up to 96 MIPS directly from Flash memory
- Single-cycle DSP instructions are supported
- Binary compatible with 16/32-bit ARM7 code

#### ■ Dual burst Flash memories, 32-bits wide

- 256KB/512KB Main Flash, 32KB 2nd Flash
- Sequential Burst operation up to 96 MHz
- 100K min erase cycles, 20 yr min retention

#### SRAM, 32-bits wide

- 64K or 96K bytes, optional battery backup
- 9 programmable DMA channels
  - One for Ethernet, eight programmable channels
- Clock, reset, and supply management
  - Two supplies required. Core: 1.8 V +/-10%, I/O: 2.7 to 3.6 V
  - Internal oscillator operating with extgraa 4-25 MHz crystal
  - Internal PLL up to 96MHz
  - Real-time clock provides calendar functions, tamper detection, and valke-up functions
  - Reset Supervisor nonitors voltage supplies, watchdog timer, wake-up unit, ext. reset
  - Brown-c.it nonitor for early warning interrupt - Run, circ, and Sleep Mode as low as 50 uA

#### Cpore ting temperature -40 to +85°C

#### Vectored interrupt controller (VIC)

- 32 IRQ vectors, 30 intr pins, any can be FIQ
  Branch cache minimizes interrupt latency
- 8-channel, 10-bit A/D converter (ADC) - 0 to 3.6V range, 0.7 usec conversion



11 Communication interfaces

- 10/100 Ethernet MAC with LVA and MII port
- USB Full-speed (12 Mbos' slave device
- CAN interface (2.03 Active)
- 3 16550-style VAR is with IrDA protocol
- 2 Fast I<sup>2</sup>C<sup>---</sup>, 400 kHz
- 2 charme's for SPI™, SSI™, or Microwire™
- 8/19 bit EMI bus on 128 packages
- ປະການ 80 I/O pins (muxed with interfaces)
- 5 V tolerant, 16 have high sink current (8 mA)
- Bit-wise manipulation of pins within a port
- 16-bit standard timers (TIM)
  - 4 timers each with 2 input capture, 2 output compare, PWM and pulse count modes
- 3-Phase induction motor controller (IMC)
  - 3 pairs of PWM outputs, adjustable centers
  - Emergency stop, dead-time gen, tach input
- JTAG interface with boundary scan
  - ARM EmbeddedICE® RT for debugging
  - In-System Programming (ISP) of Flash
- Embedded trace module (ARM ETM9)
  - Hi-speed instruction tracing, 9-pin interface

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# 1 Introduction

STR91xF is a series of ARM-powered microcontrollers which combines a 16/32-bit ARM966E-S RISC processor core, dual-bank Flash memory, large SRAM for data or code, and a rich peripheral set to form an ideal embedded controller for a wide variety of applications such as point-of-sale terminals, industrial automation, security and surveillance, vending machines, communication gateways, serial protocol conversion, and medical equipment. The ARM966E-S core can perform single-cycle DSP instructions, good for speech processing, audio algorithms, and low-end imaging.

This datasheet provides STR91xF ordering information, functional overview, mechanical information, and electrical device characteristics.

For complete information on STR91xF memory, registers, and peripherals, please refer to the STR91xF Reference Manual.

For information on programming the STR91xF Flash memory please refer to the STR9 Flash Programming Reference Manual

For information on the ARM966E-S core, please refer to the ARM966E-S Rev. 2 Technical Reference Manual.

<b>W32X6</b>	M42X6		311	STR912F		
32		M42X6 M44X6		W44X6		
	256+32 512+32		256+32 512+3			
ļ	9	65	9	96		
Kbytes         OF           Peripheral functions         CAN, 40 I/Os         CAN, EMI, 80 I/Os			Ethernet, USB, CAN, EMI, 80 I/Os			
LQFP128	LQF	P80	LQF	P128		
ducti						

#### Table 1. Device summary





# 2 Functional overview

# 2.1 System-in-a-Package (SiP)

The STR91xF is a SiP device, comprised of two stacked die. One die is the ARM966E-S CPU with peripheral interfaces and analog functions, and the other die is the burst Flash. The two die are connected to each other by a custom high-speed 32-bit burst memory interface and a serial JTAG test/programming interface.

# 2.2 Package choice

STR91xF devices are available in 128-pin (14 x 14 mm) and 80-pin (12 x 12 mm) LQFP packages. Refer to the *Table 1 on page 6* and to *Table 31 on page 70* for a list of available peripherals for each of the package choices.

# 2.3 ARM966E-S CPU core



The ARM966E-S core inherently has separate instruction and data memory interfaces (Harvard architecture), allowing the CPU to simultaneously fetch an instruction, and read or write a data item through two Tightly-Coupled Memory (TCM) interfaces as shown in *Figure 1*. The result is streamlined CPU Load and Store operations and a significant reduction in cycle count per instruction. In addition to this, a 5-stage pipeline is used to increase the amount of operational parallelism, giving the most performance out of each clock cycle.

Ten DSP-enhanced instruction extensions are supported by this core, including single-cycle execution of 32x16 Multiply-Accumulate, saturating addition/subtraction, and count leading-zeros.

The ARM966E-S core is binary compatible with 32-bit ARM7 code and 16-bit Thumb<sup>®</sup> code.

# 2.4 Burst Flash memory interface

A Burst Flash memory interface (*Figure 1*) has been integrated into the Instruction TCM (I-TCM) path of the ARM966E-S core. Also in this path is a 4-instruction Pre-Fetch Queue (PFQ) and a 4-entry Branch Cache (BC), enabling the ARM966E-S core to perform up to 96 MIPS while executing code directly from Flash memory. This architecture provides high performance levels without a costly instruction SRAM, instruction cache, or external SDRAM. Eliminating the instruction cache also means interrupt latency is reduced and code execution becomes more deterministic.

### 2.4.1 Pre-Fetch Queue (PFQ)

As the CPU core accesses sequential instructions through the I-TCM, the PFQ always looks ahead and will pre-fetch instructions, taking advantage any idle bus cycles due to variable length instructions. The PFQ will fetch 32-bits at a time from the Burst Flash memory at a rate of up to 96 MHz.



### 2.4.2 Branch Cache (BC)

When instruction addresses are not sequential, such as a program branch situation, the PFQ would have to flush and reload which would cause the CPU to stall if no BC were present. Before reloading, the PFQ checks the BC to see if it contains the desired target branch address. The BC contains up to four of the most recently taken branch addresses and the first four instructions associated with each of these branches. This check is extremely fast, checking all four BC entries simultaneously for a branch address match (cache hit). If there is a hit, the BC rapidly supplies the instruction and reduces the CPU stall. This gives the PFQ time to start pre-fetching again while the CPU consumes these four instructions from the BC. The advantage here is that program loops (very common with embedded control applications) run very fast if the address of the loops are contained in the BC.

In addition, there is a 5th branch cache entry that is dedicated to the Vectored Interrupt Controller (VIC) to further reduce interrupt latency by eliminating the stall latency typically imposed by fetching the instruction that reads the interrupt vector address from the VIC.

#### 2.4.3 Management of literals

Typical ARM architecture and compilers do not place literals (data constants) sequentially in Flash memory with the instructions that use them, but instead the literals are placed at some other address which looks like a program branch from the PFQ's point of view. The STR91xF implementation of the ARM966E-S core has special circuitry to prevent flushing the PFQ when literals are encountered in program flow to keep performance at a maximum.



#### Figure 1. STR91xF block diagram



# 2.5 SRAM (64K or 96K Bytes)

A 32-bit wide SRAM resides on the CPU's Data TCM (D-TCM) interface, providing single-cycle data accesses. As shown in *Figure 1*, the D-TCM shares SRAM access with the Advanced High-performance Bus (AHB). Sharing is controlled by simple arbitration logic to allow the DMA unit on the AHB to also access to the SRAM.

### 2.5.1 Arbitration

Zero-wait state access occurs for either the D-TCM or the AHB when only one of the two is requesting SRAM. When both request SRAM simultaneously, access is granted on an interleaved basis so neither requestor is starved, granting one 32-bit word transfer to each requestor before relinquishing SRAM to the other. When neither the D-TCM or the AHB are requesting SRAM, the arbiter leaves access granted to the most recent user (if D-TCM was last to use SRAM then the D-TCM will not have to arbitrate to get access next time).

The CPU may execute code from SRAM through the AHB. There are no wait states as long as the D-TCM is not contending for SRAM access and the AHB is not sharing bandwidth with peripheral traffic. The ARM966E-S CPU core has a small pre-fetch queue built into this instruction path through the AHB to look ahead and fetch instructions during idle bus cycles.

### 2.5.2 Battery backup

When a battery is connected to the designated battery backup pin (VBATT), SRAM contents are automatically preserved when the normal operating voltage on VDD pins is lost or sags below threshold. Automatic switchover to SRAM can be disabled by firmware if it is desired that the battery will power only the RTC and not the SRAM during standby.

# 2.6 DMA data movement

DMA channels on the Advanced High-performance Bus (AHB) take full advantage of the separate data path provided by the Harvard architecture, moving data rapidly and largely independent of the instruction path. There are two DMA units, one is dedicated to move data between the Ethernet interface and SRAM, the other DMA unit has eight programmable channels with 16 request signals to service other peripherals and interfaces (USB, SSP, I2C, UART, Timers, EMI, and external request pins). Both single word and burst DMA transfers are supported. Memory-to-memory transfers are supported in addition to memory-peripheral transfers. DMA access to SRAM is shared with D-TCM accesses, and arbitration is described in *Section 2.5.1*. Efficient DMA transfers are managed by firmware using linked list descriptor tables. Of the 16 DMA request signals, two are assigned to external inputs. The DMA unit can move data between external devices and resources inside the STR91xF through the EMI bus.

# Non-volatile memories

There are two independent 32-bit wide Burst Flash memories enabling true read-while-write operation. The Flash memories are single-voltage erase/program with 20 year minimum data retention and 100K minimum erase cycles. The primary Flash memory is much larger than the secondary Flash.

Both Flash memories are blank when devices are shipped from ST. The CPU can boot only from Flash memory (configurable selection of which Flash bank).



Flash memories are programmed half-word (16 bits) at a time, but are erased by sector or by full array.

#### 2.7.1 Primary Flash memory

Using the STR91xF device configuration software tool and 3rd party Integrated Developer Environments, it is possible to specify that the primary Flash memory is the default memory from which the CPU boots at reset, or otherwise specify that the secondary Flash memory is the default boot memory. This choice of boot memory is non-volatile and stored in a location that can be programmed and changed only by JTAG In-System Programming. See *Section 5: Memory mapping*, for more detail.

The primary Flash memory has equal length 64K byte sectors. Devices with 256 Kbytes of primary Flash have four sectors and 512K devices have eight sectors.

#### 2.7.2 Secondary Flash memory

The smaller of the two Flash memories can be used to implement a bootloader, capable of storing code to perform robust In-Application Programming (IAP) of the primary Flash memory. The CPU executes code from the secondary Flash, while updating code in the primary Flash memory. New code for the primary Flash memory can be downloaded over any of the interfaces on the STR91xF (USB, Ethernet, CAN, UART, etc.)

Additionally, the Secondary Flash memory may also be used to store small data sets by emulating EEPROM though firmware, eliminating the need for external EEPROM memories. This raises the data security level because passcodes and other sensitive information can be securely locked inside the STR91xF device.

The secondary Flash memory is 32 Kbytes and has four equal length sectors of 8 Kbytes each.

Both the primary Flash memory and the secondary Flash memory can be programmed with code and/or data using the JTAG In-System Programming (ISP) channel, totally independent of the CPU. This is excellent for iterative code development and for manufacturing.

# 2.8 One-time-programmable (OTP) memory

There are 32 bytes of OTP memory ideally suited for serial numbers, security keys, factory calibration constants, or other permanent data constants. These OTP data bytes can be programmed only one time through either the JTAG interface or by the CPU, and these bytes can never be altered afterwards. As an option, a "lock bit" can be set by the JTAG interface or the CPU which will block any further writing to the this OTP area. The "lock bit" itself is also OTP. If the OTP array is unlocked, it is always possible to go back and write to an OTP byte location that has not been previously written, but it is never possible to change an OTP byte location if any one bit of that particular byte has been written before. The last two OTP bytes are reserved for the STR91xF product ID and revision level. Byte 30 contains the device revision level. For STR91xF devices, the revision is 0x13.

# 2.9 Vectored interrupt controller (VIC)

Interrupt management in the STR91xF is implemented from daisy-chaining two standard ARM VIC units. This combined VIC has 32 prioritized interrupt request channels and generates two



interrupt output signals to the CPU. The output signals are FIQ and IRQ, with FIQ having higher priority.

#### 2.9.1 FIQ handling

FIQ (Fast Interrupt reQuest) is the only non-vectored interrupt and the CPU can execute an Interrupt Service Routine (ISR) directly without having to determine/prioritize the interrupt source, minimizing ISR latency. Typically only one interrupt source is assigned to FIQ. An FIQ interrupt has its own set of banked registers to minimize the time to make a context switch. Any of the 32 interrupt request input signals coming into the VIC can be assigned to FIQ.

#### 2.9.2 IRQ handling

IRQ is a vectored interrupt and is the logical OR of all 32 interrupt request signals coming into the 32 IRQ channels. Priority of individual vectored interrupt requests is determined by hardware (IRQ channel Intr 0 is highest priority, IRQ channel Intr 31 is lowest). However, CPU firmware may re-assign individual interrupt sources to individual hardware IRQ channels, meaning that firmware can effectively change interrupt priority levels as needed.

When the IRQ signal is activated by an interrupt request, VIC hardware will resolve the IRQ interrupt priority, then the ISR reads the VIC to determine both the interrupt source and the vector address to jump to the service code.

The STR91xF has a feature to reduce ISR response time for IRQ interrupts. Typically, it requires two memory accesses to read the interrupt vector address from the VIC, but the STR91xF reduces this to a single access by adding a 5th entry in the instruction branch cache, dedicated for interrupts. This 5th cache entry always holds the instruction that reads the interrupt vector address from the VIC, eliminating one of the memory accesses typically required in traditional ARM implementations.

#### 2.9.3 Interrupt sources

The 32 interrupt request signals coming into the VIC on 32 IRQ channels are from various sources; 5 from a wake-up unit and the remaining 27 come from internal sources on the STR91xF such as on-chip peripherals, see *Table 2*. Optionally, firmware may force an interrupt on any IRQ channel.

One of the 5 interrupt requests generated by the wake-up unit (IRQ25 in *Table 2*) is derived from the logical OR of all 32 inputs to the wake-up unit. Any of these 32 inputs may be used to wake up the CPU and cause an interrupt. These 32 inputs consist of 30 external interrupts on selected and enabled GPIO pins, plus the RTC interrupt, and the USB Resume interrupt.

Each of 4 remaining interrupt requests generated by the wake-up unit (IRQ26 in *Table 2*) are derived from groupings of 8 interrupt sources. One group is from GPIO pins P3.2 to P3.7 plus the RTC interrupt and the USB Resume interrupt; the next group is from pins P5.0 to P5.7; the next group is from pins P6.0 to P6.7; and last the group is from pins P7.0 to P7.7. This allows individual pins to be assigned directly to vectored IRQ interrupts or one pin assigned directly to the non-vectored FIQ interrupt.



See *Table 2* for recommended interrupt source assignments to physical IRQ interrupt channels. Interrupt source assignments are made by CPU firmware during initialization, thus establishing interrupt priorities.

	VIC IRQ Channel	Logic Block	Interrupt Source
	0 (high priority)	WatchDog	Timeout in WDT mode, Terminal Count in Counter Mode
ľ	1	CPU Firmware	Firmware generated interrupt
·	2	CPU Core	Debug Receive Command
	3	CPU Core	Debug Transmit Command
	4	TIM Timer 0	Logic OR of ICI0_0, ICI0_1, OCI0_0, OCI0_1, Timer overflow
Ī	5	TIM Timer 1	Logic OR of ICI1_0, ICI1_1, OCI1_0, OCI1_1, Timer overflow
Ī	6	TIM Timer 2	Logic OR of ICI2_0, ICI2_1, OCI2_0, OCI2_1, Timer overflow
Ī	7	TIM Timer 3	Logic OR of ICI3_0, ICI3_1, OCI3_0, OCI3_1, Timer overflow
Ī	8	USB	Logic OR of high priority USB interrupts
Ī	9	USB	Logic OR of low priority USB interrupts
ľ	10	CCU	Logic OR of all interrupts from Clock Control Unit
	11	Ethernet MAC	Logic OR of Ethernet MAC interrupts via its own dedicated DMA channel.
	12	DMA	Logic OR of interrupts from each of the 8 individual DMA channels
Ī	13	CAN	Logic OR of all CAN interface interrupt sources
Ī	14	IMC	Logic OR of 8 Induction Motor Control Unit interrupts
Ī	15	ADC	End of AtoD conversion interrupt
Ī	16	UART0	Logic OR of 5 interrupts from UART channel 0
Ī	17	UART1	Logic OR of 5 interrupts from UART channel 1
Ī	18	UART2	Logic OR of 5 interrupts from UART channel 2
Ī	19	I2C0	Logic OR of transmit, receive, and error interrupts of I2C channel 0
-	20	I2C1	Logic OR of transmit, receive, and error interrupts of I2C channel 1
Ī	21	SSP0	Logic OR of all interrupts from SSP channel 0
ſ	22	SSP1	Logic OR of all interrupts from SSP channel 1
ľ	23	BROWNOUT	LVD warning interrupt
Ī	24	RTC	Logic OR of Alarm, Tamper, or Periodic Timer interrupts
-	25	Wake-Up (all)	Logic OR of all 32 inputs of Wake-Up unit (30 pins, RTC, and USB Resume)
	26	Wake-up Group 0	Logic OR of 8 interrupt sources: RTC, USB Resume, pins P3.2 to P3.7
	27	Wake-up Group 1	Logic OR of 8 interrupts from pins P5.0 to P5.7
	28	Wake-up Group 2	Logic OR of 8 interrupts from pins P6.0 to P6.7
	29	Wake-up Group 3	Logic OR of 8 interrupts from pins P7.0 to P7.7
ľ	30	USB	USB Bus Resume Wake-up (also input to wake-up unit)
	31 (low priority)	PFQ-BC	Special use of interrupts from Prefetch Queue and Branch Cache

 Table 2.
 Recommended IRQ Channel assignments (set by CPU firmware)



# 2.10 Clock control unit (CCU)

The CCU generates a master clock of frequency  $f_{MSTR}$ . From this master clock the CCU also generates individually scaled and gated clock sources to each of the following functional blocks within the STR91xF.

