SLUS528A - MARCH 2002 - REVISED AUGUST 2002

## DUAL-SLOT PCI HOT-PLUG POWER CONTROLLER

## FEATURES

- 12-V, -12-V, 3.3-V, 5-V Main Power Switching and Auxiliary 3.3-V Power Switching
- 12-V, -12-V And Auxiliary 3.3-V Power FETs
- Hot-Swap Protection and Control of All Supplies
- Overcurrent Protection for All Supplies
- Isolation of Any Load Fault in One Slot from Any Other Slot
- Undervoltage Monitoring for the Main 12-V, 3.3-V, 5-V and Auxiliary 3.3-V Supplies
- Power Fault Latching
- Overtemperature Shutdown
- Slot Status Readout with Open-Drain LED Drivers
- Mechanical Switch Inputs for Attention Request and Electrical Interlock
- Serial Interface for Power Control, Power Status, Slot Control and Slot Status
- Compatible With $33-\mathrm{MHz}, 66-\mathrm{MHz}$, and 133-MHz Bus Speeds
- Compliant To PCI And PCI-X Hot Plug Specifications
- One TPS2340A Supports Two Slots


## DESCRIPTION

The TPS2340A contains main supply power control, auxiliary supply power control, power FETs for $12 \mathrm{~V},-12 \mathrm{~V}$ and auxiliary 3.3 V supplies, and a serial interface for communications with and control of slots. Each TPS2340A contains supply control and switching for two slots.

The main power control circuits start with all supplies off and hold all supplies off until power to the TPS2340A is valid on all positive supplies. When power is requested via the serial interface, the control circuit applies constant current to the
gates of the power FETs, allowing each FET to ramp load voltage linearly. Each supply can be programmed for a desired ramp rate by selecting a gate capacitor for the power FET for that supply. The power control circuits also monitor load current and latch off that slot if the load current exceeds a programmed maximum value. In addition, once the $12-\mathrm{V}$, the $5-\mathrm{V}$, and the $3.3-\mathrm{V}$ FETs are fully enhanced, the load voltage is monitored. If load voltage drops out of specification after these FETs are fully enhanced, the slot latches off. This provides another level of protection from load fault.

The auxiliary power control circuit switches, ramps, and monitors 3.3-V auxiliary power to each slot. The auxiliary control circuit also controls data switches that connect slot interrupts (power management event [PME] outputs) to the main interrupt PME bus after 3.3-V auxiliary supply is connected. PME is disconnected when a board is turned off or faulted.

Each TPS2340A contains power FETs for 12 V at $500 \mathrm{~mA},-12 \mathrm{~V}$ at 100 mA , and auxiliary 3.3 V at 375 mA for each slot. These power FETs are short-circuit protected, slew rate controlled, and over-temperature protected.

The serial interface communicates with a slot controller using a synchronous serial protocol. The interface communicates with the slot, status LEDs, and mechanical switches with individual, dedicated lines. The interface operates from 3.3-V power but inputs are 5-V tolerant. Status LED drivers are capable of driving $24-\mathrm{mA}$ LEDs via integrated open-drain MOSFETs. Mechanical switch inputs have internal pull-up and hysteresis buffers. The serial interface controls slot power, bus connection, and LED outputs, and monitors board capability, power fault, and switch input status.

## absolute maximum ratings over operating free-air temperature (unless otherwise noted) $\dagger$

| Input voltage range: | P12VIN. <br> M12VIN <br> All others | $\begin{aligned} & \ldots-0.5 \mathrm{~V} \text { to } 15 \mathrm{~V} \\ & \ldots-15.0 \mathrm{~V} \text { to } 0.5 \mathrm{~V} \\ & \ldots-0.5 \mathrm{~V} \text { to } 6 \mathrm{~V} \end{aligned}$ |
| :---: | :---: | :---: |
| Output voltage range: | $\begin{aligned} & \text { P12VO, 5V3VG } \\ & \text { P12VG ........ } \end{aligned}$ | $\begin{aligned} & -0.5 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{P} 12 \mathrm{VIN}}+0.5 \mathrm{~V} \\ & \ldots . . . . . .-0.5 \mathrm{~V} \text { to } 28 \mathrm{~V} \end{aligned}$ |
|  | M12VO, M12VG | $\mathrm{V}_{\mathrm{M} 12 \mathrm{VIN}}{ }^{-0.5} \mathrm{~V}$ to 0.5 V |
| Output current pulse: | P12VO (DC internally limited) | 4 A |
|  | M12VO | 0.8 A |
| Operating virtual temperature range, $\mathrm{T}_{J}$ |  | $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |
| Storage temperature range, $\mathrm{T}_{\text {stg }}$ |  | $-55^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$ |
| Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds |  | $260^{\circ} \mathrm{C}$ |

$\dagger$ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTE 1: All voltages are respect to DGND.
electrical characteristics over recommended operating temperature range, P12VINA $=\mathrm{P} 12 \mathrm{VINB}=$ $12 \mathrm{~V}, \mathrm{~V} 5 I \mathrm{~N}=5 \mathrm{~V}$, DIGVCC = 3.3 V, M12VINA $=\mathrm{M} 12 \mathrm{VINB}=-12 \mathrm{~V}, 3 \mathrm{VAUXI}=3.3 \mathrm{~V}$, all outputs unloaded, $T_{A}=T_{J}$ (unless otherwise noted)

## 5-V/3.3-V Supply

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $5 \mathrm{~V}_{\text {OC }}$ input threshold voltage | ROCSET $=6.04 \mathrm{k} \Omega$ | 43 | 53 | 63 | mV |
| 5VISA, 5VISB voltage fault threshold | After P12VG and 5V3VG good | 4.16 | 4.65 | 4.92 | V |
| 5VISA, 5VISB voltage fault minimum captured pulse |  |  | 75 | 135 | ns |
| 5VSA input bias current | PWRENx = high | -100 |  | 100 | $\mu \mathrm{A}$ |
| 5VISA, 5VISB input bias current | PWREN $x=$ high | 100 | 250 | 500 |  |
| 5VISA, 5VISB bleed current | PWRENx = low, $5 \mathrm{VISx}=5 \mathrm{~V}$ | 5 | 10 | 20 | mA |
| $3 \mathrm{~V}_{\text {OC }}$ Input threshold voltage | ROCSET $=6.04 \mathrm{k} \Omega$ | 53 | 63 | 72 | mV |
| 3VISA, 3VISB voltage fault threshold | After P12VG and 5V3VG good | 2.64 | 2.86 | 3.08 | V |
| 3VISA, 3VISB voltage fault minimum captured pulse time |  |  | 75 | 135 | ns |
| 3VSA, 3VSB input bias current | PWREN $x=$ high | -100 |  | 100 | $\mu \mathrm{A}$ |
| 3VISA, 3VISB input bias current | PWRENx = high | 100 | 250 | 500 |  |
| 3VISA, 3VISB bleed current | PWRENx = low, 5VISx $=5 \mathrm{~V}$ | 5 | 10 | 20 | mA |
| 5V3VGA, 5V3VGB charge current |  | -25 | -20 | -14.5 | $\mu \mathrm{A}$ |
| 5V3VGA, 5V3VGB discharge current |  |  | 200 |  | mA |
| 5V3VGA, 5V3VGB good threshold | $\mathrm{P} 12 \mathrm{VIN}=12 \mathrm{~V}$ | 9.5 | 11 | 11.5 | V |
| 5V3VGA, 5V3VGB turn-off time | $\mathrm{C}_{5 \mathrm{~V} 3 \mathrm{VG}}=0.022 \mu \mathrm{~F},$ <br> 5V3VG falling from $90 \%$ to $10 \%$ |  | 1 | 3.5 | $\mu \mathrm{s}$ |

electrical characteristics over recommended operating temperature range, P12VINA $=\mathrm{P} 12 \mathrm{VINB}=$ $12 \mathrm{~V}, \mathrm{~V} 5 \mathrm{~N}=5 \mathrm{~V}$, DIGVCC $=3.3 \mathrm{~V}$, M12VINA $=\mathrm{M} 12 \mathrm{VINB}=-12 \mathrm{~V}, 3 \mathrm{VAUXI}=3.3 \mathrm{~V}$, all outputs unloaded, $\mathrm{T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{J}}$ (unless otherwise noted) (continued)