- CPU, f<sub>CPUCLK</sub>
- Advanced High-performance Bus (AHB), f<sub>HCLK</sub>
- Advanced Peripheral Bus (APB), f<sub>PCLK</sub>
- Flash Memory Interface (FMI), f<sub>FMICLK</sub>
- External Memory Interface (EMI), f<sub>BCLK</sub>
- UART Baud Rate Generators, f<sub>BAUD</sub>
- USB, f<sub>USB</sub>

#### 2.10.1 Master clock sources

The master clock in the CCU ( $f_{MSTR}$ ) is derived from one of three clock input sources. Under firmware control, the CPU can switch between the three CCU inputs without introducing any glitches on the master clock output. Inputs to the CCU are:

- Main Oscillator (f<sub>OSC</sub>). The source for the main oscillator input is a 4 to 25 MHz external crystal connected to STR91xF pins X1\_CPU and X2\_CPU, or an external oscillator device connected to pin X1\_CPU.
- PLL (f<sub>PLL</sub>). The PLL takes the 4 to 25 MHz oscillator clock as input and generates a master clock output up to 96 MHz (programmable). By default, at power-up the master clock is sourced from the main oscillator until the PLL is ready (locked) and then the CPU may switch to the PLL source under firmware control. The CPU can switch back to the main oscillator source at any time and turn off the PLL for low-power operation. The PLL is always turned off in Sleep mode.
- RTC (f<sub>RTC</sub>). A 32.768 kHz external crystal can be connected to pins X1\_RTC and X2\_RTC, or an external oscillator connected to pin X1\_RTC to constantly run the real-time clock unit. This 32.768 kHz clock source can also be used as an input to the CCU to run the CPU in slow clock mode for reduced power.

As an option, there are a number of peripherals that do not have to receive a clock sourced from the CCU. The USB interface can receive an external clock on pin P2.7, TIM timers TIM0/ TIM1 can receive an external clock on pin P2.4, and timers TIM2/TIM3 on pin P2.5.



#### Figure 2. Clock control

### 2.10.2 Reference clock (RCLK)

The main clock (f<sub>MSTR</sub>) can be divided to operate at a slower frequency reference clock (RCLK) for the ARM core and all the peripherals. The RCLK provides the divided clock for the ARM core, and feeds the dividers for the AHB, APB, External Memory Interface, and FMI units.

#### 2.10.3 AHB clock (HCLK)

The RCLK can be divided by 1, 2 or 4 to generate the AHB clock. The AHB clock is the bus clock for the AHB bus and all bus transfers are synchronized to this clock. The maximum HCLK frequency is 96 MHz.

### 2.10.4 APB clock (PCLK)

The RCLK can be divided by 1, 2, 4 or 8 to generate the APB clock. The APB clock is the bus clock for the APB bus and all bus transfers are synchronized to this clock. Many of the peripherals that are connected to the AHB bus also use the PCLK as the source for external bus data transfers. The maximum PCLK frequency is 48 MHz.

#### 2.10.5 Flash memory interface clock (FMICLK)

The FMICLK clock is an internal clock derived from RCLK, defaulting to RCLK frequency at power up. The clock can be optionally divided by 2. The FMICLK determines the bus bandwidth between the ARM core and the Flash memory. Typically, codes in the Flash memory can be fetched one word per FMICLK clock in burst mode. The maximum FMICLK frequency is 96MHz.



### 2.10.6 Baud rate clock (BRCLK)

The baud rate clock is an internal clock derived from  $f_{MSTR}$  that is used by the three on-chip UART peripherals for baudrate generation. The frequency can be optionally divided by 2.

#### 2.10.7 External memory interface bus clock (BCLK)

The BCLK is an internal clock that controls the EMI bus. All EMI bus signals are synchronized to the BCLK. The BCLK is derived from the HCLK and the frequency can be configured to be the same or half that of the HCLK. Refer to *Table 9 on page 50* for the maximum BCLK frequency ( $f_{BCLK}$ ).

#### 2.10.8 USB interface clock

Special consideration regarding the USB interface: The clock to the USB interface must operate at 48 MHz and comes from one of three sources, selected under firmware control:

- CCU master clock output of 48 MHz.
- CCU master clock output of 96 MHz. An optional divided-by-two circuit is available to produce 48 MHz for the USB while the CPU system runs at 96MHz.
- STR91xF pin P2.7. An external 48 MHz oscillator connected to pin P2.7 can directly source the USB while the CCU master clock can run at some frequency other than 48 or 96 MHz.

### 2.10.9 Ethernet MAC clock

Special consideration regarding the Ethernet MAC: The external Ethernet PHY interface device requires it's own 25 MHz clock source. This clock can come from one of two sources:

- A 25 MHz clock signal coming from a dedicated output pin (P5.2) of the STR91xF. In this case, the STR91xF must use a 25 MHz signal on its main oscillator input in order to pass this 25 MHz clock back out to the PHY device through pin P5.2. The advantage here is that an inexpensive 25 MHz crystal may be used to source a clock to both the STR91xF and the external PHY device.
- An external 25 MHz oscillator connected directly to the external PHY interface device. In this case, the STR91xF can operate independent of 25 MHz.

#### 2.10.10 External RTC calibration clock

The RTC\_CLK ( $f_{RTC/8}$ ) can be enabled as an output on the JRTCK pin. The RTC\_CLK is used for RTC oscillator calibration. The RTC\_CLK is active in Sleep mode and can be used as a system wake up control clock.

### 2.10.11 Operation example

As an example of CCU operation, a 25 MHz crystal can be connected to the main oscillator input on pins X1\_CPU and X2\_CPU, a 32.768 kHz crystal connected to pins X1\_RTC and X2\_RTC, and the clock input of an external Ethernet PHY device is connected to STR91xF output pin P5.2. In this case, the CCU can run the CPU at 96 MHz from PLL, the USB interface at 48 MHz, and the Ethernet interface at 25 MHz. The RTC is always running in the background at 32.768 kHz, and the CPU can go to very low power mode dynamically by running from 32.768 kHz and shutting off peripheral clocks and the PLL as needed.



# 2.11 Flexible power management

The STR91xF offers configurable and flexible power management control that allows the user to choose the best power option to fit the application. Power consumption can be dynamically managed by firmware and hardware to match the system's requirements. Power management is provided via clock control to the CPU and individual peripherals.

Clocks to the CPU and peripherals can be individually divided and gated off as needed. In addition to individual clock divisors, the CCU master clock source going to the CPU, AHB, APB, EMI, and FMI can be divided dynamically by as much as 1024 for low power operation. Additionally, the CCU may switch its input to the 32.768 kHz RTC clock at any time for low power.

The STR91xF supports the following three global power control modes:

- Run Mode: All clocks are on with option to gate individual clocks off via clock mask registers.
- Idle Mode: CPU and FMI clocks are off until an interrupt, reset, or wake-up occurs. Preconfigured clock mask registers selectively allow individual peripheral clocks to continue run during Idle Mode.
- Sleep Mode: All clocks off except RTC clock. Wake up unit remains powered, PLL is forced off.

A special mode is used when JTAG debug is active which never gates off any clocks even if the CPU enters Idle or Sleep mode.

#### 2.11.1 Run mode

This is the default mode after any reset occurs. Firmware can gate off or scale any individual clock. Also available is a special Interrupt Mode which allows the CPU to automatically run full speed during an interrupt service and return back to the selected CPU clock divisor rate when the interrupt has been serviced. The advantage here is that the CPU can run at a very low frequency to conserve power until a periodic wake-up event or an asynchronous interrupt occurs at which time the CPU runs full speed immediately.

#### 2.11.2 Idle mode

In this mode the CPU suspends code execution and the CPU and FMI clocks are turned off immediately after firmware sets the Idle Bit. Various peripherals continue to run based on the settings of the mask registers that exist just prior to entering Idle Mode. There are 3 ways to exit Idle Mode and return to Run Mode:

Any reset (external reset pin, watchdog, low-voltage, power-up, JTAG debug command)

Any interrupt (external, internal peripheral, RTC alarm or interval)

Input from wake-up unit on GPIO pins

Note:

It is possible to remain in Idle Mode for the majority of the time and the RTC can be programmed to periodically wake up to perform a brief task or check status.

#### 2.11.3 Sleep mode

In this mode all clock circuits except the RTC are turned off and main oscillator input pins X1\_CPU and X2\_CPU are disabled. The RTC clock is required for the CPU to exit Sleep Mode. The entire chip is quiescent (except for RTC and wake-up circuitry). There are three means to exit Sleep Mode and re-start the system:



- Some resets (external reset pin, low-voltage, power-up, JTAG debug command)
- RTC alarm
- Input from wake-up unit

# 2.12 Voltage supplies

The STR91xF requires two separate operating voltage supplies. The CPU and memories operate from a 1.65V to 2.0V on the VDD pins, and the I/O ring operates at 2.7V to 3.6V on the VDDQ pins.

#### 2.12.1 Independent A/D converter supply and reference voltage

The ADC unit on 128-pin packages has an isolated analog voltage supply input at pin AVDD to accept a very clean voltage source, independent of the digital voltage supplies. The analog voltage supply range on pin AVDD is the same range as the digital voltage supply on pin VDDQ. Additionally, an isolated analog supply ground connection is provided on pin AVSS only on 128-pin packages for further ADC supply isolation. On 80-pin packages, the analog voltage supply is shared with the ADC reference voltage pin (as described next), and the analog ground is shared with the digital ground at a single point in the STR91xF device on pin AVSS\_VSSQ.

A separate external analog reference voltage input for the ADC unit is available on 128-pin packages at the AVREF pin for better accuracy on low voltage inputs, and the voltage on AVREF can range from 1.0V to  $V_{DDQ}$ . For 80-pin packages, the ADC reference voltage is tied internally to the ADC unit supply voltage at pin AVCC\_AVREF, meaning the ADC reference voltage is fixed to the ADC unit supply voltage.

#### 2.12.2 Battery supply

An optional stand-by voltage from a battery or other source may be connected to pin VBATT to retain the contents of SRAM in the event of a loss of the  $V_{DD}$  supply. The SRAM will automatically switch its supply from the internal  $V_{DD}$  source to the VBATT pin when the voltage of  $V_{DD}$  drops below that of VBATT.

The VBATT pin also supplies power to the RTC unit, allowing the RTC to function even when the main digital supplies ( $V_{DD}$  and  $V_{DDQ}$ ) are switched off. By configuring the RTC register, it is possible to select whether or not to power from VBATT only the RTC unit, or power the RTC unit and the SRAM when the STR91xF device is powered off.



**Dbsolete** 



# 2.13 System supervisor

The STR91xF monitors several system and environmental inputs and will generate a global reset, a system reset, or an interrupt based on the nature of the input and configurable settings. A global reset clears all functions on the STR91xF, a system reset will clear all but the Clock Control Unit (CCU) settings and the system status register. At any time, firmware may reset individual on-chip peripherals. System supervisor inputs include:

- GR: CPU voltage supply (V<sub>DD</sub>) drop out or brown out
- GR: I/O voltage supply (V<sub>DDQ</sub>) drop out or brown out
- GR: Power-Up condition
- SR: Watchdog timer timeout
- SR: External reset pin (RESET\_INn)
- SR: JTAG debug reset command
- Note: GR: means the input causes Global Reset, SR: means the input causes System Reset

The CPU may read a status register after a reset event to determine if the reset was caused by a watchdog timer timeout or a voltage supply drop out. This status register is cleared only by a power up reset.

#### 2.13.1 Supply voltage brownout

Each operating voltage source ( $V_{DD}$  and  $V_{DDQ}$ ) is monitored separately by the Low Voltage Detect (LVD) circuitry. The LVD will generate an early warning interrupt to the CPU when voltage sags on either  $V_{DD}$  or  $V_{DDQ}$  voltage inputs. This is an advantage for battery powered applications because the system can perform an orderly shutdown before the batteries become too weak. The voltage trip point to cause a brown out interrupt is typically 0.25V above the LVD dropout thresholds that cause a reset.

CPU firmware may prevent all brown-out interrupts by writing to interrupt mask registers at runtime.

### 2.13.2 Supply voltage dropout

LVD circuitry will always cause a global reset if the CPU's V<sub>DD</sub> source drops below it's fixed threshold of 1.4V.

However, the LVD trigger threshold to cause a global reset for the I/O ring's  $V_{DDQ}$  source is set to one of two different levels, depending if  $V_{DDQ}$  will be operated in the range of 2.7V to 3.3V, or 3.0V to 3.6V. If  $V_{DDQ}$  operation is at 2.7V to 3.3V, the LVD dropout trigger threshold is 2.4V. If  $V_{DDQ}$  operation is 3.0V and 3.6V, the LVD threshold is 2.7V. The choice of trigger level is made by STR91xF device configuration software from STMicroelectronics or IDE from 3rd parties, and is programmed into the STR91xF device along with other configurable items through the JTAG interface when the Flash memory is programmed.

CPU firmware may prevent some LVD resets if desired by writing a control register at run-time. Firmware may also disable the LVD completely for lowest-power operation when an external LVD device is being used.

#### 2.13.3 Watchdog timer

The STR91xF has a 16-bit down-counter (not one of the four TIM timers) that can be used as a watchdog timer or as a general purpose free-running timer/counter. The clock source is the



peripheral clock from the APB, and an 8-bit clock pre-scaler is available. When enabled by firmware as a watchdog, this timer will cause a system reset if firmware fails to periodically reload this timer before the terminal count of 0x0000 occurs, ensuring firmware sanity. The watchdog function is off by default after a reset and must be enabled by firmware.

#### 2.13.4 External RESET\_INn pin

This input signal is active-low with hystereses ( $V_{RHYS}$ ). Other open-drain, active-low system reset signals on the circuit board (such as closure to ground from a push-button) may be connected directly to the RESET\_INn pin, but an external pull-up resistor to  $V_{DDQ}$  must be present as there is no internal pullup on the RESET\_INn pin.

A valid active-low input signal of  $t_{RINMIN}$  duration on the RESET\_INn pin will cause a system reset within the STR91xF. There is also a RESET\_OUTn pin on the STR91xF that can drive other system components on the circuit board. RESET\_OUTn is active-low and has the same timing of the Power-On-Reset (POR) shown next,  $t_{POR}$ .

#### 2.13.5 Power-up

The LVD circuitry will always generate a global reset when the STR91xF powers up, meaning internal reset is active until  $V_{DDQ}$  and  $V_{DD}$  are both above the LVD thresholds. This POR condition has a duration of  $t_{POR}$ , after which the CPU will fetch its first instruction from address 0x0000.0000 in Flash memory. It is not possible for the CPU to boot from any other source other than Flash memory.

### 2.13.6 JTAG debug command

When the STR91xF is in JTAG debug mode, an external device which controls the JTAG interface can command a system reset to the STR91xF over the JTAG channel.

#### 2.13.7 Tamper detection

On 128-pin STR91xF devices only, there is a tamper detect input pin, TAMPER\_IN, used to detect and record the time of a tamper event on the end product such as malicious opening of an enclosure, unwanted opening of a panel, etc. The activation mode of the tamper pin is programmable to one of two modes. One is Normally Closed/Tamper Open, the other mode will detect when a signal on the tamper input pin is driven from low-to-high, or high-to-low depending on firmware configuration. Once a tamper event occurs, the RTC time (millisecond resolution) and the date are recorded in the RTC unit. Simultaneously, the SRAM standby voltage source will be cut off to invalidate all SRAM contents. Tamper detection control and status logic are part of the RTC unit.

# Real-time clock (RTC)

The RTC combines the functions of a complete time-of-day clock (millisecond resolution) with an alarm programmable up to one month, a 9999-year calender with leap-year support, periodic interrupt generation from 1 to 512 Hz, tamper detection (described in *Section 2.13.7*), and an optional clock calibration output on the JRTCK pin. The time is in 24 hour mode, and time/calendar values are stored in binary-coded decimal format.

The RTC also provides a self-isolation mode that is automatically activated during power down. This feature allows the RTC to continue operation when  $V_{DDO}$  and  $V_{DD}$  are absent, as long as



an alternate power source, such as a battery, is connected to the VBATT input pin. The current drawn by the RTC unit on the VBATT pin is very low in this standby mode, I<sub>RTC STBY</sub>.

# 2.15 JTAG interface

An IEEE-1149.1 JTAG interface on the STR91xF provides In-System-Programming (ISP) of all memory, boundary scan testing of pins, and the capability to debug the CPU.

STR91xF devices are shipped from ST with blank Flash memories. The CPU can only boot from Flash memory (selection of which Flash bank is programmable). Firmware must be initially programmed through JTAG into one of these Flash memories before the STR91xF is used.

Six pins are used on this JTAG serial interface. The five signals JTDI, JTDO, JTMS, JTCK, and JTRSTn are all standard JTAG signals complying with the IEEE-1149.1 specification. The sixth signal, JRTCK (Return TCK), is an output from the STR91xF and it is used to pace the JTCK clock signal coming in from the external JTAG test equipment for debugging. The frequency of the JTCK clock signal coming from the JTAG test equipment must be at least 10 times less than the ARM966E-S CPU core operating frequency (f<sub>CPUCLK</sub>). To ensure this, the signal JRTCK is output from the STR91xF and is input to the external JTAG test equipment to hold off transitions of JTCK until the CPU core is ready, meaning that the JTAG equipment cannot send the next rising edge of JTCK until the equipment receives a rising edge of JRTCK from the STR91xF. The JTAG test equipment must be able to interpret the signal JRTCK and perform this adaptive clocking function. If it is known that the CPU clock will always be at least ten times faster than the incoming JTCK clock signal, then the JRTCK signal is not needed.