## 12-V Supply

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| +12-V Internal NMOS on-resistance | $\begin{array}{\|ll} \hline \text { PWREN }=\mathrm{HIGH}, & \mathrm{I} \mathrm{D}=0.5 \mathrm{~A} \\ \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C} & \\ \hline \end{array}$ |  | 0.18 | 0.3 | $\Omega$ |
|  | PWREN $=$ HIGH, $\quad \mathrm{I}$ D $=0.5 \mathrm{~A}$ |  |  | 0.4 | $\Omega$ |
| -12-V Internal NMOS on-resistance | $\begin{array}{\|ll} \hline \text { PWREN }=\mathrm{HIGH}, & \mathrm{I} \mathrm{D}=0.1 \mathrm{~A} \\ \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C} & \\ \hline \end{array}$ |  | 0.5 | 0.75 | $\Omega$ |
|  | PWREN $=\mathrm{HIGH}, \quad \mathrm{I}, ~=0.1 \mathrm{~A}$ |  |  | 0.9 |  |
| +12-V overcurrent threshold | ROCSET $=6.04 \mathrm{k} \Omega$ | 0.83 | 1.00 | 1.17 | A |
| -12-V overcurrent threshold | ROCSET $=6.04 \mathrm{k} \Omega$ | 0.12 | 0.19 | 0.25 |  |
| P12VOA, P12VOB fault threshold voltage | After P12VG and 5V3VG good | 9.50 | 10.80 | 11.15 | V |
| P12VOA, P12VOB voltage fault minimum captured pulse time |  |  | 75 | 135 | ns |
| M12VGA, M12VGB gate charge current |  | -25 | -20 | -14.5 | $\mu \mathrm{A}$ |
| M12VGA, M12VGB gate discharge current |  |  | 200 |  | mA |
| P12VGA, P12VGB, charge current | Derived from charge pump | 1.0 | 4.0 | 8.5 | $\mu \mathrm{A}$ |
| P12VGA, P12VGB, discharge current |  |  | 100 |  | mA |
| P12VGA, P12VGB good threshold | $\mathrm{P} 12 \mathrm{VIN}=12 \mathrm{~V}$ | 19 | 20.5 | 22 | V |
| Turn-on time | $\begin{array}{ll} \hline \text { PWREN }=\text { HIGH to } \mathrm{M} 12 \mathrm{VO}=-10.4 \mathrm{~V}, \\ \mathrm{C}_{\mathrm{M} 12 \mathrm{VG}}=0.022 \mu \mathrm{~F}, & \mathrm{R}_{\mathrm{L}}=120 \Omega \\ \mathrm{C}_{\mathrm{M} 12 \mathrm{VO}}=50 \mu \mathrm{~F} & \\ \hline \end{array}$ |  | 15 | 20 | ms |
|  | $\begin{array}{lr} \hline \text { PWREN }=\mathrm{HIGH} \text { to } \mathrm{P} 12 \mathrm{VO}=11.4 \mathrm{~V}, \\ \mathrm{C}_{12 \mathrm{PVG}}=0.022 \mu \mathrm{~F}, & \mathrm{R}=24 \Omega \\ \mathrm{C}_{\mathrm{P} 12 \mathrm{VO}}=200 \mu \mathrm{~F} & \\ \hline \end{array}$ |  | 60 | 75 |  |
| Turn-off time | $\begin{aligned} & \text { PWREN }=\text { LOW to P12VO }=0.6 \mathrm{~V}, \\ & \text { CP12VG }=0.022 \mu \mathrm{~F} \end{aligned}$ |  | 1.5 | 3.5 | $\mu \mathrm{s}$ |
|  | $\begin{aligned} & \text { PWREN }=\text { LOW to } \mathrm{M} 12 \mathrm{VO}=-0.6 \mathrm{~V}, \\ & \text { CM12VG }=0.022 \mu \mathrm{~F} \end{aligned}$ |  | 1.5 | 3.5 | $\mu \mathrm{s}$ |
| M12VO bleed current |  |  | -20 | -5 | mA |
| P12VO bleed current |  | 5 | 10 |  |  |

input/output control

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| P12VIN supply current | P12VIN = 12 V |  | 1 | 2 | mA |
| V5IN supply current | $\mathrm{V} 5 \mathrm{IN}=5 \mathrm{~V}$ |  | 1.00 | 2.75 |  |
| DIGVCC supply current | DIGVCC $=3.3 \mathrm{~V}$ |  | 500 | 2000 | $\mu \mathrm{A}$ |
| M12VIN supply current | M12VIN -12 V |  | 250 | 2000 |  |
| 3VAUXI supply current | $3 \mathrm{VAUXI}=3.3 \mathrm{~V}$ |  | 200 | 2000 |  |
| Overcurrent fault response time |  |  | 500 | 960 | ns |
| DIGVCC start-up threshold voltage |  | 2.60 | 2.80 | 2.95 | V |
| DIGVCC stop threshold voltage |  | 2.40 | 2.55 | 2.80 |  |
| V5IN start-up threshold voltage |  | 4.2 | 4.4 | 4.6 |  |
| V5IN stop threshold voltage |  | 3.8 | 4.0 | 4.4 |  |
| P12VIN start-up threshold voltage |  | 10.2 | 10.3 | 11.2 |  |
| P12VIN stop threshold voltage |  | 9.5 | 9.4 | 10.6 |  |

electrical characteristics over recommended operating temperature range, $\mathrm{P} 12 \mathrm{VINA}=\mathrm{P} 12 \mathrm{VINB}=$ $12 \mathrm{~V}, \mathrm{~V} 5 I \mathrm{~N}=5 \mathrm{~V}$, DIGVCC = 3.3 V, M12VINA = M12VINB $=-12 \mathrm{~V}$, 3VAUXI $=3.3 \mathrm{~V}$, all outputs unloaded, $T_{A}=T_{J}$ (unless otherwise noted) (continued)

## noise filter

| PARAMETER | TEST CONDITIONS | MIN | TYP |
| :--- | :---: | :---: | :---: |
| Ignored spike from overcurrent |  | 250 |  |
| Latched spike from overcurrent |  | ns |  |

### 3.3 V AUX and PME

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 3VAUXx overcurrent shutdown |  | 0.95 | 1.15 | 1.40 | A |
| 3VAUXI to 3VAUXx on-resistance | $\mathrm{I}_{3} \mathrm{VAUXX}=-500 \mathrm{~mA}$ |  | 300 | 425 | $\mathrm{m} \Omega$ |
| 3VAUXI undervoltage lockout |  | 1.9 | 2.2 | 2.9 | V |
| 3VAUXx turn-on slew rate |  |  | 1.6 | 3.3 | V/ms |
| 3VAUXx turn-on time from SWx | $\begin{aligned} & \text { from } S W x<0.8 \mathrm{~V}, \\ & \mathrm{C}_{3 V A U X}=150 \mu \mathrm{~F} \end{aligned}$ |  | 3 | 5 | ms |
| 3VAUXx turn-off time from SWx | from $\mathrm{SWx}>2.0 \mathrm{~V}$ |  | 2.5 | 7.0 |  |
| 3VAUXx turn-off time from Faultx | from 3VAUXx overcurrent fault detected |  | 5 | 10 | $\mu \mathrm{s}$ |
| PMEx turn-on time from 3VAUXx | $\begin{aligned} & \text { from 3VAUXx > 3.0 V, } \\ & \mathrm{C}_{3 V A U X x}=150 \mu \mathrm{~F} \end{aligned}$ | 6 | 10 | 17 | ms |
| PMEx turn-off time from SWx | from $S W x>2.0 \mathrm{~V}$, |  | 2 | 4 | $\mu \mathrm{s}$ |
| PMEx turn-off time from Faultx | from 3VAUXx overcurrent fault detected |  | 2 | 4 | $\mu \mathrm{s}$ |
| PMEx switch on-resistance | $\begin{array}{ll} \text { from } S W x<0.8 \mathrm{~V}, & \mathrm{I} D=10 \mathrm{~mA} \\ \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{J}}=25^{\circ} \mathrm{C} & \\ \hline \end{array}$ |  | 5 | 10 | $\Omega$ |