The two die inside the STR91xF (CPU die and Flash memory die) are internally daisy-chained on the JTAG bus, see *Figure 3 on page 22*. The CPU die has two JTAG Test Access Ports (TAPs), one for boundary scan functions and one for ARM CPU debug. The Flash memory die has one TAP for program/erase of non-volatile memory. Because these three TAPs are daisychained, only one TAP will converse on the JTAG bus at any given time while the other two TAPs are in BYPASS mode. The TAP positioning order within this JTAG chain is the boundary scan TAP first, followed by the ARM debug TAP, followed by the Flash TAP. All three TAP controllers are reset simultaneously by one of two methods:

- A chip-level global reset, caused only by a Power-On-Reset (POR) or a Low Voltage Detect (LVD).
- A reset command issued by the external JTAG test equipment. This can be the assertion of the JTAG JTRSTn input pin on the STR91xF or a JTAG reset command shifted into the STR91xF serially.

This means that chip-level system resets from watchdog time-out or the assertion of RESET\_INn pin do not affect the operation of any JTAG TAP controller. Only global resets effect the TAPs.





Figure 3. JTAG chaining inside the STR91xF

### 2.15.1 In-system-programming

The JTAG interface is used to program or erase all memory areas of the STR91xF device. The pin RESET\_INn must be asserted during ISP to prevent the CPU from fetching invalid instructions while the Flash memories are being programmed.

Note that the 32 bytes of OTP memory locations cannot be erased by any means once programmed by JTAG ISP or the CPU.

# 2.15.2 Boundary scan

Standard JTAG boundary scan testing compliant with IEEE-1149.1 is available on the majority of pins of the STR91xF for circuit board test during manufacture of the end product. STR91xF pins that are not serviced by boundary scan are the following:

- JTAG pins JTCK, JTMS, JTDI, JTDO, JTRSTn, JRTCK
- Oscillator input pins X1 CPU, X2 CPU, X1 RTC, X2 RTC
- Tamper detect input pin TAMPER\_IN (128-pin packages only)

### 2.15.3 CPU debug

The ARM966E-S CPU core has standard ARM EmbeddedICE-RT logic, allowing the STR91xF to be debugged through the JTAG interface. This provides advanced debugging features making it easier to develop application firmware, operating systems, and the hardware itself.



Debugging requires that an external host computer, running debug software, is connected to the STR91xF target system via hardware which converts the stream of debug data and commands from the host system's protocol (USB, Ethernet, etc.) to the JTAG EmbeddedICE-RT protocol on the STR91xF. These protocol converters are commercially available and operate with debugging software tools.

The CPU may be forced into a Debug State by a breakpoint (code fetch), a watchpoint (data access), or an external debug request over the JTAG channel, at which time the CPU core and memory system are effectively stopped and isolated from the rest of the system. This is known as Halt Mode and allows the internal state of the CPU core, memory, and peripherals to be examined and manipulated. Typical debug functions are supported such as run, halt, and single-step. The EmbeddedICE-RT logic supports two hardware compare units. Each can be configured to be either a watchpoint or a breakpoint. Breakpoints can also be data-dependent.

Debugging (with some limitations) may also occur through the JTAG interface while the CPU is running full speed, known as Monitor Mode. In this case, a breakpoint or watchpoint will not force a Debug State and halt the CPU, but instead will cause an exception which can be tracked by the external host computer running monitor software. Data can be sent and received over the JTAG channel without affecting normal instruction execution. Time critical code, such as Interrupt Service Routines may be debugged real-time using Monitor Mode.

#### 2.15.4 JTAG security bit

This is a non-volatile bit (Flash memory based), which when set will not allow the JTAG debugger or JTAG programmer to read the Flash memory contents.

Using JTAG ISP, this bit is typically programmed during manufacture of the end product to prevent unwanted future access to firmware intellectual property. The JTAG Security Bit can be cleared only by a JTAG "Full Chip Erase" command, making the STR91xF device blank and ready for programming again. The CPU can read the status of the JTAG Security Bit, but it may not change the bit value.

# 2.16 Embedded trace module (ARM ETM9, v. r2p2)

The ETM9 interface provides greater visibility of instruction and data flow happening inside the CPU core by streaming compressed data at a very high rate from the STR91xF though a small number of ETM9 pins to an external Trace Port Analyzer (TPA) device. The TPA is connected to a host computer using USB, Ethernet, or other high-speed channel. Real-time instruction flow and data activity can be recorded and later formatted and displayed on the host computer running debugger software, and this software is typically integrated with the debug software used for EmbeddedICE-RT functions such as single-step, breakpoints, etc. Tracing may be triggered and filtered by many sources, such as instruction address comparators, data watchpoints, context ID comparators, and counters. State sequencing of up to three triggers is also provided. TPA hardware is commercially available and operates with debugging software tools.

The ETM9 interface is nine pins total, four of which are data lines, and all pins can be used for GPIO after tracing is no longer needed. The ETM9 interface is used in conjunction with the JTAG interface for trace configuration. When tracing begins, the ETM9 engine compresses the data by various means before broadcasting data at high speed to the TPA over the four data lines. The most common ETM9 compression technique is to only output address information when the CPU branches to a location that cannot be inferred from the source code. This means



the host computer must have a static image of the code being executed for decompressing the ETM9 data. Because of this. self-modified code cannot be traced.

#### Ethernet MAC interface with DMA 2.17

STR91xF devices in 128-pin packages provide an IEEE-802.3-2002 compliant Media Access Controller (MAC) for Ethernet LAN communications through an industry standard Medium Independent Interface (MII). The STR91xF requires an external Ethernet physical interface device (PHY) to connect to the physical LAN bus (twisted-pair, fiber, etc.). The PHY is connected to the STR91xF MII port using as many as 18 signals (see pins which have signal names MII \* in Table 3).

The MAC corresponds to the OSI Data Link layer and the PHY corresponds to the OSI Physical layer. The STR91xF MAC is responsible for:

- Data encapsulation, including frame assembly before transmission, and frame parsing/ error detection during and after reception.
- Productls Media access control, including initiation of frame transmission and recover from transmission failure.

The STR91xF MAC includes the following features:

- Supports 10 and 100 Mbps rates •
- Tagged MAC frame support (VLAN support)
- Half duplex (CSMA/CD) and full duplex operation •
- MAC control sublayer (control frames) support
- 32-bit CRC generation and removal
- Several address filtering modes for physical and multicast address (multicast and group addresses)
- 32-bit status code for each transmitted or received frame
- Internal FIFOs to buffer transmit and receive frames. Transmit FIFO depth is 4 words (32 bits each), and the receive FIFO is 16 words deep.

A 32-bit burst DMA channel residing on the AHB is dedicated to the Ethernet MAC for highspeed data transfers, side-stepping the CPU for minimal CPU impact during transfers. This DMA channel includes the following features:

- Direct SRAM to MAC transfers of transmit frames with the related status, by descriptor chain
- Direct MAC to SRAM transfers of receive frames with the related status, by descriptor chain
  - Open and Closed descriptor chain management

#### 2.18 USB 2.0 slave device interface with DMA

The STR91xF provides a USB slave controller that implements both the OSI Physical and Data Link layers for direct bus connection by an external USB host on pins USBDP and USBPN. The USB interface detects token packets, handles data transmission and reception, and processes handshake packets as required by the USB 2.0 standard.

The USB slave interface includes the following features:



- Supports USB low and full-speed transfers (12 Mbps), certified to comply with the USB 2.0 specification
- Supports isochronous, bulk, control, and interrupt endpoints
- Configurable number of endpoints allowing a mixture of up to 20 single-buffered monodirectional endpoints or up to 10 double-buffered bidirectional endpoints
- Dedicated, dual-port 2 Kbyte USB Packet Buffer SRAM. One port of the SRAM is connected by a Packet Buffer Interface (PBI) on the USB side, and the CPU connects to the other SRAM port.
- CRC generation and checking
- NRZI encoding-decoding and bit stuffing
- USB suspend resume operations

#### 2.18.1 Packet buffer interface (PBI)

The PBI manages a set of buffers inside the 2 Kbyte Packet Buffer, both for transmission and reception. The PBI will choose the proper buffer according to requests coming from the USB Serial Interface Engine (SIE) and locate it in the Packet SRAM according to addresses pointed by endpoint registers. The PBI will also auto-increment the address after each exchanged byte until the end of packet, keeping track of the number of exchanged bytes and preventing buffer overrun. Special support is provided by the PBI for isochronous and bulk transfers, implementing double-buffer usage which ensures there is always an available buffer for a USB packet while the CPU uses a different buffer.

#### 2.18.2 DMA

A programmable DMA channel may be assigned by CPU firmware to service the USB interface for fast and direct transfers between the USB bus and SRAM with little CPU involvement. This DMA channel includes the following features:

- Direct USB Packet Buffer SRAM to system SRAM transfers of receive packets, by descriptor chain for bulk or isochronous endpoints.
- Direct system SRAM to USB Packet Buffer SRAM transfers of transmit packets, by descriptor chain for bulk or isochronous endpoints.
- Linked-list descriptor chain support for multiple USB packets

### 2.18.3 Suspend mode

CPU firmware may place the USB interface in a low-power suspend mode when required, and the USB interface will automatically wake up asynchronously upon detecting activity on the USB pins.

# CAN 2.0B interface

The STR91xF provides a CAN interface complying with CAN protocol version 2.0 parts A and B. An external CAN transceiver device connected to pins CAN\_RX and CAN\_TX is required for connection to the physical CAN bus.

The CAN interface manages up to 32 Message Objects and Identifier Masks using a Message SRAM and a Message Handler. The Message Handler takes care of low-level CAN bus activity such as acceptance filtering, transfer of messages between the CAN bus and the Message



2.19

SRAM, handling of transmission requests, and interrupt generation. The CPU has access to the Message SRAM via the Message Handler using a set of 38 control registers.

The follow features are supported by the CAN interface:

- Bitrates up to 1 Mbps
- Disable Automatic Retransmission mode for Time Triggered CAN applications
- 32 Message Objects
- Each Message Object has its own Identifier Mask
- Programmable FIFO mode
- Programmable loopback mode for self-test operation

The CAN interface is not supported by DMA.

### 2.20 UART interfaces with DMA

The STR91xF supports three independent UART serial interfaces, designated UART0, UART1, and UART2. Each interface is very similar to the industry-standard 16C550 UART device. All three UART channels support IrDA encoding/decoding, requiring only an external LED transceiver to pins UARTx\_RX and UARTx\_Tx for communication. One UART channel (UART0) supports full modem control signals.

UART interfaces include the following features:

- Maximum baud rate of 1.5 Mbps
- Separate FIFOs for transmit and receive, each 16 deep, each FIFO can be disabled by firmware if desired
- Programmable FIFO trigger levels between 1/8 and 7/8
- Programmable baud rate generator based on CCU master clock, or CCU master clock divided by two
- Programmable serial data lengths of 5, 6, 7, or 8 bits with start bit and 1 or 2 stop bits
- Programmable selection of even, odd, or no-parity bit generation and detection
- False start-bit detection
- Line break generation and detection
- Support of IrDA SIR ENDEC functions for data rates of up to 115.2K bps
- IrDA bit duration selection of 3/16 or low-power (1.14 to 2.23 µsec)
- Channel UART0 supports modem control functions CTS, DCD, DSR, RTS, DTR, and RI

For your reference, only two standard 16550 UART features are not supported, 1.5 stop bits and independent receive clock.

#### 2.20.1 DMA

A programmable DMA channel may be assigned by CPU firmware to service channels UART0 and UART1 for fast and direct transfers between the UART bus and SRAM with little CPU involvement. Both DMA single-transfers and DMA burst-transfers are supported for transmit and receive. Burst transfers require that UART FIFOs are enabled.



# 2.21 I<sup>2</sup>C interfaces with DMA

The STR91xF supports two independent I2C serial interfaces, designated I2C0, and I2C1. Each interface allows direct connection to an I2C bus as either a bus master or bus slave device (firmware configurable). I2C is a two-wire communication channel, having a bidirectional data signal and a single-directional clock signal based on open-drain line drivers, requiring external pull-up resistors.

Byte-wide data is transferred between a Master device and a Slave device on two wires. More than one bus Master is allowed, but only one Master may control the bus at any given time. Data is not lost when another Master requests the use of a busy bus because I2C supports collision detection and arbitration. More than one Slave device may be present on the bus, each having a unique address. The bus Master initiates all data movement and generates the clock that permits the transfer. Once a transfer is initiated by the Master, any device that is addressed is considered a Slave. Automatic clock synchronization allows I2C devices with different bit rates to communicate on the same physical bus. A single device can play the role of Master or Slave, or a single device can be a Slave only. A Master or Slave device has the ability to suspend data transfers if the device needs more time to transmit or receive data.

Each I2C interface on the STR91xF has the following features:

- Programmable clock supports various rates up to I2C Standard rate (100 KHz) or Fast rate (400 KHz).
- Serial I/O Engine (SIOE) takes care of serial/parallel conversion; bus arbitration; clock generation and synchronization; and handshaking
- Multi-master capability
- 7-bit or 10-bit addressing

#### 2.21.1 DMA

A programmable DMA channel may be assigned by CPU firmware to service each I2C channel for fast and direct transfers between the I2C bus and SRAM with little CPU involvement. Both DMA single-transfers and DMA burst-transfers are supported for transmit and receive.

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# 2.22 SSP interfaces (SPI, SSI, and Microwire) with DMA

The STR91xF supports two independent Synchronous Serial Port (SSP) interfaces, designated SSP0, and SSP1. Primary use of each interface is for supporting the industry standard Serial Peripheral Interface (SPI) protocol, but also supporting the similar Synchronous Serial Interface (SSI) and Microwire communication protocols.

SPI is a three or four wire synchronous serial communication channel, capable of full-duplex operation. In three-wire configuration, there is a clock signal, and two data signals (one data signal from Master to Slave, the other from Slave to Master). In four-wire configuration, an additional Slave Select signal is output from Master and received by Slave.

The SPI clock signal is a gated clock generated from the Master and regulates the flow of data bits. The Master may transmit at a variety of baud rates, up to 24 MHz

In multi-Slave operation, no more than one Slave device can transmit data at any given time. Slave selection is accomplished when a Slave's "Slave Select" input is permanently grounded or asserted active-low by a Master device. Slave devices that are not selected do not interfere with SPI activities. Slave devices ignore the clock signals and keep their data output pins in



high-impedance state when not selected. The STR91xF supports SPI multi-Master operation because it provides collision detection.

Each SSP interface on the STR91xF has the following features:

- Full-duplex, three or four-wire synchronous transfers
- Master or Slave operation
- Programmable clock bit rate with prescaler, up to 24MHz for Master mode and 4MHz for Slave mode
- Separate transmit and receive FIFOs, each 16-bits wide and 8 locations deep
- Programmable data frame size from 4 to 16 bits
- Programmable clock and phase polarity
- Specifically for Microwire protocol:
  - Half-duplex transfers using 8-bit control message
- Specifically for SSI protocol:
  - Full-duplex four-wire synchronous transfer
  - Transmit data pin tri-stateable when not transmitting

#### 2.22.1 DMA

A programmable DMA channel may be assigned by CPU firmware to service each SSP channel for fast and direct transfers between the SSP bus and SRAM with little CPU involvement. Both DMA single-transfers and DMA burst-transfers are supported for transmit and receive. Burst transfers require that FIFOs are enabled.

# 2.23 General purpose I/O

There are up to 80 GPIO pins available on 10 I/O ports for 128-pin devices, and up to 40 GPIO pins on 5 I/O ports for 80-pin devices. Each and every GPIO pin by default (during and just after a reset condition) is in high-impedance input mode, and some GPIO pins are additionally routed to certain peripheral function inputs. CPU firmware may initialize GPIO pins to have alternate input or output functions as listed in *Table 3*. At any time, the logic state of any GPIO pin may be read by firmware as a GPIO input, regardless of its reassigned input or output function.

Bit masking is available on each port, meaning firmware may selectively read or write individual port pins, without disturbing other pins on the same port during a write.

Firmware may designate each GPIO pin to have open-drain or push-pull characteristics.

All GPIO pins are 5V tolerant, meaning in they can drive a voltage level up to  $V_{DDQ}$ , and can be safely driven by a voltage up to 5.5V.

There are no internal pull-up or pull-down resistors on GPIO pins. As such, it is recommended to ground, or pull up to  $V_{DDQ}$  with a 100K $\Omega$  resistor, all unused GPIO pins to minimize power consumption and noise generation.



# 2.24 A/D converter (ADC)

The STR91xF provides an eight-channel, 10-bit successive approximation analog-to-digital converter. The ADC input pins are multiplexed with other functions on Port 4 as shown in *Table 3*. Following are the major ADC features:

- Fast conversion time, as low as 0.7 usec
- Accuracy. Integral and differential non-linearity are typically within 4 conversion counts.
- 0 to 3.6V input range. External reference voltage input pin (AVREF) available on 128-pin packages for better accuracy on low-voltage inputs. The voltage on AVREF can range from 1.0V to V<sub>DDQ</sub>.
- CPU Firmware may convert one ADC input channel at a time, or it has the option to set the ADC to automatically scan and convert all eight ADC input channels sequentially before signalling an end-of-conversion
- Automatic continuous conversion mode is available for any number of designated ADC input channels
- Analog watchdog mode provides automatic monitoring of any ADC input, comparing it against two programmable voltage threshold values. The ADC unit will set a flag or it will interrupt the CPU if the input voltage rises above the higher threshold, or drops below the lower threshold.
- The ADC unit goes to stand-by mode (very low-current consumption) after any reset event. CPU firmware may also command the ADC unit to stand-by mode at any time.

# 2.25 Standard timers (TIM) with DMA

The STR91xF has four independent, free-running 16-bit timer/counter modules designated TIM0, TIM1, TIM2, and TIM3. Each general purpose timer/counter can be configured by firmware for a variety of tasks including; pulse width and frequency measurement (input capture), generation of waveforms (output compare and PWM), event counting, delay timing, and up/down counting.