ac switching characteristics


## electrical characteristics over recommended operating temperature range, P 12VINA $=\mathrm{P} 12 \mathrm{VINB}=$ $12 \mathrm{~V}, \mathrm{~V} 5 I \mathrm{~N}=5 \mathrm{~V}$, DIGVCC $=3.3 \mathrm{~V}, \mathrm{M} 12 \mathrm{VINA}=\mathrm{M} 12 \mathrm{VINB}=-12 \mathrm{~V}, 3 \mathrm{VAUXI}=3.3 \mathrm{~V}$, all outputs unloaded, $\mathrm{T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{J}}$ (unless otherwise noted) (continued)

dc electrical characteristics

| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Input threshold voltage ( $\overline{\text { SIL }}$, SOC, SIDI, $\overline{\text { SORR }}$, SORLC, TEST, M66EN, SOLC, $\overline{\text { SOR, SODI, SIC) }}$ |  | 0.8 | 1.4 | 2.0 | V |
| High-level input threshold voltage (SWA, SWB, BUTTONA, BUTTONB, PRSNT1A, PRSNT2A, PRSNT1B, PRSNT2B, PGOOD) |  | 2.0 | 2.4 | 2.8 |  |
| Low-level input threshold voltage (SWA, SWB, BUTTONA, BUTTONB, PRSNT1A, PRSNT2A, $\overline{\text { PRSNT1B, }}, \overline{\text { PRSNT2B }}$ |  | 0.8 | 1.2 | 1.6 |  |
| Low-level input threshold voltage (PGOOD) |  | 0.1 | 0.4 | 0.8 |  |
| $\frac{\text { Input hysteresis (SWA }}{\text { PRSNT1A }}, \frac{\text { SWB, } \overline{\text { BUT }}}{\text { PRSNT2A }}, \frac{\text { PRSNT1B }}{\text { PRSN }}, \overline{\text { BUTTONB }},$ |  | 0.4 | 1.0 | 1.6 |  |
| Input hysteresis (PGOOD) |  | 1.5 | 2.0 | 2.5 |  |
| High-level output voltage ( $\overline{\text { BUSENA }}$, $\overline{\text { BUSENB }}$ ) | $\mathrm{I}_{\mathrm{L}}=-8 \mathrm{~mA}$ | 2.4 | 2.8 |  |  |
| Low-level output voltage ( $\overline{\text { BUSENA }}$, $\overline{B U S E N B}$ ) | $\mathrm{L}=16 \mathrm{~mA}$ |  | 0.2 | 0.5 |  |
| Low-level output voltage ( $\overline{\text { PWRLEDA }}, \overline{\text { PWRLEDB }}$, $\overline{\text { ATTLEDA, }}$ ATTLEDB | $\mathrm{L}=24 \mathrm{~mA}$ |  | 0.4 | 0.8 |  |
| Low-level output voltage (all other outputs) | $\mathrm{L}_{\mathrm{L}}=4 \mathrm{~mA}$ |  | 0.2 | 0.5 |  |
| 3.3 V pull-up resistor impedance (inputs pulled up to 3.3 V ) |  | 30 |  | 200 | $\mathrm{k} \Omega$ |
| 5 V pull-up resistor impedance (inputs pulled up to 5 V ) |  | 30 |  | 200 |  |
| Pull-down resistor impedance (inputs with pull-down) |  | 30 |  | 200 |  |
| PCIXCAPA, PCIXCAPB resistor for 133 MHz | Open circuit recommended | 30 |  |  |  |
| PCIXCAPA, PCIXCAPB resistor for 66 MHz | $10 \mathrm{k} \Omega$ connection to DIGGND recommended | 6 |  | 14 |  |
| PCIXCAPA, PCIXCAPB resistor for 33 MHz | $0 \mathrm{k} \Omega$ connection to DIGGND recommended |  |  | 1 |  |

recommended operating conditions

|  | MIN | MAX | UNIT |
| :---: | :---: | :---: | :---: |
| Input voltage, P12VINA, P12VINB | 10.8 | 13.2 | V |
| Input voltage, V5IN | 4.75 | 5.25 |  |
| Input voltage, DIGVCC | 3.1 | 3.5 |  |
| Input voltage, M12VINA, M12VINB | -13.2 | -10.8 |  |
| Input voltage, 3VAUXI | 3.1 | 3.5 |  |
| Load current, $\overline{\text { PWRLEDA }}$, $\overline{\text { PWRLEDB }}$, $\overline{\text { ATTLEDA }}$, $\overline{\text { ATTLEDB }}$ | 0 | 24 | mA |
| Load current, P12VOA, P12VOB | 0 | 500 |  |
| Load current, M12VOA, M12VOB | 0 | 100 |  |
| Load current, 3VAUXA, 3VAUXB | 0 | 375 |  |



## Terminal Functions

| TERMINAL |  | 1/0 | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| NAME | NO. |  |  |
| 3VAUXI | 66 | 1 | 3.3Vaux voltage supply input. A $0.1-\mu \mathrm{F}$ bypass capacitor to PWRGND is recommended. |
| 3VAUXA | 65 | 0 | 3.3Vaux voltage supply outputs. A $0.01-\mu \mathrm{F}$ bypass capacitor to PWRGND is recommended. |
| 3VAUXB | 64 | O |  |
| 3VISA | 57 | 1 | Connect to the load side of the sense resistor. See definition for 3VS. This pin has a switched FET to ground to discharge any output load capacitance when the output is turned off. A $0.01-\mu \mathrm{F}$ bypass capacitor to ANAGND is recommended. |
| 3VISB | 4 | 1 |  |
| 3VSA | 58 | 1 | Connect to the source side of the 3.3-V FET switch. This pin in conjunction with the 3VIS pin sense the current to the 3.3-V load by sensing the voltage drop across a sense resistor. A $0.01-\mu \mathrm{F}$ bypass capacitor to ANAGND is recommended. |
| 3VSB | 3 | 1 |  |
| 5V3VGA | 59 | 0 | Gate drive for the $5-\mathrm{V}$ and $3.3-\mathrm{V}$ FET switches. Ramp rate is programmed by external capacitance connected from this pin to PWRGND. The capacitor is charged with a $20-\mu \mathrm{A}$ current source and discharged with a switch. The output UV circuitry is disabled until the voltage on this pin is greater than 11 V and the voltage on P12VGx is greater than 20 V . |
| 5V3VGB | 2 | O |  |
| 5VISA | 56 | 1 | Connect to the load side of the sense resistor. See definition for 5VS. 5VIS is also used to sense the output voltage for the 5-V UV circuit. This pin has a switched FET to ground to discharge any output load capacitance when the output is turned off. A $0.01-\mu \mathrm{F}$ bypass capacitor to ANAGND is recommended. |
| 5VISB | 5 | 1 |  |
| 5VSA | 55 | 1 | Connect to the source of the 5-V FET switch. This pin in conjunction with the 5VIS pin senses the current to the 5 V load by sensing the voltage drop across the sense resistor. A $0.01-\mu \mathrm{F}$ bypass capacitor to ANAGND is recommended. |
| 5VSB | 6 | 1 |  |
| ANAGND | 54 | - | Ground pin for the low level analog section |
| $\overline{\text { ATTLEDA }}$ | 47 | 0 | $\overline{\text { ATTLEDx }}$ is a high-current, low-true, open-drain output with a $100-\mathrm{k} \Omega$ pull-up resistor to V5IN. |
| $\overline{\text { ATTLEDB }}$ | 13 | 0 |  |
| BUSENA | 32 | 0 | Output-to-bus enable FET switches. These outputs typically enable PCI clocks to the PCI connector. |
| $\overline{\text { BUSENB }}$ | 29 | O |  |
| $\overline{\text { BUTTONA }}$ | 45 | 1 | PCI hot-plut attention notification (momentary) button inputs. Low indicates attention. These input have hysteresis and a $100-\mathrm{k} \Omega$ pull-up to DIGVCC, requiring only a capacitor to ground for debouncing mechanical noise. |