Each of the four timer units have the following features:

- 16-bit free running timer/counter
- Internal timer/counter clock source from a programmable 8-bit prescale of the CCU PCLK clock output
- Optional external timer/counter clock source from pin P2.4 shared by TIM0/TIM1, and pin P2.5 shared by TIM2/TIM3. Frequency of these external clocks must be at least 4 times less the frequency of the internal CCU PCLK clock output. The Master clock (f<sub>MSTR</sub>) with a 16-bit prescaler can also be selected as an external clock source
- Two dedicated 16-bit Input Capture registers for measuring up to two input signals. Input Capture has programmable selection of input signal edge detection
- Two dedicated 16-bit Output Compare registers for generation up to two output signals
- PWM output generation with 16-bit resolution of both pulse width and frequency
- One pulse generation in response to an external event
- A dedicated interrupt to the CPU with five interrupt flags



### 2.25.1 DMA

A programmable DMA channel may be assigned by CPU firmware to service each timer/ counter module TIM0 and TIM1 for fast and direct transfers.

# 2.26 Three-phase induction motor controller (IMC)

The STR91xF provides an integrated controller for variable speed motor control applications.

Six PWM outputs are generated on high current drive pins P6.0 to P6.5 for controlling a threephase AC induction motor drive circuit assembly. Rotor speed feedback is provided by capturing a tachometer input signal on pin P6.6, and an asynchronous hardware emergency stop input is available on pin P6.7 to stop the motor immediately if needed, independently of firmware.

The IMC unit has the following features:

- Three PWM outputs generated using a 10-bit PWM counter, one for each phase U, V, W. Complimentary PWM outputs are also generated for each phase.
- Choice of classic or zero-centered PWM generation modes
- 10-bit PWM counter clock is supplied through a programmable 8-bit prescaler of the APB clock.
- Programmable 6-bit dead-time generator to add delay to each of the three complimentary PWM outputs
- 8-bit repetition counter
- Automatic rotor speed measurement with 16-bit resolution. Schmitt trigger tachometer input with programmable edge detection
- Hardware asynchronous emergency stop input
- A dedicated interrupt to CPU with eight flags

# 2.27 External memory interface (EMI)

STR91xF devices in 128-pin packages offer an external memory bus for connecting external parallel peripherals and memories. The EMI bus resides on ports 7, 8, and 9 and operates with either an 8 or 16-bit data path. The configuration of 8 or 16 bit mode is specified by CPU firmware writing to configuration registers at run-time. If the application does not use the EMI bus, then these port pins may be used for general purpose I/O as shown in *Table 3*.

The EMI has the following features:

- Supports static asynchronous memory access cycles, including page mode for non-mux operation.
- Four configurable memory regions, each with a chip select output (EMI\_CS0n ... EMI\_CS3n)
- Programmable wait states per memory region for both write and read operations
- **16-bit multiplexed data mode** (*Figure 4*): 16 bits of data and 16 bits of low-order address are multiplexed together on ports 8 and 9, while port 7 contains eight more high-order address signals. The output signal on pin EMI\_ALE is used to demultiplex the signals on ports 8 and 9, and the polarity of EMI\_ALE is programmable. The output signals on pins EMI\_BWR\_WRLn and EMI\_WRHn are the write strobes for the low and high data bytes



respectively. The output signal EMI\_RDn is the read strobe for both the low and high data bytes.

8-bit multiplexed data mode: This is a variant of the 16-bit multiplexed mode. Although this mode can provide 24 bits of address and 8 bits of data, it does require an external latch device on Port 8. However, this mode is most efficient when connecting devices that only require 8 bits of address on an 8-bit multiplexed address/data bus, and have simple read, write, and latch inputs as shown in *Figure 5* 

To use all 24 address bits, the following applies: 8 bits of lowest-order data and 8 bits of lowest-order address are multiplexed on port 8. On port 9, 8-bits of mid-order address are multiplexed with 8 bits of data, but these 8 data values are always at logic zero on this port during a write operation, and these 8 data bits are ignored during a read operation. An external latch device (such as a '373 latch) is needed to de-multiplex the mid-order 8 address bits that are generated on port 8. Port 7 outputs the 8 highest-order address signals directly (not multiplexed). The output signal on pin EMI\_ALE is used to demultiplex the signal on pin EMI\_BWR\_WRLn is the data write strobe, and the output on pin EMI\_RDn is the data read strobe.

• 8-bit non-multiplexed data mode (*Figure 6*): Eight bits of data are on port 8, while 16 bits of address are output on ports 7 and 9. The output signal on pin EMI\_BWR\_BWLn is the data write strobe and the output on pin EMI\_RDn is the data read strobe.

	_					-	
	STF	891xx EMI_CS3n			16-BIT		
		EMI_CS2n			DEVICE		
		EMI_CS1n					
		EMI_CS0n					
		EIVII_CSUN		-	CHIP_SELECT		
		EMI_WRHn			WRITE_HIGH_BYTE		
		EMI_BWR_WRLn			- WRITE_LOW_BYTE		
		EMI_RDn			- READ		
		EMI_ALE			ADDR_LATCH		
			D7 7				
		EMI_A23	P7.7 P7.6		A23		
		EMI_A22			A22		
		EMI_A21	P7.5		A21	1	
		EMI_A20	P7.4		A20		
		EMI_A19	P7.3		A19		
		EMI_A18	P7.2		A18		
		EMI_A17	P7.1		A17		
		EMI_A16	P7.0		A16		
		EIMI_AIO			A10		
105018		EMI_AD15	P9.7		AD15		
	$\circ$	EMI_AD13	P9.6		AD13 AD14		
			P9.5				
		EMI_AD13	P9.4		AD13		
		EMI_AD12	P9.3		AD12		
		EMI_AD11	P9.2		AD11		
		EMI_AD10	P9.1		AD10		
		EMI_AD9	P9.0		AD9		
		EMI_AD8			AD8		
			P8.7				
		EMI_AD7	P8.6		AD7		
		EMI_AD6	P8.5	<b></b>	AD6		
		EMI_AD5	P8.4		AD5		
		EMI_AD4	P8.3		AD4		
		EMI_AD3	P8.2		AD3		
		EMI_AD2	P8.2		AD2		
		EMI_AD1			AD1		
		EMI_AD0	P8.0		AD0		
						J	

#### Figure 4. EMI 16-bit multiplexed connection example

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EM	L_CS3n	8-BIT DEVICE ————————————————————————————————————
	I_WRLn ► /I_RDn ► /I_ALE ►	
EN EN EN EN EN EN	P8.7           P8.6           P8.5           P8.5           P8.5           P8.4           P8.3           P8.3           P8.4           P8.3           P8.4           P8.3           P8.4           P8.3           P8.4           P8.2           P8.1           P8.1           P8.0	AD7       AD6       AD5       AD4       AD3       AD1       AD0



STR91xx EMI_CS3n		8-BIT
EMI_CS3n EMI_CS2n		DEVICE
EMI_CS1n		
EMI_CS0n		CHIP_SELECT
		CHIF_SELECT
EMI_BWR_WRLn	► C	WRITE
EMI_RDn	► C	READ
	P9.7	
EMI_A15	P9.6	A15
EMI_A14	P9.5	A14
EMI_A13	P9 /	A13
EMI_A12	P9.3	A12
EMI_A11	P0.2	A11
EMI_A10	P9 1	A10
EMI_A9	P9.0	A9
EMI_A8		A8
	P7.7	A7
EMI_A7 EMI_A6		A7 A6
EMI_AG	P7.5	A0 A5
EMI_A3	P7.4	A4
EMI_A3	P7.3	A3
EMI_A2	P7.2	A2
EMI_A1	P7.1	A1
EMI_A0	P7.0	AO
	P8.7	
EMI_D7	P8.6	D7
EMI_D6	Dº 6	D6
EMI_D5	D9 /	D5
EMI_D4	Do 2	D4
EMI_D3	P9 2	D3
EMI_D2	P8.1	D2 D1
EMI_D1 EMI_D0	P8.0	D1
EMI_DO		80
XV		
CU'		
Obsolete		

# 3 Related documentation

#### Available from www.arm.com:

ARM966E-S Rev 2 Technical Reference Manual

Available from www.st.com:

STR91xF Reference Manual

STR9 Flash Programming Manual (PM0020)

The above is a selected list only, a full list STR91xF application notes can be viewed at *http://www.st.com*.

obsolete Product(s) - Obsolete Product(s)

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# 4 Pin description





ns



Figure 8. STR91xFW 128-pin package pinout



# 4.1 Default pin functions

During and just after reset, all pins on ports 0-9 default to high-impedance input mode until CPU firmware assigns other functions to the pins. This initial input mode routes all pins on ports 0-9 to be read as GPIO inputs as shown in the "Default Pin Function" column of *Table 3*. Simultaneously, certain port pin signals are also routed to other functional inputs as shown in the "Default Input Function" column of *Table 3*, and these pin input functions will remain until CPU firmware makes other assignments. At any time, even after the CPU assigns pins to alternate functions, the CPU may always read the state of any pin on ports 0-9 as a GPIO input. CPU firmware may assign alternate functions to port pins as shown in columns "Alternate Input 1" or "Alternate Output 1, 2, 3" of *Table 3* by writing to control registers at run-time.

#### Notes for Table 3:

- Notes: 1 STMicroelectronics advises to ground, or pull up to  $V_{DDQ}$  using a 100 K $\Omega$  resistor, all unused pins on port 0 9 to reduce noise susceptibility, noise generation, and minimize power consumption. There are no internal or programmable pull-up resistors on ports 0-9.
  - 2 All pins on ports 0 9 are 5V tolerant
  - 3 Pins on ports 0,1,2,4,5,7,8,9 have 4 mA drive and 4mA sink. Ports 3 and 6 have 8 mA drive and 8 mA sink.
  - 4 For 8-bit non-muxed EMI operation: Port 8 is eight bits of data, ports 7 and 9 are 16 bits of address.
  - 5 For 16-bit muxed EMI operation: Ports 8 and 9 are 16 bits of muxed address and data bits, port 7 is up to eight additional bits of high-order address
  - 6 Signal polarity is programmable for interrupt request inputs, EMI\_ALE, timer input capture inputs and output compare/PWM outputs, motor control tach and emergency stop inputs, and motor control phase outputs.
  - 7 HiZ = High Impedance, V = Voltage Source, G = Ground, I/O = Input/Output
  - 8 STR910F devices do not support USB. On these devices USBDP and USBDN signals are "Not Used" (USBDN is not connected, USBDP must be pulled up by a 1.5K ohm resistor to VDDQ), and all functions named "USB" are not available.
  - 9 STR910F 128-pin devices do not support Ethernet. On these devices PHYCLK and all functions named "MII\*" are not available.

	P	kg		Type			Alternate functions			
	LQFP80	LQFP128	Pin Name	Signal Ty	Default Pin Function	Default Input Function	Alternate Input 1	Alternate Output 1	Alternate Output 2	Alternate Output 3
	5	67	P0.0	I/O	GPIO_0.0, GP Input, HiZ	MII_TX_CLK, PHY Xmit clock	I2C0_CLKIN, I2C clock in	GPIO_0.0, GP Output	I2C0_CLKOUT, I2C clock out	ETM_PCK0, <i>ETM</i> Packet
U	-	69	P0.1	I/O	GPIO_0.1, GP Input, HiZ	-	I2C0_DIN, I <i>2C data in</i>	GPIO_0.1, GP Output	I2C0_DOUT, I2C data out	ETM_PCK1, ETM Packet
	-	71	P0.2	I/O	GPIO_0.2, GP Input, HiZ	MII_RXD0, PHY Rx data0	I2C1_CLKIN, I2C clock in	GPIO_0.2, GP Output	I2C1_CLKOUT, I2C clock out	ETM_PCK2, ETM Packet
	-	76	P0.3	I/O	GPIO_0.3, GP Input, HiZ	MII_RXD1, PHY Rx data	I2C1_DIN, I2C data in	GPIO_0.3, GP Output	I2C1_DOUT, I2C data out	ETM_PCK3, ETM Packet
	-	78	P0.4	I/O	GPIO_0.4, GP Input, HiZ	MII_RXD2, PHY Rx data	TIM0_ICAP1, Input Capture	GPIO_0.4, GP Output	EMI_CS0n, EMI Chip Select	ETM_PSTAT0, ETM pipe status
	-	85	P0.5	I/O	GPIO_0.5, GP Input, HiZ	MII_RXD3, PHY Rx data	TIM0_ICAP2, Input Capture	GPIO_0.5, GP Output	EMI_CS1n, EMI Chip Select	ETM_PSTAT1, ETM pipe status

Table 3. Device pin description


Ρ	'nkg		ре				Alternate	e functions	
LQFP80	LQFP128	Pin Name	Signal Type	Default Pin Function	Default Input Function	Alternate Input 1	Alternate Output 1	Alternate Output 2	Alternate Output 3
_	88	P0.6	I/O	GPIO_0.6,	MII_RX_CLK,	TIM2_ICAP1,	GPIO_0.6,	EMI_CS2n,	ETM_PSTAT2
-	00	1 0.0	1/0	GP Input, HiZ	PHY Rx clock	Input Capture	GP Output	EMI Chip Select	ETM pipe statu
-	90	P0.7	I/O	GPIO_0.7,	MII_RX_DV,	TIM2_ICAP2,	GPIO_0.7,	EMI_CS3n,	ETM_TRSYNC
			., 0	GP Input, HiZ	PHY data valid	Input Capture	GP Output	EMI Chip Select	ETM trace syn
	-								0001 0014
-	98	P1.0	I/O	GPIO_1.0,	MII_RX_ER,	ETM_EXTRIG,	GPIO_1.0,	UART1_TX, UART xmit data	SSP1_SCLK
				GP Input, HiZ	PHY rcv error	ETM ext. trigger	GP Output		SSP mstr clk o
-	99	P1.1	I/O	GPIO_1.1, GP Input, HiZ	-	UART1_RX, UART rcv data	GPIO_1.1, GP Output	MII_TXD0, MAC Tx data	SSP1_MOSI, SSP mstr dat o
							•		
-	101	P1.2	I/O	GPIO_1.2, GP Input, HiZ	-	SSP1_MISO, SSP mstr data in	GPIO_1.2, GP Output	MII_TXD1, MAC Tx data	UART0_TX, UART xmit dat
				GPIO 1.3,		UART2_RX,	GPIO 1.3,	MII_TXD2,	SSP1 NSS,
-	106	P1.3	I/O	GP Input, HiZ	-	UART rcv data	GP Output	MAC Tx data	SSP mstr sel o
				GPIO_1.4,		12C0_CLKIN, <i>12C</i>	GPIO_1.4,	MII_TXD3,	I2C0_CLKOU
-	109	P1.4	I/O	GP Input, HiZ	-	clock in	GP Output	MAC Tx data	12C0_CLKOU
				GPIO_1.5,	MII COL.	CAN_RX,	GPIO_1.5,	UART2_TX,	ETM_TRCLK
-	110	P1.5	I/O	GP Input, HiZ	PHY collision	CAN rcv data	GP Output	UART xmit data	ETM trace close
				GPIO_1.6,	MIL_CRS,	I2C0_DIN,	GPIO_1.6,	CAN TX,	I2C0_DOUT
-	114	P1.6	I/O	GP Input, HiZ	PHY carrier sns	12C data in	GP Output	CAN Tx data	I2C data out
				GPIO_1.7,		ETM_EXTRIG,	GPIO_1.7,	MII_MDC,	ETM TRCLK
-	116	P1.7	I/O	GP Input, HiZ	-	ETM ext. trigger	GP Output	MAC mgt dat ck	ETM trace clo
				GPIO_2.0,	UART0_CTS,	I2C0_CLKIN, 12C	GPIO_2.0,	12C0_CLKOUT, 12C	ETM_PCK0, E
7	10	P2.0	I/O	GP Input, HiZ	Clear To Send	clock in	GP Output	clock out	Packet
				GPIO_2.1,	UART0_DSR,	I2C0_DIN,	GPIO_2.1,	I2C0_DOUT,	ETM_PCK1, E
8	11	P2.1	I/O	GP Input, HiZ	Data Set Ready	1200_DIN,	GP Output	I2C data out	Packet
				GPIO 2.2,	UARTO DCD,	12C1 CLKIN, 12C	GPIO 2.2,	I2C1_CLKOUT, I2C	ETM_PCK2, E
21	33	P2.2	I/O	GP Input, HiZ	Dat Carrier Det	clock in	GP Output	clock out	Packet
				GPIO_2.3,	UART0_RI,	I2C1_DIN,	GPIO_2.3,	I2C1_DOUT,	ETM_PCK3, E
22	35	P2.3	I/O	GP Input, HiZ	Ring Indicator	I2C data in	GP Output	I2C data out	Packet
				GPIO_2.4,	EXTCLK TOT1Ext	SSP0_SCLK,	GPIO 2.4,	SSP0 SCLK,	ETM PSTATO
23	37	P2.4	I/O	GP Input, HiZ	clk timer0/1	SSP slv clk in	GP Output	SSP mstr clk out	ETM pipe stat
				GPIO_2.5,	EXTCLK T2T3Ext	SSP0_MOSI,	GPIO_2.5,	SSP0 MOSI,	ETM_PSTAT
29	45	P2.5	I/O	GP Input, HiZ	clk timer2/3	SSP slv dat in	GP Output	SSP mstr dat out	ETM pipe stat
				GPIO_2.6,		SSP0 MISO,	GPIO 2.6,	SSP0 MISO,	ETM PSTAT
32	53	P2.6	I/O	GP Input, HiZ	-	SSP mstr data in	GP Output	SSP slv data out	ETM pipe stat
~~		USBCLK		GPIO_2.7,	USB_CLK48M,	SSP0_NSS,	GPIO_2.7,	SSP0_NSS,	ETM_TRSYN
33	54	_P2.7	1/0	GP Input, HiZ	48MHz to USB	SSP slv sel in	GP Output	SSP mstr sel out	ETM trace syr
				GPIO_3.0,	DMA_RQST0,	UART0_RxD,	GPIO_3.0,	UART2_TX,	TIM0 OCMP1,
34	55	P3.0	1/0	GP Input, HiZ	Ext DMA requst	UART rcv data	GP Output	UART xmit data	comp/PWM
				GPIO 3.1,	DMA RQST1.	UART2_RxD,	GPIO_3.1,	UART0_TX,	TIM1_OCMP1,
37	59	P3.1	I/O	GP Input, HiZ	Ext DMA requst	UART rcv data	GP Output	UART xmit data	comp/PWM
	2			GPIO_3.2,	EXINT2,	UART1_RxD,	GPIO_3.2,	CAN_TX,	UART0_DTR
38	60	P3.2	I/O	GP Input, HiZ	External Intr	UART rcv data	GP Output	CAN Tx data	Data Trmnl Ro
		D		GPIO_3.3,	EXINT3,	CAN_RX,	GPIO_3.3,	UART1_TX,	UART0_RTS
39	61	P3.3	I/O	GP Input, HiZ	External Intr	CAN rcv data	GP Output	UART xmit data	Ready To Sen
				GPIO_3.4,	EXINT4,	SSP1_SCLK,	GPIO_3.4,	SSP1_SCLK,	UART0_TX,
40	63	P3.4	I/O	GP Input, HiZ	External Intr	SSP slv clk in	GP Output	SSP mstr clk out	UART xmit da
		<b>F a c</b>	110	GPIO_3.5,	EXINT5,	SSP1_MISO,	GPIO_3.5,	SSP1_MISO,	UART2_TX,
41	65	P3.5	I/O	GP Input, HiZ	External Intr	SSP mstr data in	GP Output	SSP slv data out	UART xmit da
		P3.6	I/O	GPIO_3.6,	EXINT6,	SSP1_MOSI,	GPIO_3.6,	SSP1_MOSI,	CAN_TX,
42	66								

Р	kg		Type				Alternate	e functions	
LQFP80	LQFP128	Pin Name	Signal Ty	Default Pin Function	Default Input Function	Alternate Input 1	Alternate Output 1	Alternate Output 2	Alternate Output 3
43	68	P3.7	I/O	GPIO_3.7, GP Input, HiZ	EXINT7, External Intr	SSP1_NSS, SSP slv select in	GPIO_3.7, GP Output	SSP1_NSS, SSP mstr sel out	TIM1_OCMP1, Out comp/PWM
4	3	P4.0	I/O	GPIO_4.0, GP Input, HiZ	ADC0, ADC input chnl	TIM0_ICAP1, Input Capture	GPIO_4.0, GP Output	TIM0_OCMP1, Out comp/PWM	ETM_PCK0, ETM Packet
3	2	P4.1	I/O	GPIO_4.1, GP Input, HiZ	ADC1, ADC input chnl	TIM0_ICAP2, Input Capture	GPIO_4.1, GP Output	TIM0_OCMP2, Out comp	ETM_PCK1, ETM Packet
2	1	P4.2	I/O	GPIO_4.2, GP Input, HiZ	ADC2, ADC input chnl	TIM1_ICAP1, Input Capture	GPIO_4.2, GP Output	TIM1_OCMP1, Out comp/PWM	ETM_PCK2, ETM Packet
1	128	P4.3	I/O	GPIO_4.3, GP Input, HiZ	ADC3, ADC input chnl	TIM1_ICAP2, Input Capture	GPIO_4.3, GP Output	TIM1_OCMP2, Out comp	ETM_PCK3, ETM Packet
80	127	P4.4	I/O	GPIO_4.4, GP Input, HiZ	ADC4, ADC input chnl	TIM2_ICAP1, Input Capture	GPIO_4.4, GP Output	TIM2_OCMP1, Out comp/PWM	ETM_PSTAT0, ETM pipe status
79	126	P4.5	I/O	GPIO_4.5, GP Input, HiZ	ADC5, ADC input chnl	TIM2_ICAP2, Input Capture	GPIO_4.5, GP Output	TIM2_OCMP2, Out comp	ETM_PSTAT1, ETM pipe status
78	125	P4.6	I/O	GPIO_4.6, GP Input, HiZ	ADC6, ADC input chnl	TIM3_ICAP1, Input Capture	GPIO_4.6, GP Output	TIM3_OCMP1, Out comp/PWM	ETM_PSTAT2, ETM pipe status
77	124	P4.7	I/O	GPIO_4.7, GP Input, HiZ	ADC7, ADC input chnl	TIM3_ICAP2, Input Capture	GPIO_4.7, GP Output	TIM3_OCMP2, Out comp	ETM_TRSYNC, ETM trace sync
9	12	P5.0	I/0	GPIO_5.0, GP Input, HiZ	EXINT8, External Intr	CAN_RX, CAN rcv data	GPIO_5.0, GP Output	ETM_TRCLK, ETM trace clock	UART0_TX, UART xmit data
12	18	P5.1	I/0	GPIO_5.1, GP Input, HiZ	EXINT9, External Intr	UART0_RxD, UART rcv data	GPIO_5.1, GP Output	CAN_TX, CAN Tx data	UART2_TX, UART xmit data
17	25	PHYCLK _P5.2	I/O	GPIO_5.2, GP Input, HiZ	EXINT10, External Intr	UART2_RxD, UART rcv data	GPIO_5.2, GP Output	MII_PHYCLK, 25Mhz to PHY	TIM3_OCMP1, Out comp/PWM
18	27	P5.3	I/O	GPIO_5.3, GP Input, HiZ	EXINT11, External Intr	ETM_EXTRIG, ETM ext. trigger	GPIO_5.3, GP Output	MII_TX_EN, MAC xmit enbl	TIM2_OCMP1, Out comp/PWM
44	70	P5.4	I/O	GPIO_5.4, GP Input, HiZ	EXINT12, External Intr	SSP0_SCLK, SSP slv clk in	GPIO_5.4, GP Output	SSP0_SCLK, SSP mstr clk out	EMI_CS0n, EMI Chip Select
47	77	P5.5	I/O	GPIO_5.5, GP Input, HiZ	EXINT13, External Intr	SSP0_MOSI, SSP slv dat in	GPIO_5.5, GP Output	SSP0_MOSI, SSP mstr dat out	EMI_CS1n, EMI Chip Select
48	79	P5.6	I/O	GPIO_5.6, GP Input, HiZ	EXINT14, External Intr	SSP0_MISO, SSP mstr dat in	GPIO_5.6, GP Output	SSP0_MISO, SSP slv data out	EMI_CS2n, EMI Chip Select
49	80	P5.7	I/O	GPIO_5.7, GP Input, HiZ	EXINT15, External Intr	SSP0_NSS, SSP slv select in	GPIO_5.7, GP Output	SSP0_NSS, SSP mstr sel out	EMI_CS3n, EMI Chip Select
19	29	P6.0	1/0	GPIO_6.0, GP Input, HiZ	EXINT16, External Intr	TIM0_ICAP1, Input Capture	GPIO_6.0, GP Output	TIM0_OCMP1, Out comp/PWM	MC_UH, IMC phase U hi
20	31	P6.1	1/0	GPIO_6.1, GP Input, HiZ	EXINT17, External Intr	TIM0_ICAP2, Input Capture	GPIO_6.1, GP Output	TIM0_OCMP2, Out comp	MC_UL, IMC phase U lo
13	19	P6.2	I/O	GPIO_6.2, GP Input, HiZ	EXINT18, External Intr	TIM1_ICAP1, Input Capture	GPIO_6.2, GP Output	TIM1_OCMP1, Out comp/PWM	MC_VH, IMC phase V hi
14	20	P6.3	I/O	GPIO_6.3, GP Input, HiZ	EXINT19, External Intr	TIM1_ICAP2, Input Capture	GPIO_6.3, GP Output	TIM1_OCMP2, Out comp	MC_VL, IMC phase V lo
52	83	P6.4	I/O	GPIO_6.4, GP Input, HiZ	EXINT20, External Intr	TIM2_ICAP1, Input Capture	GPIO_6.4, GP Output	TIM2_OCMP1, Out comp/PWM	MC_WH, IMC phase W hi
53	84	P6.5	I/O	GPIO_6.5, GP Input, HiZ	EXINT21, External Intr	TIM2_ICAP2, Input Capture	GPIO_6.5, GP Output	TIM2_OCMP2, Out comp	MC_WL, IMC phase W lo
57	92	P6.6	I/O	GPIO_6.6, GP Input, HiZ	EXINT22_TRIG, Ext Intr & Tach	UART0_RxD, UART rcv data	GPIO_6.6, GP Output	TIM3_OCMP1, Out comp/PWM	ETM_TRCLK, ETM trace clock
58	93	P6.7	I/O	GPIO_6.7, GP Input, HiZ	EXINT23_STOP, Ext Intr & Estop	ETM_EXTRIG, ETM ext. trigger	GPIO_6.7, GP Output	TIM3_OCMP2, Out comp	UART0_TX, UART xmit data



Ρ	kg		be				Alternate	e functions	
LQFP80	LQFP128	Pin Name	Signal Type	Default Pin Function	Default Input Function	Alternate Input 1	Alternate Output 1	Alternate Output 2	Alternate Output 3
-	5	P7.0	I/O	GPIO_7.0,	EXINT24,	TIM0_ICAP1,	GPIO_7.0,	8b) EMI_A0,	ETM_PCK0, ET
	Ŭ	17.0	1/0	GP Input, HiZ	External Intr	Input Capture	GP Output	16b) EMI_A16	Packet
-	6	P7.1	I/O	GPIO_7.1,	EXINT25,	TIM0_ICAP2,	GPIO_7.1,	8b) EMI_A1,	ETM_PCK1, ET
				GP Input, HiZ	External Intr	Input Capture	GP Output	16b) EMI_A17	Packet
-	7	P7.2	I/O	GPIO_7.2,	EXINT26,	TIM2_ICAP1,	GPIO_7.2,	8b) EMI_A2,	ETM_PCK2, E1 Packet
				GP Input, HiZ GPIO 7.3,	External Intr	Input Capture TIM2 ICAP2,	GP Output GPIO_7.3,	16b) EMI_A18	
-	13	P7.3	I/O	GPI0_7.3, GP Input, HiZ	EXINT27, External Intr	Input Capture	GPIO_7.3, GP Output	8b) EMI_A3, 16b) EMI_A19	ETM_PCK3, ET Packet
				GPIO_7.4,	EXINT28,	UARTO_RxD,	GPIO_7.4,	8b) EMI_A4,	EMI_CS3n,
-	14	P7.4	I/O	GP Input, HiZ	External Intr	UART rcv data	GP Output	16b) EMI_A20	EMI Chip Sele
				GPIO 7.5,	EXINT29,	ETM_EXTRIG,	GPIO 7.5,	8b) EMI_A5,	EMI_CS2n,
-	15	P7.5	I/O	GP Input, HiZ	External Intr	ETM ext. trigger	GP Output	16b) EMI_A21	EMI Chip Sele
				GPIO_7.6,	EXINT30,	TIM3_ICAP1,	GPIO_7.6,	8b) EMI_A6,	EMI_CS1n,
-	118	P7.6	I/O	GP Input, HiZ	External Intr	Input Capture	GP Output	16b) EMI A22	EMI Chip Sele
		_		GPIO_7.7,	EXINT31,	TIM3_ICAP2,	GPIO_7.7,	EMI_CS0n,	16b) EMI_A23
-	119	P7.7	I/O	GP Input, HiZ	External Intr	Input Capture	GP Output	EMI chip select	8b) EMI_A7
	1								X
	00	D0.0	I/O	GPIO_8.0,			GPIO_8.0,	8b) EMI_D0,	
-	26	P8.0	1/0	GP Input, HiZ	-	-	GP Output	16b) EMI_AD0	<u> </u>
	28	P8.1	I/O	GPIO_8.1,			GPIO_8.1,	8b) EMI_D1,	
-	20	F0.1	1/0	GP Input, HiZ	-	-	GP Output	16b) EMI_AD1	-
-	30	P8.2	1/0	GPIO_8.2,		_	GPIO_8.2,	8b) EMI_D2,	_
	00	1 0.2	1/0	GP Input, HiZ			GP Output	16b) EMI_AD2	
-	32	P8.3	I/O	GPIO_8.3,	-	-	GPIO_8.3,	8b) EMI_D3,	-
				GP Input, HiZ			GP Output	16b) EMI_AD3	
-	34	P8.4	I/O	GPIO_8.4,	-	-	GPIO_8.4,	8b) EMI_D4,	-
				GP Input, HiZ			GP Output	16b) EMI_AD4	
-	36	P8.5	I/O	GPIO_8.5,	-	$\sim$	GPIO_8.5,	8b) EMI_D5,	-
				GP Input, HiZ			GP Output	16b) EMI_AD5	
-	38	P8.6	I/O	GPIO_8.6, GP Input, HiZ	-	-	GPIO_8.6, GP Output	8b) EMI_D6, 16b) EMI_AD6	-
				GPIIO 8.7,			GPIO_8.7,	8b) EMI_D7,	
-	44	P8.7	I/O	GP Input, HiZ		-	GP Output	16b) EMI_AD7	-
							Gi Output	100) EMI_AD7	
	1			GPIO_9.0,	0		GPIO_9.0,	8b) EMI_A8	1
-	46	P9.0	I/O	GP Input, HiZ	-	-	GP Output	16b) EMI_AD8	-
				GPIO_9.1,			GPIO_9.1,	8b) EMI_A9,	
-	47	P9.1	I/O	GP Input, HiZ	-	-	GP Output	16b) EMI_AD9	-
				GPIO_9.2,			GPIO_9.2,	8b) EMI_A10,	
-	50	P9.2	I/O	GP Input, HiZ	-	-	GP Output	16b)EMI_AD10	-
	- 1		1/0	GPIO_9.3,			GPIO_9.3,	8b) EMI_A11,	
-	51	P9.3	I/O	GP Input, HiZ	-	-	GP Output	16b)EMI_AD11	-
-			1/0	GPIO_9.4,			GPIO_9.4,	8b) EMI_A12,	
$\frown$	52	P9.4	I/O	GP Input, HiZ	-	-	GP Output	16b)EMI_AD12	-
Y	E0		I/O	GPIO_9.5,			GPIO_9.5,	8b) EMI_A13,	
-	58	P9.5	1/0	GP Input, HiZ	-	-	GP Output	16b)EMI_AD13	-
_	62	P9.6	I/O	GPIO_9.6,	_	-	GPIO_9.6,	8b) EMI_A14,	_
_	02	F 9.0	1/0	GP Input, HiZ	-	-	GP Output	16b)EMI_AD14	
-	64	P9.7	I/O	GPIO_9.7,			GPIO_9.7,	8b) EMI_A15,	-
	04	1 3.7		GP Input, HiZ		-	GP Output	16b)EMI_AD15	-

Ρ	kg		be				Alternate	functions	
LQFP80	LQFP128	Pin Name	Signal Type	Default Pin Function	Default Input Function	Alternate Input 1	Alternate Output 1	Alternate Output 2	Alternate Output 3
-	21	EMI_BWR _WRLn (used as EMI_LBn in future rev.)	0	EMI byte write strobe (8 bit mode) or low byte write strobe (16 bit mode)			N/A		
-	22	EMI_WRHn (used as EMI_UBn in future rev.)	0	EMI high byte write strobe (16- bit mode)			N/A		
-	74	EMI_ALE	0	EMI address latch enable (mux mode)			N/A		
-	75	EMI_RDn	0	EMI read strobe			N/A		
-	-	EMI_BAAn	0	TBD			N/A		
-	-	EMI_WAITn	1	TBD			N/A		
-	-	EMI_BCLK	0	TBD Reserved for			N/A		-
-	-	EMI_WEn	0	future use			N/A		16)
-	91	TAMPER _IN	I	Tamper detection input			N/A		CIC .
-	94	MII_MDIO	I/O	MAC/PHY management data line			N/A	du	
59	95	USBDN	I/O	USB data (-) bus connect			N/A		
60	96	USBDP	I/O	USB data (+) bus connect			N/A		
56	89	RESET _INn RESET	Ι	External reset input Global or System			N/A		
62	100	_OUTn	0	reset output CPU oscillator or		~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~	N/A		
65	104	X1_CPU	1	crystal input CPU crystal		0	N/A		
64	103	X2_CPU	0	connection RTC oscillator or	.15		N/A		
27	42	X1_RTC	I	crystal input (32.768 kHz) RTC crystal	CIL		N/A		
26	41	X2_RTC	0	JTAG return	A _		N/A		
61	97	JRTCK	0	clock or RTC clock			N/A		
67	107	JTRSTn	I	JTAG TAP controller reset			N/A		
68	108	JTCK	I	JTAG clock			N/A		
69	111	JTMS	I	JTAG mode select			N/A		
72	115	JTDI	1	JTAG data in			N/A		
73	117	JTDO	0	JTAG data out			N/A		
-	122	AVDD	۷	voltage source, 2.7V - 3.6V			N/A		
-	4	AVSS	G	ADC analog ground			N/A		
5	-	AVSS _VSSQ	G	Common ground point for digital I/ O & analog ADC			N/A		



Pin Name AVREF AVREF AVDD VBATT VDDQ VDDQ VDDQ VDDQ	<pre>c &lt; Signal Type</pre>	ADC reference voltage input Combined ADC ref voltage and ADC analog voltage source, 2.7V - 3.6V Standby voltage input for RTC and SRAM backup	Default Input Function	Alternate Input 1	Alternate Output 1 N/A	Alternate Output 2	Alternat Output \$
AVREF _AVDD VBATT VDDQ VDDQ VDDQ	v v v	voltage input Combined ADC ref voltage and ADC analog voltage source, 2.7V - 3.6V Standby voltage input for RTC and SRAM					
_AVDD VBATT VDDQ VDDQ VDDQ	V	ref voltage and ADC analog voltage source, 2.7V - 3.6V Standby voltage input for RTC and SRAM			N/A		
VDDQ VDDQ VDDQ	V	input for RTC and SRAM					
VDDQ VDDQ					N/A		
VDDQ							
	V	ļ					
	۷						
VDDQ	۷	V Source for					
VDDQ	۷	I/O and USB.			N/A		
VDDQ	V	2.7V to 3.6V					16
VDDQ	V						*(2)
VDDQ	۷						
VDDQ	V					11	9
	G						
VSSQ	G					~ YOY	
VSSQ	G				<pre></pre>		
		Digital Ground					
		for			N/A		
		1/0 and 05b			16,		
		-		5			
				$\Delta 0^{\circ}$			
		V Source for		U.			
		CPU.			N/A		
		1.65V - 2.0V					
		[	,151				
		_	$\sim$				
		Digital Ground			N/A		
			· ·				
		V Source for PLL					
					N/A		
PLLVSSQ	G	for PLL					
	VDDQ VSSQ VSSQ VSSQ VSSQ VSSQ VSSQ VSSQ	VDDQVVSSQGVSSQGVSSQGVSSQGVSSQGVSSQGVSSQGVSSQGVSSQGVDDVVDDVVDDVVDDVVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSGVSSQVLVDDQV	VDDQ       V         VSSQ       G         VDD       V         VDD       V         VDD       V         VSS       G         VSS       G         VSS       G         VSS       G         PLLVDDQ       V         VSource for PLL 2.7 to 3.6 V	VDDQ       V         VSSQ       G         VDD       V         VDD       V         VDD       V         VSS       G         VSS       G         VSS       G         VSS       G         PLLVDDQ       V         VSource for PLL         2.7 to 3.6 V	VDD         V         1.65V - 2.0V           VDD         V         1.65V - 2.0V           VSS         G         Digital Ground for CPU           VSS         G         Digital Ground for CPU           VSS         G         VSS           PLLVDDQ         V         V Source for PLL 2.7 to 3.6 V	VDD         V         1.65V - 2.0V           VDD         V         1.65V - 2.0V           VSS         G         Digital Ground for CPU           VSS         G         Digital Ground for CPU           VSS         G         2.7 to 3.6 V           PLLVDDQ         V         VSource for PLL 2.7 to 3.6 V	VDDQ         V           VSSQ         G           VDD         V           VDD         V           VDD         V           VSS         G           VSS         G           VSS         G           VSS         G           VSS         G           PLLVDDQ         V           VSource for PLL           2.7 to 3.6 V



# 5 Memory mapping

The ARM966E-S CPU addresses a single linear address space of 4 giga-bytes (2<sup>32</sup>) from address 0x0000.0000 to 0xFFFF.FFFF as shown in *Figure 9*. Upon reset the CPU boots from address 0x0000.0000, which is chip-select zero at address zero in the Flash Memory Interface (FMI).