| $\overline{\text { BUTTONB }}$ | 15 | 1 |  |
| $\overline{\text { CLKENA }}$ | 33 | 0 | Output-to-clock enable FET switches. These outputs typically enable PCI clocks to the PCI connector. |
| $\overline{\text { CLKENB }}$ | 28 | 0 |  |
| DIGVCC | 30 | 1 | Power pin for the digital section, connect to 3.3 V . A $0.1-\mu \mathrm{F}$ bypass capacitor from DIGVCC to DIGGND is recommended. |
| DIGGND1 | 31 | - | Ground pins for the digital section. |
| DIGGND2 | 18 | - |  |
| DIGGND3 | 43 | - |  |
| M12VGA | 69 | 0 | A capacitor connected from this pin to M12VO programs the ramp rate of the 12-V switched output. The capacitor is charged with a $20-\mu \mathrm{A}$ current source and discharged with a switch. |
| M12VGB | 78 | 0 |  |
| M12VINA | 68 | 1 | $-12-\mathrm{V}$ input voltage to the device and the -12-V power FET. M12VINA and M12VINB must be tied together and are internally connected by a high-resistance path. The tab on the back of the package is also connected to M12VIN. A $0.1-\mu \mathrm{F}$ bypass capacitor from M12VIN to PWRGND is recommended. |
| M12VINB | 79 | 1 |  |
| M12VOA | 67 | 0 | -12-V Switched output. This pin has a switched FET to ground to discharge any output load capacitance when the output is turned off. A $0.01-\mu \mathrm{F}$ bypass capacitor to PWRGND is recommended. |
| M12VOB | 80 | 0 |  |
| M66ENA | 34 | 1 | PCI 66 MHz -capable bit for slot A . This pin has a $100-\mathrm{k} \Omega$ pull-up to 3 VISA. This pin is typically tied to the PCl connector. |
| M66ENB | 27 | 1 | PCI 66 MHz -capable bit for slot B . This pin has a $100-\mathrm{k} \Omega$ pull-up to 3 VISB. This pin is typically tied to the PCl connector. |
| OCSET | 7 | 1 | A resistor connected between this pin and ANAGND sets the overcurrent threshold of the 4 FET switches. The $+12-\mathrm{V}$ and $-12-\mathrm{V}$ switches are set for the maximum permissible currents per the PCl specification when a $1 \%, 6.04-\mathrm{k} \Omega$ resistor is used. A $0.1-\mu \mathrm{F}$ bypass capacitor from OCSET to ANAGND is recommended. |

Terminal Functions

| TERMINAL |  | I/O |  |  |
| :--- | :---: | :---: | :--- | :--- |
| NAME | NO. |  | DESCRIPTION |  |

NOTE 1: PGOOD input: diagram:


Terminal Functions

| TERMINAL |  | I/O |  |  |
| :--- | :---: | :---: | :--- | :--- |
| NAME | NO. | DESCRIPTION |  |  |

functional block diagram


## APPLICATION INFORMATION

Figure 1 shows three TPS2340A devices cascaded in a system to control six PCI or PCI-X hot-plug slots. A hot-plug controller communicates with three TPS2340A devices over a nine-signal serial bus. Three signals are used to gather status information from the slots, and six signals are used to control the slots. In this bus, seven of the nine signals are connected in parallel. The remaining signals facilitate a cascaded bus and are comprised of the signal pair SIDI/SIDO and SODI/SODO. The SIDO signal drives the PCI hot-plug controller serial input data pin or the SIDI pin of another up-stream TPS2340. The SODI signal is driven by the PCI hot-plug controller and is cascaded down to the next TPS2340A via the SODO signal.
The SIDO and SODO outputs are dual-function pins that also serve as mode-select pins upon the rising edge of PGOOD. See Digital Circuits section for more details.