The Instruction TCM and Data TCM enable high-speed CPU operation without incurring any performance or power penalties associated with accessing the system buses (AHB and APB). I-TCM and D-TCM address ranges are shown at the bottom of the memory map in *Figure 9*.

# 5.1 Buffered and non-buffered writes

The CPU makes use of write buffers on the AHB and the D-TCM to decouple the CPU from any wait states associated with a write operation. The user may choose to use write with buffers on the AHB by setting bit 3 in control register CP15 and selecting the appropriate AHB address range when writing. By default at reset, buffered writes are disabled (bit 3 of CP15 is clear) and all AHB writes are non-buffered until enabled. *Figure 9* shows that most addressable items on the AHB are aliased at two address ranges, one for buffered writes and another for non-buffered writes. A buffered write will allow the CPU to continue program execution while the write-back is performed through a FIFO to the final destination on the AHB. If the FIFO is full, the CPU is stalled until FIFO space is available. A non-buffered write will impose an immediate delay to the CPU, but results in a direct write to the final AHB destination, ensuring data coherency. Read operations from AHB locations are always direct and never buffered.

# 5.2 System (AHB) and peripheral (APB) buses

The CPU will access SRAM, higher-speed peripherals (USB, Ethernet, Programmable DMA), and the external bus (EMI) on the AHB at their respective base addresses indicated in *Figure 9*. Lower-speed peripherals reside on the APB and are accessed using two separate AHB-to-APB bridge units (APB0 and APB1). These bridge units are essentially address windows connecting the AHB to the APB. To access an individual APB peripheral, the CPU will place an address on the AHB bus equal to the base address of the appropriate bridge unit APB0 or APB1, plus the offset of the particular peripheral, plus the offset of the individual data location within the peripheral. *Figure 9* shows the base addresses of bridge units APB0 and APB1, and also the base address of each APB peripheral. Please consult the STR91xF Reference manual for the address of data locations within each individual peripheral.



# SRAM

The SRAM is aliased at three separate address ranges as shown in *Figure 9*. When the CPU accesses SRAM starting at 0x0400.0000, the SRAM appears on the D-TCM. When CPU access starts at 0x4000.0000, SRAM appears in the buffered AHB range. Beginning at CPU address 0x5000.0000, SRAM is in non-buffered AHB range. The SRAM size must be specified by CPU initialization firmware writing to a control register after any reset condition. Default SRAM size is 32K bytes, with option to set to 64K bytes on STR91xFx32 devices, and to 96K bytes on STR91xFx44 devices.



When other AHB bus masters (such as a DMA controller) write to SRAM, their access is never buffered. Only the CPU can make use of buffered AHB writes.

# 5.4 Two independent Flash memories

The STR91xF has two independent Flash memories, the larger primary Flash and the small secondary Flash. It is possible for the CPU to erase/write to one of these Flash memories while simultaneously reading from the other.

One or the other of these two Flash memories may reside at the "boot" address position of 0x0000.0000 at power-up or at reset as shown in *Figure 9*. The default configuration is that the first sector of primary Flash memory is enabled and residing at the boot position, and the secondary Flash memory is disabled. This default condition may be optionally changed as described below.

### 5.4.1 Default configuration

When the primary Flash resides at boot position, typical CPU initialization firmware would set the start address and size of the main Flash memory, and go on to enable the secondary Flash, define it's start address and size. Most commonly, firmware would place the secondary Flash start address at the location just after the end of the primary Flash memory. In this case, the primary Flash is used for code storage, and the smaller secondary flash can be used for data storage (EEPROM emulation).

## 5.4.2 Optional configuration

Using the STR91xF device configuration software tool, or IDE from 3rd party, one can specify that the smaller secondary Flash memory is at the boot location at reset and the primary Flash is disabled. The selection of which Flash memory is at the boot location is programmed in a non-volatile Flash-based configuration bit during JTAG ISP. The boot selection choice will remain as the default until the bit is erased and re-written by the JTAG interface. The CPU cannot change this choice for boot Flash, only the JTAG interface has access.

In this case where the secondary Flash defaults to the boot location upon reset, CPU firmware would typically initialize the Flash memories the following way. The secondary Flash start address and size is specified, then the primary Flash is enabled and its start address and size is specified. The primary Flash start address would typically be located just after the final address location of the secondary Flash. This configuration is particularly well-suited for In-Application-Programming (IAP). The CPU would boot from the secondary Flash memory, initialize the system, then check the contents of the primary Flash memory (by checksum or other means). If the contents of primary Flash is OK, then CPU execution continues from either Flash memory. If the main Flash contents are incorrect, the CPU, while executing code from the secondary Flash, can download new data from any STR91xF communication channel and program into primary Flash memory. Application code then starts after the new contents of primary Flash are verified.



#### Notes for Figure 9: STR91xF memory map on page 45:

- Notes: 1 Either of the two Flash memories may be placed at CPU boot address 0x0000.0000. By default, the primary Flash memory is in boot position starting at CPU address 0x0000.0000 and the secondary Flash memory may be placed at a higher address following the end of the primary Flash memory. This default option may be changed using the STR91xx device configuration software, placing the secondary Flash memory at CPU boot location 0x0000.0000, and then the primary Flash memory may be placed at a higher address.
  - 2 The local SRAM (64KB or 96KB) is aliased in three address windows. A) At 0x0400.0000 the SRAM is accessible through the CPU's D-TCM, at 0x4000.0000 the SRAM is accessible through the CPU's AHB in buffered accesses, and at 0x5000.0000 the SRAM is accessible through the CPU's AHB in non-buffered accesses. An AHB bus master other than the CPU can access SRAM in all three aliased windows, but these accesses are always non-buffered. The CPU is the only AHB master that can performed buffered writes.
  - 3 APB peripherals reside in two AHB-to-APB peripheral bridge address windows, APB0 and APB1. These peripherals are accessible with buffered AHB access if the CPU addresses them in the address range of 0x4800.0000 to 0x4FFF.FFFF, and non-buffered access in the address range of 0x5800.0000 to 0x5FFF.FFFF.
- 4 Individual peripherals on the APB are accessed at the listed address offset plus the base address of the appropriate AHB-to-APB bridge.

	TOTAL 4 GB CPU		<u>APB BASE +</u> <u>OFFSET</u> APB1+0x03FF.FFFF	PERIPHERAL BUS MEMORY SPACE <sup>(4)</sup>	
	MEMORY SPACE			RESERVED	、 、
0xFFFF.FFFF			APB1+0x0000.E000	I2C1 4 КВ	
0xFFFF.F000	VIC0	4 КВ АНВ	APB1+0x0000.D000	12С0 4 КВ	
0xFC01.0000	RESERVED	> NON- ( BUFFERED	APB1+0x0000.C000	WATCHDOG 4 KB	
0xFC00.0000	VIC1	64 КВ 🜙	APB1+0x0000.B000	ADC 4 KB	
			APB1+0x0000.A000	CAN 4 KB	
			APB1+0x0000.9000	SSP1 4 KB	AP
	RESERVED		APB1+0x0000.8000	SSP0 4 KB	∖ AH
			APB1+0x0000.7000	UART2 4 KB	to-/
0x8000.0000			APB1+0x0000.6000	UART1 4 KB	
0x7C00.0000	ENET	64 MB	APB1+0x0000.5000	UARTO 4 KB	
0x7800.0000	8-CH DMA	64 MB AHB	APB1+0x0000.4000	IMC 4 KB	
0x7400.0000	EMI	64 MB BUFFERED	APB1+0x0000.3000	SCU 4 KB	
0x7000.0000	USB	64 MB	APB1+0x0000.2000	RTC 4 KB	
0x6C00.0000	ENET	64 MB	APB1+0x0000.1000		/
0x6800.0000	8-CH DMA	64 MB ( AHB	APB1+0x0000.0000	APB1 CONFIG 4 KB	<i>,</i>
0x6400.0000	EMI	64 MB BUFFERED	APB0+0x03FF.FFFF		
0x6000.0000	USB	64 MB	APB0+0x0001.0000	RESERVED	\
0x5C00.0000	APB1	64 MB PERIPHERAL BUS,	APB0+0x0000.F000	GPIO PORT P9 4 KB	
0x5800.0000	APB0	64 MB ( AHB ACCESS (3)	APB0+0x0000.E000	GPIO PORT P8 4 KB	
0x5400.0000	FMI	64 MB BUFFERED	APB0+0x0000.D000	GPIO PORT P7 4 KB	
0x5400.0000	SRAM, AHB (2)	64 MB	APB0+0x0000.C000	GPIO PORT P6 4 KB	
0x3000.0000	APB1	64 MB PERIPHERAL BUS,	APB0+0x0000.C000	GPIO PORT P5 4 KB	
0x4800.0000	APB0	64 MB BUFFERED ACCESS (3)	APB0+0x0000.A000	GPIO PORT P4 4 KB	
0x4800.0000	FMI	64 MB (BUFFERED	APB0+0x0000.9000	GPIO PORT P3 4 KB	AP
0x4400.0000	SRAM, AHB (2)	64 MB	APB0+0x0000.9000	GPIO PORT P2 4 KB	} A⊢ to-
	Ext. MEM, CS0	64 MB		GPIO PORT P1 4 KB	Bri
0x3C00.0000	Ext. MEM, CS1	64 MB AHB	APB0+0x0000.7000	GPIO PORT P0 4 KB	
0x3800.0000	Ext. MEM, CS2	64 MB BUFFERED	APB0+0x0000.6000	TIM3 4 кв	
0x3400.0000	Ext. MEM, CS3	64 MB	APB0+0x0000.5000	TIM2 4 KB	
0x3000.0000	Ext. MEM, CS0	64 MB	APB0+0x0000.4000	ТІМ1 4 КВ	
0x2C00.0000	Ext. MEM, CS1	64 MB	APB0+0x0000.3000	TIMO 4 KB	
0x2800.0000	Ext. MEM, CS2	64 MB BUFFERED	APB0+0x0000.2000	WAKE-UP UNIT 4 KB	
0x2400.0000	Ext. MEM, CS3	64 MB	APB0+0x0000.1000	APB0 CONFIG 4 KB	/
0x2000.0000	RESERVED	°	APB0+0x0000.0000 rder of the two Flash men (1) SECONDARY FLASH (BANK 1), 32KB		, ,
0x0800.0000			MAIN FLASH (BANK 0),	SECONDARY	
0x0400.0000 _	SRAM, D-TCM (2)	Using 64 KB or 96 KB	256KB or 512KB	FLASH (BANK 1),	
0x0000.0000	FLASH, I-TCM <sup>(1)</sup>	Using 288 KB or 544 KB 0x0000.0000		32KB	

#### Figure 9. STR91xF memory map



# 6 Electrical characteristics

# 6.1 Absolute maximum ratings

This product contains devices to protect the inputs against damage due to high static voltages. However, it is advisable to take normal precautions to avoid application of any voltage higher than the specified maximum rated voltages. It is also recommended to ground any unused input pin to reduce power consumption and minimize noise.

Symbol	Parameter	Va	lue	Unit
Symbol	Parameter	Min	Max	Onit
V <sub>DD</sub>	Voltage on VDD pin with respect to ground $\mathrm{V}_{\mathrm{SS}}$	-0.3	2.4	V
V <sub>DDQ</sub>	Voltage on VDDQ pin with respect to ground $\mathrm{V}_{\mathrm{SS}}$	-0.3	4.0	V
V <sub>BATT</sub>	Voltage on VBATT pin with respect to ground $\mathrm{V}_{\mathrm{SS}}$	-0.3	4.0	V
AV <sub>DD</sub>	Voltage on AVDD pin with respect to ground V <sub>SS</sub> (128-pinpackage)	-0.3	4.0	v
AV <sub>REF</sub>	Voltage on AVREF pin with respect to ground $V_{SS}$ (128-pin package)	-0.3	4.0	v
AV <sub>REF_AVDD</sub>	Voltage on AVREF_AVDD pin with respect to Ground V <sub>SS</sub> (80-pin package)	-0.3	4.0	v
V <sub>IN</sub>	Voltage on 5V tolerant pins with respect to ground $V_{SS}$	-0.3	5.5	v
♥ IN	Voltage on any other pin with respect to ground $V_{SS}$	-0.3	4.0	v
I <sub>OV</sub>	Input current on any pin during overload condition	-10	+10	mA
I <sub>TDV</sub>	Absolute sum of all input currents during overload condition		200	mA
Τ <sub>ST</sub>	Storage Temperature	-55	+150	°C
TJ	Junction Temperature		+125	°C
ESD	ESD Susceptibility (Human Body Model)	20	00	V

Table 4. Absolute maximum ratings

Note:

Stresses exceeding above listed recommended "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. During overload conditions ( $V_{IN}$ > $V_{DDQ}$  or  $V_{IN}$ < $V_{SSQ}$ ) the voltage on pins with respect to ground ( $V_{SSQ}$ ) must not exceed the recommended values.

#### **Operating conditions** 6.2

Parameter	Test	Va	lue	Unit
Faiametei	Conditions	Min	Max	Unit
Digital CPU supply voltage		1.65	2.0	V
Digital I/O supply voltage		2.7	3.6	V
SRAM backup and RTC supply voltage		2.5	3.6	V
Analog ADC supply voltage (128-pin package)		2.7	3.6	V
Analog ADC reference voltage (128-pin package)		1.0	3.6	V
Combined analog ADC reference and ADC supply voltage (80-pin package)		2.7	3.6	>
Ambient temperature under bias		-40	+85	c
hould be connected to V <sub>DDQ</sub> if no	battery is ins	talled	400	
	Digital I/O supply voltageSRAM backup and RTC supply voltageAnalog ADC supply voltage (128-pin package)Analog ADC reference voltage (128-pin package)Combined analog ADC reference and ADC supply voltage (80-pin package)Ambient temperature under bias	ParameterConditionsDigital CPU supply voltageDigital I/O supply voltageSRAM backup and RTC supply voltageAnalog ADC supply voltage (128-pin package)Analog ADC reference voltage (128-pin package)Combined analog ADC reference and ADC supply voltage (80-pin package)Ambient temperature under bias	ParameterTest ConditionsDigital CPU supply voltage1.65Digital I/O supply voltage2.7SRAM backup and RTC supply voltage2.5Analog ADC supply voltage (128-pin package)2.7Analog ADC reference voltage (128-pin package)1.0Combined analog ADC reference and ADC supply voltage (80-pin package)2.7	ParameterConditionsMinMaxDigital CPU supply voltage1.652.0Digital I/O supply voltage2.73.6SRAM backup and RTC supply voltage2.53.6Analog ADC supply voltage (128-pin package)2.73.6Analog ADC reference voltage (128-pin package)1.03.6Combined analog ADC reference and ADC supply voltage (80-pin package)2.73.6Ambient temperature under bias-40+85

#### Table 5. **Operating conditions**

Notes: 1 The  $V_{BATT}$  pin should be connected to  $V_{DDQ}$  if no battery is installed

#### 6.3 LVD electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

Symbol	Parameter	Test		Value		Unit
Symbol	i didilleter	Conditions	Min	Тур	Max	Onit
V <sub>DD_LVD</sub>	V <sub>DD</sub> LVD Threshold		1.35	1.4	1.45	V
V <sub>DDQ_LVD</sub>	V <sub>DDQ</sub> LVD Threshold	(1)	2.35	2.4	2.45	V
DDQ_LVD	DDQ 1111	(2)	2.65	2.7	2.75	
V <sub>DD_BRN</sub>	V <sub>DD</sub> Brown Out Warning Threshold		1.6	1.65	1.7	V
V <sub>DDQ_BRN</sub>	V <sub>DDQ</sub> Brown Out Warning	(1)	2.6	2.65	2.7	V
BRN	Threshold	(2)	2.9	2.95	3.0	

#### Table 6. **LVD Electrical Characteristics**

Notes: 1 For VDDQ I/O voltage operating at 2.7 - 3.3V.

- 2 For VDDQ I/O voltage operating at 3.0 3.6V.
- 3 Selection of  $V_{DDQ}$  operation range is made using configuration software from ST, or IDE from 3rd parties. The default condition is V<sub>DDQ</sub>=2.7V - 3.3V.



#### 6.4 **DC** electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

Min     Typ     Max $V_{IH}$ Input High Level     General inputs     2.0     (1) $V_{IL}$ Input Low Level     General inputs $0.8V_{DDQ}$ 0.8 $V_{IL}$ Input Low Level     General inputs $0.8V_{DDQ}$ 0.8	Symbol	Parameter	Test Conditions		Value		Uni
VIHInput High LevelRESET and TCK inputs $0.8V_{DDQ}$ InputVILInput Low LevelGeneral inputs $0.8V_{DDQ}$ $0.8$ VHYSInput Hysteresis Schmitt TriggerGeneral inputs $0.4$ $0.2V_{DDQ}$ VHYSInput Hysteresis Schmitt TriggerGeneral inputs $0.4$ $V$ VOHOutput High Level High current pinsI/O ports 3 and 6: Push-Pull, I <sub>OH</sub> = 8mA $V_{DDQ}$ -0.7 $V_{DDQ}$ VOHOutput High Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OH</sub> = 4mA $V_{DDQ}$ -0.7 $V_{DDQ}$ VOLOutput Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 8mA $0.4$ $V_{DDQ}$ VOLOutput Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 4mA $0.4$ $V_{DDQ}$	Symbol	Parameter	Test Conditions	Min	Тур	Мах	
VILInput Low LevelGeneral inputs $0.8V_{DDQ}$ VVHYSInput Hysteresis Schmitt TriggerGeneral inputs $0.4$ $0.2V_{DDQ}$ VHYSInput Hysteresis Schmitt TriggerGeneral inputs $0.4$ $V$ VOHOutput High Level High current pinsI/O ports 3 and 6: Push-Pull, I <sub>OH</sub> = 8mA $V_{DDQ}$ -0.7 $V$ VOHOutput High Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OH</sub> = 4mA $V_{DDQ}$ -0.7 $V$ VOLOutput Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 8mA $0.4$ $V$ VOLOutput Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 4mA $0.4$ $V$ Votes: 1Input pins are 5V tolerant, max input voltage is 5.5V $0.4$ $V$		land that the	General inputs	2.0		(1)	
VILInput Low LevelGeneral inputs0.8 $V_{HYS}$ Input Hysteresis Schmitt TriggerGeneral inputs0.40.2V_{DDQ} $V_{HYS}$ Input Hysteresis Schmitt TriggerGeneral inputs0.4V $V_{OH}$ Output High Level High current pinsI/O ports 3 and 6: Push-Pull, I <sub>OH</sub> = 8mA $V_{DDQ}$ -0.7V $V_{OH}$ Output High Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OH</sub> = 4mA $V_{DDQ}$ -0.7V $V_{OL}$ Output Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 8mA0.4V $V_{OL}$ Output Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 4mA0.4V $V_{OL}$ Notes: 1 Input pins are 5V tolerant, max input voltage is 5.5V0.4V	vIH	Input High Level	RESET and TCK inputs	0.8V <sub>DDQ</sub>			
VHYSInput Hysteresis Schmitt TriggerGeneral inputs0.4VVOH $\begin{array}{c} \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\$	V.		General inputs			0.8	v
VHYSSchmitt TriggerGeneral Inputs0.4V $V_{OH}$ Output High Level High current pinsI/O ports 3 and 6: Push-Pull, I <sub>OH</sub> = 8mA $V_{DDQ}$ -0.7VOutput High Level Standard current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OH</sub> = 4mA $V_{DDQ}$ -0.7V $V_{OL}$ Output Low Level High current pinsI/O ports 3 and 6: Push-Pull, I <sub>OL</sub> = 8mA0.4 $V_{OL}$ Output Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 8mA0.4 $V_{OL}$ Vortuput Low Level High current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, I <sub>OL</sub> = 4mA0.4Notes: 1Input pins are 5V tolerant, max input voltage is 5.5V0.4	۹Ľ		RESET and TCK inputs			0.2V <sub>DDQ</sub>	1
$V_{OH} = \frac{High current pins}{High current pins} = \frac{Push-Pull, I_{OH} = 8mA}{Push-Pull, I_{OH} = 8mA} = \frac{V_{DDQ}-0.7}{V_{DDQ}-0.7}$ $\frac{Output High Level}{Standard current pins} = \frac{I/O ports 0, 1, 2, 4, 5, 7, 8, 9:}{Push-Pull, I_{OH} = 4mA} = \frac{V_{DDQ}-0.7}{0.4}$ $V_{OL} = \frac{Output Low Level}{High current pins} = \frac{I/O ports 3 and 6:}{Push-Pull, I_{OL} = 8mA} = \frac{0.4}{0.4}$ $V_{OL} = \frac{Output Low Level}{Standard current pins} = \frac{I/O ports 0, 1, 2, 4, 5, 7, 8, 9:}{Push-Pull, I_{OL} = 8mA} = \frac{0.4}{0.4}$ $V_{OL} = \frac{Output Low Level}{Standard current pins} = \frac{I/O ports 0, 1, 2, 4, 5, 7, 8, 9:}{Push-Pull, I_{OL} = 4mA} = \frac{0.4}{0.4}$ $V_{OL} = \frac{V_{OL}}{V_{OL}} = \frac{V_{OL}}{V$	V <sub>HYS</sub>		General inputs	0.4			V
Output High Level Standard current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, $I_{OH} = 4mA$ $V_{DDQ}$ -0.7V_{OL}Output Low Level High current pinsI/O ports 3 and 6: Push-Pull, $I_{OL} = 8mA$ 0.4Output Low Level Standard current pinsI/O ports 0,1,2,4,5,7,8,9: Push-Pull, $I_{OL} = 4mA$ 0.4Notes: 1Input pins are 5V tolerant, max input voltage is 5.5V0.4	M			V <sub>DDQ</sub> -0.7			
V <sub>OL</sub> High current pins       Push-Pull, I <sub>OL</sub> = 8mA       0.4       V         Output Low Level       I/O ports 0,1,2,4,5,7,8,9:       0.4       V         Standard current pins       Push-Pull, I <sub>OL</sub> = 4mA       0.4       V         Notes: 1       Input pins are 5V tolerant, max input voltage is 5.5V       0.4       V	∨он			V <sub>DDQ</sub> -0.7			5
Output Low Level       I/O ports 0, 1,2,4,5,7,8,9:         Standard current pins       Push-Pull, I <sub>OL</sub> = 4mA         Notes: 1       Input pins are 5V tolerant, max input voltage is 5.5V						0.4	
005010	Voi				~ <b>(</b> )		v
		Standard current pins	Push-Pull, I <sub>OL</sub> = 4mA	blete	2	0.4	

Table 7. **DC Electrical Characteristics** 

# 6.5 AC electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

Symbol	Parameter	Tost Co	nditions		Value		Unit
Symbol	Faidilletei	Test Co	nunions	Min	Тур	Max	Onit
<b>1</b>	Run Mode Current	All peripherals on	CPU_CLK =		1.7	2.3	mA/
IDDRUN		All peripherals off	96MHz [1] [5]		1.3	1.6	MHz
lini e	Idle Mode Current	All peripherals	on [2] [5]		1.14	1.7	mA/ MHz
IDLE		All peripherals	off [3] [5]		0.45	0.75	mA/ MHz
la. ===	Sleep Mode Current	LVD On [4] [5	]		55	825	μA
ISLEEP		LVD Off [4] [5	]		50	820	μA
I <sub>RTC_STBY</sub>	RTC Standby Current	Measured on V	/BATT pin		0.3	0.9	μA
I <sub>SRAM_STBY</sub>	SRAM Standby Current	Measured on V	/BATT pin		5	85	μA

Table 8.AC electrical characteristics

Notes: 1 ARM core and peripherals active with all clocks on. Power can be conserved by turning off clocks to peripherals which are not required.

- 2 ARM core stopped and all peripheral clocks active.
- 3 ARM core stopped and all peripheral clocks stopped.
- 4 ARM core and all peripheral clocks stopped (with exception of RTC).
- 5 Current measured on the V<sub>DD</sub> pins. V<sub>DDQ</sub> current is not included.





Symbol	Parameter	Test Conditions		Value		Unit
Symbol	Farameter	lest conditions	Min	Тур	Max	Unit
f <sub>MSTR</sub>	CCU Master Clk Output		32.768		96,000	kHz
farment	CPU Core Frequency	Executing from SRAM			96	MHz
fCPUCLK	CFO Cole Flequency	Executing from Flash			96	MHz
f <sub>PCLK</sub>	Peripheral Clock for APB				48	MHz
f <sub>HCLK</sub>	Peripheral Clock for AHB				96	MHz
f <sub>OSC</sub>	Clock Input		4		25	MHz
f <sub>FMICLK</sub>	FMI Flash Bus clock (internal clock)				96	MHz
f <sub>BCLK</sub>	External Memory Bus clock (internal clock)				66	MHz
f <sub>RTC</sub>	RTC Clock			32.768		kHz
f <sub>EMAC</sub>	EMAC PHY Clock			25	, CV	MHz
f <sub>USB</sub>	USB Clock			48	N.	MHz

Table 9. AC electrical characteristics

# 6.6 **RESET\_INn and power-on-reset characteristics**

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

	Symbol	Parameter	Test Conditions	Value			Unit
	Symbol	Farameter	Test conditions	Min	Тур	Max	Onit
	t <sub>RINMIN</sub>	RESET_INn Valid Active Low		100			ns
	t <sub>POR</sub>	Power-On-Reset Condition duration	V <sub>DDQ</sub> ,V <sub>DD</sub> ramp time is less than 10ms	10			ms
	t <sub>RSO</sub>	RESET_OUT Duration (Watchdog reset)		one PCLK			ns
01050	lete						

Table 10. RESET\_INn and Power-On-Reset Characteristics



# 6.7 Main oscillator electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85  $^\circ C$  unless otherwise specified.

#### Table 11. Main oscillator electrical characteristics

Symbol	Parameter	Test Conditions		Unit		
Symbol	i didineter		Min	Тур	Max	Onit
t <sub>STUP(OSC)</sub>	Oscillator Start-up Time	Stable V <sub>DDQ</sub>			3	mS

# 6.8 **RTC oscillator electrical characteristics**

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

#### Table 12. RTC oscillator electrical characteristics

Symbol	Parameter	Test Conditions	Value			Unit
Cymbol	i didineter		Min	Тур	Max	51
9м(RTC)	Oscillator Start _voltage		LVD <sup>1)</sup>			v
t <sub>STUP(RTC)</sub>	Oscillator Start-up Time	Stable V <sub>DDQ</sub>				S

Notes: 1 Min oscillator start voltage is the same as low voltage detect level (2.4V or 2.7V) for VDDQ

#### Table 13. RTC crystal electrical characteristics

	Symbol	Parameter	Test Conditions	Value			Unit
	Symbol	Faidilielei	rest conditions	Min	Тур	Max	
	f <sub>O</sub>	Resonant frequency	<b>O</b> Y		32.768		kHz
	R <sub>S</sub>	Series resistance				40	kΩ
	CL	Load capacitance			8		pF
01050	lete	produc					



# 6.9 PLL electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

 Table 14.
 PLL Electrical Characteristics



# 6.10 Flash memory characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

Parameter			Value			
		Test Conditions	Typ <sup>2)</sup>	Typ after 100K W/E cycles <sup>2)</sup>	Max	Unit
	Primary Bank (512 Kbytes) <sup>1)</sup>		8	9	11.5	s
Bank erase	Primary Bank (256 Kbytes) <sup>1)</sup>		4	4.5	6	s
	Secondary Bank (32 Kbytes)		700	750	950	ms
Sector erase	Of Primary Bank (64 Kbytes)		1300	1400	1800	ms
	Of Secondary Bank (8 Kbytes)		300	320	450	ms
	Primary Bank (512 Kbytes) <sup>1)</sup>		3700	4700	5100	ms
Bank program	Primary Bank (256 Kbytes) <sup>1)</sup>	40	1900	2000	2550	ms
	Secondary Bank (32 Kbytes)	COLOR	250	260	320	ms
Sector program	Of Primary Bank (64 Kbytes)	005	500	520	640	ms
	Of Secondary Bank (8 Kbytes)		60	62	80	ms
Word program			8	9	11	μs
Sector erase timeout	4110					μs

Table 15.	Flash memory program/erase characteristics
-----------	--------------------------------------------

- Notes: 1 STR91xFx44 devices have 512 Kbytes primary Flash, STR91xFx32 devices have 256 Kbytes primary Flash
  - 2  $V_{DD} = 1.8V$ ,  $V_{DDQ} = 3.3V$ ,  $T_A = 25^{\circ}C$ .
  - 3 Flash read access for synchronous addresses is 96 MHz maximum.
  - 4 Flash read access for asynchronous accesses requires 2 wait states when FMI clock is above 66 MHz. See STR91xF Flash Programming Manual for more information.

#### Table 16. Flash memory endurance

Parameter	Test Conditions		Value		Unit
Falameter		Min	Тур	Max	Unit
Program/erase cycles	Per word	100K			cycles
Data retention		20			years



# 6.11 EMC characteristics

Susceptibility tests are performed on a sample basis during product characterization.

## 6.11.1 Functional EMS (Electro Magnetic Susceptibility)

Based on a simple running application on the product (toggling 2 LEDs through I/O ports), the product is stressed by two electro magnetic events until a failure occurs (indicated by the LEDs).

- **ESD**: Electro-Static Discharge (positive and negative) is applied on all pins of the device until a functional disturbance occurs. This test conforms with the IEC 1000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub>, V<sub>DDQ</sub> and V<sub>SS</sub> through a 100pF capacitor, until a functional disturbance occurs. This test conforms with the IEC 1000-4-4 standard.

A device reset allows normal operations to be resumed.

Symbol	Parameter	Conditions	Neg.	Pos.	Unit
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V <sub>DD</sub> =1.8V, V <sub>DDQ</sub> =3.3V, T <sub>A</sub> =+25°C, f <sub>OSC</sub> /f <sub>CPUCLK</sub> =4 MHz/96MHz PLL	-1(1)	>2 <sup>(1)</sup>	
V <sub>FFTB</sub>	Fast transient voltage burst limits to be applied through 100pF on $V_{DD}$ and $V_{DDQ}$ pins to induce a functional disturbance	$V_{DD}$ =1.8V, $V_{DDQ}$ =3.3V, $T_A$ =+25°C, $f_{OSC}/f_{CPUCLK}$ =4 MHz/96 MHz PLL conforms to IEC 1000-4-4	-4(1)	4 <sup>(1)</sup>	kV

Table 17. EMS data

1. Data based on characterization results, not tested in production.

# 6.11.2 Electro Magnetic Interference (EMI)

Based on a simple application running on the product (toggling 2 LEDs through the I/O ports), the product is monitored in terms of emission. This emission test is in line with the norm SAE J 1752/3 which specifies the board and the loading of each pin.

	Symbol	Parameter	Conditions	Monitored Frequency Band	Max vs. [fosc/fcpucLk]	Unit
	0,00				4 MHz/ 96 MHz	
-0	Y			0.1MHz to 30 MHz	10	
5	c	Peak level	$T_A=+25$ °C, LQFP128 package conforming to SAE J 1752/3	30 MHz to 130 MHz	10	dBμV
	S <sub>EMI</sub>			130 MHz to 1GHz	22	
				SAE EMI Level	4	-

# Table 18. EMI data

Notes:

1. Data based on characterization results, not tested in production.

2. BGA and LQFP devices have similar EMI characteristics.



# 6.11.3 Absolute Maximum Ratings (Electrical Sensitivity)

Based on three different tests (ESD, LU and DLU) using specific measurement methods, the product is stressed in order to determine its performance in terms of electrical sensitivity. For more details, refer to the application note AN1181.

# 6.11.4 Electro-Static Discharge (ESD)

Electro-Static Discharges (3 positive then 3 negative pulses separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts\*(n+1) supply pin). Two models can be simulated: Human Body Model and Charge Device Model. This test conforms to the JESD22-A114A/A115A standard.

Table 19. ESD Absolute Maximum ratings

Symbol	Ratings	Conditions	Maximum value <sup>1)</sup>	Unit	
V <sub>ESD(HBM)</sub>	Electro-static discharge voltage (Human Body Model)		+/-2000		
V <sub>ESD(CDM)</sub>	Electro-static discharge voltage (Charge Device Model)		1000		
	terization results, not tested in production.	Pre	201		
Static an	d Dynamic Latch-Up	10,00			

#### Notes:

1. Data based on characterization results, not tested in production.

## 6.11.5 Static and Dynamic Latch-Up

- LU: 3 complementary static tests are required on 10 parts to assess the latch-up performance. A supply overvoltage (applied to each power supply pin) and a current injection (applied to each input, output and configurable I/O pin) are performed on each sample. This test conforms to the EIA/JESD 78 IC latch-up standard. For more details, refer to the application note AN1181.
- DLU: Electro-Static Discharges (one positive then one negative test) are applied to each pin of 3 samples when the micro is running to assess the latch-up performance in dynamic mode. Power supplies are set to the typical values, the oscillator is connected as near as possible to the pins of the micro and the component is put in reset mode. This test conforms to the IEC1000-4-2 and SAEJ1752/3 standards. For more details, refer to the application note AN1181.

# 6.11.6 Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

#### Software recommendations:

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset



• Critical Data corruption (control registers...)

Prequalification trials:

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the RESET pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

### 6.11.7 Electrical Sensitivities

Symbol	Parameter	Conditions	Class <sup>1)</sup>
LU	Static latch-up class	T <sub>A</sub> =+25°C	A
DLU	Dynamic latch-up class	V <sub>DDQ</sub> =3.3V, V <sub>DD</sub> =1.8V, f <sub>OSC</sub> /f <sub>CPUCLK</sub> =4 MHz/96 MHz	А

Notes:

# 6.12 External memory bus timings

 $V_{DDQ} = 2.7 - 3.6V$ ,  $V_{DD} = 1.65 - 2V$ ,  $T_A = -40 / 85$  °C,  $C_L = 30$  pF unless otherwise specified.

#### Table 20. EMI Bus Clock Period

Symbol	Parameter Value	
t <sub>BCLK</sub>	EMI Bus Clock period	1 /(f <sub>HCLK</sub> x EMI_ratio)

- Notes: 1 The internal EMI Bus clock signal is available externally only on LFBGA144 packages (ball M8), and not available on LQFP packages.
  - 2 EMI\_ratio =1/2 by default (can be programmed to be 1 by setting the proper bits in the SCU\_CLKCNTR register)

#### Table 21. EMI read operation

Symbol	Parameter	Value			
Symbol	Farameter	Min	Мах	Unit	
t <sub>RCR</sub>	Read to CSn inactive	-1	+1	ns	
t <sub>RP</sub>	Read Pulse Width	(WSTRD-WSTOEN+1) x t <sub>BCLK</sub> - 1	(WSTRD-WSTOEN+1) x t <sub>BCLK</sub> + 1	ns	
t <sub>RDS</sub>	Read Data Setup Time	4		ns	
t <sub>RDH</sub>	Read Data Hold Time	0		ns	
t <sub>RAS</sub>	Read Address Setup Time	(WSTOEN) x t <sub>BCLK</sub> - 1	(WSTOEN) x t <sub>BCLK</sub> + 1	ns	
t <sub>RAH</sub>	Read Address Hold Time	0		ns	
t <sub>AW</sub>	ALE pulse width	(ALE_LENGTH) x t <sub>BCLK</sub> - 1	(ALE_LENGTH) x t <sub>BCLK</sub> + 1	ns	

<sup>1.</sup> Class description: A Class is an STMicroelectronics internal specification. All its limits are higher than the JEDEC specifications, that means when a device belongs to Class A it exceeds the JEDEC standard. B Class strictly covers all the JEDEC criteria (international standard).

Symbol	Parameter	Value			
Symbol	runneter	Min	Мах	Unit	
t <sub>AAH</sub>	Address to ALE hold time	t <sub>BCLK/2</sub> - 1	t <sub>BCLK/2</sub> + 1	ns	
t <sub>AAS</sub>	Address to ALE setup time	(ALE_LENGTH) x t <sub>BCLK</sub> - 1		ns	

- Notes: 1 ALE\_LENGTH = 1 by default (can be programmed to be 2 by setting the bits In the SCU\_SCR0 register)
  - 2 WSTRD = 1Fh by default (RD wait state time = WSTRD x t<sub>BCLK</sub>, WSTRD can be programmed in the EMI\_RCRx Register)
  - 3 WSTOEN = 1 by default (RD assertion delay from chip select. WSTOEN can be programmed in the EMI\_OECRx Register)









Symbol Parameter		Test	Value			
Symbol	Parameter	Conditions	Min	Мах	Unit	
t <sub>WCR</sub>	WRn to CSn inactive		(t <sub>BCLK</sub> /2) - 1	$(t_{BCLK}/2) + 1$	ns	
t <sub>WP</sub>	Write Pulse Width		(WSTWR-WSTWEN + 1) x t <sub>BCLK</sub> - 1	(WSTWR-WSTWEN + 1) x t <sub>BCLK</sub> + 1	ns	
	Write Data Setup Time (non-mux mode)		(WSTWEN + 1/2) x t <sub>BCLK</sub>		ns	
t <sub>WDS</sub> Write Data Setup Time		ALE length=1 WSTWEN>2	(WSTWEN - 1.5) x t <sub>BCLK</sub>			
	(mux mode )	ALE length=2 WSTWEN>3	(WSTWEN - 2.5) x t <sub>BCLK</sub>			
t <sub>WDH</sub>	Write Data Hold Time		(t <sub>BCLK</sub> /2) - 1	(t <sub>BCLK</sub> /2) + 1	ns	
t <sub>WAS</sub>	Write Address Setup Time		(WSTWEN + 1/2) x t <sub>BCLK</sub> -1 <sup>4)</sup>	(WSTWEN + 1/2) x t <sub>BCLK</sub> +1 <sup>4)</sup>	ns	
t <sub>WAH</sub>	Write Address Hold Time		t <sub>BCLK</sub> /2	01001	ns	
t <sub>AW</sub>	ALE pulse width		(ALE_LENGTH x t <sub>BCLK</sub> )-1	(ALE_LENGTH x t <sub>BCLK</sub> ) +1	ns	
t <sub>AAH</sub>	Address to ALE hold time		(t <sub>BCLK</sub> /2) -1	(t <sub>BCLK</sub> /2) + 1	ns	
t <sub>AAS</sub>	Address to ALE setup time		(ALE_LENGTH x t <sub>BCLK</sub> )- 1		ns	

Table 22. EMI write operation

Notes: 1 ALE\_LENGTH = 1 by default (can be programmed to be 2 by setting the bits In the SCU\_SCR0 register)

2 WSTWR =1Fh by default (WR wait state time = WSTWR x t<sub>BCLK</sub>, WSTWR can be programmed in the EMI\_WCRx Register)

- 3 WSTWEN= 0 by default (WR assertion delay from chip select. WSTWEN can be programmed in the EMI\_WECRx Register)
- 4 When the CPU executes a 16-bit write to a x8 EMI bus, the second write cycle's address setup time is defined as t<sub>WAS</sub>=(WSTWEN ½) x t<sub>BCLK</sub>

ps





Figure 13. Non-Mux Bus (8-bit) write timings





# 6.13 ADC electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

Symbol	Parameter	Test Conditions		Value	Unit	
Symbol	Farameter	Test Conditions	Min	Тур	Max	Unit
V <sub>AIN</sub>	Input Voltage Range		0		AV <sub>REF</sub>	V
RES	Resolution				10	Bits
N <sub>CH</sub>	Number of Input Channels				8	N
f <sub>ADC</sub>	ADC Clock Frequency				25	MHz
	POR bit set to Standby mode				500	ms
-	Conversion Time	f <sub>ADC</sub> = 25 MHz		0.7		μs
	Throughput Rate	f <sub>ADC</sub> = 25 MHz		1400		ksps
C <sub>IN</sub>	Input Capacitance			5		pF
E <sub>D</sub>	Differential Non-Linearity	[1] [2]		1	3	LSB[3]
EL	Integral Non-Linearity	[1]		3	6	LSB
E <sub>O</sub>	Offset Error	[1]		3	6	LSB
E <sub>G</sub>	Gain Error	[1]		0.5	2	LSB
E <sub>T</sub>	Absolute Error	[1]	-0.	4	6	LSB
I <sub>ADC</sub>	Power Consumption		20	4.6	6	mA

 Table 23.
 ADC Electrical Characteristics

Notes: 1 Conditions:  $A_{VSS} = 0$  V,  $AV_{DD} = 3.3$  V  $f_{ADC} = 25$  MHz.

- 2 The A/D is monotonic, there are no missing codes.
- 3 1 LSB = (V<sub>DDA</sub> V<sub>SSA</sub>)/1024





#### Figure 15. ADC conversion characteristics



S

# 6.14 Communication interface electrical characteristics

### 6.14.1 10/100 Ethernet MAC electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85  $^\circ C$  unless otherwise specified.

### **Ethernet MII Interface Timings**

#### Figure 16. MII\_RX\_CLK and MII\_TX\_CLK timing diagram



#### Table 24. MII\_RX\_CLK and MII\_TX\_CLK timing table

Symbol			lue	Unit	
Symbol	i didileter	Symbol	Min	Max	Gim
1	Cycle time	t <sub>c</sub> (CLK)	40	Ċ	ns
2	Pulse duration HIGH	t <sub>HIGH</sub> (CLK)	40%	60%	
3	Pulse duration LOW	t <sub>LOW</sub> (CLK)	40%	60%	
4	Transition time	t <sub>t</sub> (CLK)	2	1	ns

#### Figure 17. MDC timing diagram



#### Table 25. MDC timing table

	Symbol	Parameter	Symbol	Va	Unit	
	Gymbol	landineter	Symbol	Min	Max	onit
Î	10	Cycle time	t <sub>c</sub> (MDC)	266		ns
Î	2	Pulse duration HIGH	t <sub>HIGH</sub> (MDC)	40%	60%	
	3	Pulse duration LOW	t <sub>LOW</sub> (MDC)	40%	60%	
	4	Transition time	t <sub>t</sub> (MDC)		1	ns

#### **Ethernet MII management timings**

### Figure 18. Ethernet MII management timing diagram



Symbol	Parameter	Symbol	Value		Unit
	Falanetei	Symbol	Min	Max	Onit
1	MDIO delay from rising edge of MDC	t <sub>c</sub> (MDIO)		2.83	ns
2	MDIO setup time to rising edge of MDC	T <sub>su</sub> (MDIO)	2.70		ns
3	MDIO hold time from rising edge of MDC	T <sub>h</sub> (MDIO)	-2.03		ns

 Table 26.
 Ethernet MII management timing table

### **Ethernet MII transmit timings**

#### Figure 19. Ethernet MII transmit timing diagram



#### Table 27. Ethernet MII transmit timing table

	Symbol	Parameter	Symbol	Val	ue	Unit
	Symbol	Falameter	Symbol	Min	Max	onit
	1	MII_TX_CLK high to MII_TX_EN valid	t <sub>VAL</sub> (MII_TX_EN)		4.20	ns
	2	MII_TX_CLK high to MII_TX_EN invalid	T <sub>inval</sub> (MII_TX_EN)		4.86	ns
	3	MII_CRS valid to MII_TX_CLK high	T <sub>su</sub> (MII_CRS)	0.61		ns
	4	MII_TX_CLK high to MII_CRS invalid	T <sub>h</sub> (MII_CRS)	0.00		ns
	5	MII_COL valid to MII_TX_CLK high	T <sub>su</sub> (MII_COL)	0.81		ns
<u> </u>	6	MII_TX_CLK high to MII_COL invalid	T <sub>h</sub> (MII_COL)	0.00		ns
305°	7	MII_TX_CLK high to MII_TXD valid	t <sub>VAL</sub> (MII_TXD)		5.02	ns
	8	MII_TXCLK high to MII_TXD invalid	T <sub>inval</sub> (MII_TXD		5.02	ns



# **Ethernet MII Receive timings**





#### Figure 21. Ethernet MII receive timing table

Symbol	Parameter	Symbol	Value		Unit			
Symbol	Farameter	Symbol	Min	Max	Unit			
1	MII_RXD valid to MII_RX_CLK high	T <sub>su</sub> (MII_RXD)	0.81		ns			
2	MII_RX_CLK high to MII_RXD invalid	T <sub>h</sub> (MII_RXD)	0.00		ns			
USB 2.0 Compliant in Full Speed Mode								
CAN interfa	ce electrical characte	eristics	0					
Conforms to CAN 2.0B protocol specification								
	16)							

## 6.14.2 USB electrical interface characteristics

## 6.14.3 CAN interface electrical characteristics

Conforms to CAN 2.0B protocol specification obsolete Product(s)



# 6.14.4 I<sup>2</sup>C electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85 °C unless otherwise specified.

Symbol	Parameter	Standard I <sup>2</sup> C		Fast I <sup>2</sup> C		Unit	
0,		Min	Max	Min	Max	onic	
t <sub>BUF</sub>	Bus free time between a STOP and START condition	4.7		1.3		ms	
t <sub>HD:STA</sub>	Hold time START condition. After this period, the first clock pulse is generated	4.0		0.6		μs	
t <sub>LOW</sub>	LOW period of the SCL clock	4.7		1.3		μs	
t <sub>HIGH</sub>	HIGH period of the SCL clock	4.0		0.6		μs	
t <sub>SU:STA</sub>	Set-up time for a repeated START condition	4.7		0.6		μs	
t <sub>HD:DAT</sub>	Data hold time	0		0	- KI	ns	
t <sub>SU:DAT</sub>	Data set-up time	250		100		ns	
t <sub>R</sub>	Rise time of both SDA and SCL signals		1000	20+0.1C <sub>b</sub>	300	ns	
t <sub>F</sub>	Fall time of both SDA and SCL signals		300	20+0.1C <sub>b</sub>	300	ns	
t <sub>SU:STO</sub>	Set-up time for STOP condition	4.0	105	0.6		μs	
C <sub>b</sub>	Capacitive load for each bus line	5	400		400	pF	

 Table 28.
 I<sup>2</sup>C Electrical Characteristics

Notes: 1 The device must internally provide a hold time of at least 300 ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL

- 2 The maximum hold time of the START condition has only to be met if the interface does not stretch the low period of SCL signal
- 3  $C_b$  = total capacitance of one bus line in pF



## 6.14.5 SPI electrical characteristics

 $V_{DDQ}$  = 2.7 - 3.6V,  $V_{DD}$  = 1.65 - 2V,  $T_A$  = -40 / 85  $^\circ C$  unless otherwise specified.

Symbol	Parameter	Toot Conditions	Value		Unit
	Parameter Test Conditions		Тур	Мах	Unit
f <sub>SCLK</sub>	SPI clock frequency	Master		24	MHz
1/t <sub>c(SCLK)</sub>	SFI Clock frequency	Slave		4	
t <sub>r(SCLK)</sub>	- SPI clock rise and fall times	50pF load	0	.1	V/ns
t <sub>f(SCLK)</sub>		Supr load	0	. 1	V/115
t <sub>su(SS)</sub>	SS setup time	Slave	1		
t <sub>h(SS)</sub>	SS hold time	Slave	1		
t <sub>w(SCLKH)</sub>	SCI K high and low time	Master	1		
t <sub>w(SCLKL)</sub>	SOLK high and low time	LK high and low time Slave			
t <sub>su(MI)</sub>	Data input setup time	Master	5	×	51
t <sub>su(SI)</sub>		Slave	5	$C^{\vee}$	
t <sub>h(MI)</sub>	Data input hold time	Master	6	2	
t <sub>h(SI)</sub>		Slave	01		t <sub>PCLK</sub>
t <sub>a(SO)</sub>	Data output access time	Slave		6	
t <sub>dis(SO)</sub>	Data output disable time	Slave		6	
t <sub>v(SO)</sub>	Data output valid time	Slave (after enable		6	
t <sub>h(SO)</sub>	Data output hold time	edge)	0		
t <sub>v(MO)</sub>	Data output valid time	Master (before capture	0.25		
t <sub>h(MO)</sub>	Data output hold time	edge)	0.25		

 Table 29.
 SPI electrical characteristics







Figure 23. SPI slave timing diagram with CPHA=1







# 7 Package mechanical data



#### Figure 25. 80-Pin Low Profile Quad Flat Package





#### 7.1 Thermal characteristics

The average chip-junction temperature, T<sub>.1</sub> must never exceed 125° C.

The average chip-junction temperature, T<sub>J</sub>, in degrees Celsius, may be calculated using the following equation:

 $T_{J} = T_{A} + (P_{D} \times \Theta_{JA})(1)$ 

Where:

- $T_A$  is the Ambient Temperature in °C,
- Θ<sub>JA</sub> is the Package Junction-to-Ambient Thermal Resistance, in ° C/W,
- $P_D$  is the sum of  $P_{INT}$  and  $P_{I/O}$  ( $P_D = P_{INT} + P_{I/O}$ ),
- $P_{INT}$  is the product of  $I_{DD}$  and  $V_{DD}$ , expressed in Watts. This is the Chip Internal Power.

P<sub>I/O</sub> represents the Power Dissipation on Input and Output Pins;

Most of the time for the applications P<sub>I/O</sub> < P<sub>INT</sub> and may be neglected. On the other hand, P<sub>I/O</sub> may be significant if the device is configured to drive continuously external modules and/or memories. The worst case PINT of the STR91xF is 500mW (IDD x VDD, or 250mA x 2.0V).

An approximate relationship between P<sub>D</sub> and T<sub>J</sub> (if P<sub>I/O</sub> is neglected) is given by: tepro

 $P_{D} = K / (T_{J} + 273^{\circ}C) (2)$ 

Therefore (solving equations 1 and 2):

$$K = P_D x (T_A + 273^{\circ}C) + \Theta_{JA} x P_D^2(3)$$

where:

K is a constant for the particular part, which may be determined from equation (3) by measuring  $P_D$  (at equilibrium) for a known  $T_{A}$ . Using this value of K, the values of  $P_D$ and  $T_{,1}$  may be obtained by solving equations (1) and (2) iteratively for any value of  $T_{A}$ .

Table 30.	Thermal	characteristics
-----------	---------	-----------------

	Symbol	Parameter	Value	Unit
	$\Theta_{JA}$	Thermal Resistance Junction-Ambient LQFP 80 - 12 x 12 mm / 0.5 mm pitch	41.5	°C/W
	Θ <sub>JA</sub>	Thermal Resistance Junction-Ambient LQFP128 - 14 x 14 mm / 0.4 mm pitch	38	°C/W
	ete			
0050				



# 8 Ordering information

	Flash KB	RAM KB	Major Peripherals	Package <sup>1)</sup>
STR910FM32X6	256+32	64	CAN, 40 I/Os	LQFP80, 12x12 mm
STR910FW32X6	256+32	64	CAN, EMI, 80 I/Os	LQFP128, 14x14 mm
STR911FM42X6	256+32	96		LQFP80,
STR911FM44X6	512+32	96	USB, CAN, 40 I/Os	12x12mm
STR912FW42X6	256+32	96	Ethernet, USB, CAN, EMI, 80 I/Os	LQFP128
STR912FW44X6	512+32	96	Ethernet, USB, CAN, EMI, 80 I/Os	LQFF 120

Table 31. Ordering information











# 9 Revision history

	Date	Revision	Changes
	12-Apr-2006	1	Initial release
	28-June-2006	2	Added LFBGA144 package
	28-June-2006		Updated electrical characteristics section
			Changed number of GPIOs in 80 pin packge to 40
		3	Changed EMI_RDYn pin name to EMI_WAITn
	04-Sep-2006		Added RTC clock to description of JRTCK in <i>Table 3</i>
			UART max baud rate changed to 1.5 Mbps in Section 2.20 on page 26
			Modified Figure 2: Clock control on page 15
	01-Feb-2007	4	Removed LFBGA144 package, (transferred to separate STR91xFA datasheet).
obsc	hete Pro	duct	Removed LFBGA144 package, (transferred to separate STR91xFA datasheet).

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