Figure 1. Hot-Plug System Block Diagram

## analog circuits

## power controller

The functional block diagram shows the TPS2340A with detailed information on the analog functions. For clarity, circuits for only one slot are shown in detail.

## 3.3-V auxiliary supply (3.3-Vaux) support

In today's hot-plug systems, a 3.3Vaux rail remains live while the system power is shut down to allow implementation of the PCI Power Management Specification, version 1.1. The TPS2340A provides a 3.3-Vaux input pin that is monitored for undervoltage and generates a 3.3-Vaux supply pin for each hot-plug slot that is monitored for overcurrent. These supply pins are switched by the slot-specific SWx slot switch inputs. When the slot switch is opened, the slot-specific 3.3-Vaux supply is disabled to allow safe removal of the adapter card. When the slot switch is closed (SWx grounded), the slot-specific 3.3-Vaux supply is restored.

## APPLICATION INFORMATION

## power cycling and $\overline{\text { PME }}$

The PCI Power Management Specification defines a signal called $\overline{\text { PME ( }}$ (power management event) to allow power management events to be communicated back to the system. The TPS2340A provides a slot-specific $\overline{\text { PMEx }}$ input and a gated PMEO output pin that is monitored by the system. The gated PMEO output is controlled by the SWx slot switch similar to the 3.3 -Vaux supply pins described above, but with a delay during connection as shown below. The purpose of the delay is to ensure that 3.3 -Vaux power is stable to the slot before connecting the $\overline{\text { PMEx }}$ signal. (If the device were to observe the $\overline{\mathrm{PMEx}}$ signal while 3.3 -Vaux power was still ramping up, a false trigger could result.)

The 3.3-Vaux circuitry provides short-circuit fault detection. In the event of a fault, the slot 3.3-Vaux and $\overline{\text { PME }}$ signals are immediately disconnected. The fault state is latched internally in the TPS2340A and is cleared by opening the SWx slot switch or by the removal of the 3.3-Vaux power supplying the TPS2340.


Figure 2. 3.3Vaux and PME Gating
When SWx is closed, 3VAUXx power is immediately applied to the slot with controlled slew rate, minimizing inrush current into 3VAUXx bypass capacitors. After 3VAUXx power is above threshold, a delay timer starts. At the end of the delay timer cycle, the PMEx switches close, allowing connection of the PMEx signal to the PMEO output. When multiple TPS2340A devices are used in a system, the PMEO output pins can be connected to the same node, creating a wired-OR PME bus that can be connected to a system interrupt input.

When SWx is opened or if there is a power fault on channel x , the $\overline{\mathrm{PMEx}}$ switch for that channel is immediately opened and the 3VAUXx power for that channel is removed. Although these events happen at approximately the same time, the 3VAUXx power should remain high until the PMEx switch is open so that falling 3VAUXx power doesn't cause a nuisance $\overline{\text { PMEx }}$ interrupt. To insure that 3VAUXx remains high during a power fault, 3VAUXx should have a bypass capacitance of at least $20 \mu \mathrm{~F}$. If the capacitor is not available on the inserted card, it should be provided on the system board.

## APPLICATION INFORMATION

## fault handling

When PGOOD is asserted (main power is valid), the serial interface is available for use. At this time, $\overline{\text { Auxflt }}$ and $\overline{\text { Faultx }}$ (i.e. $\overline{\text { Faultx }}=\overline{\text { Pwrfltx }} \bullet \overline{\text { Auxflt }}$ ) signals are present in the registers and can be read back over the serial interface.
When PGOOD is deasserted (main power is not valid), the serial interface is inactive, so the TPS2340A provides an alternate path for observing the Auxfltx signals. In this case, $\overline{\text { Auxfltx }}$ is presented on the $\overline{\text { ATTLEDx }}$ open-drain output, so 3Vaux faults can be observed. Systems that do not require Auxfltx indications when main power is off should use main power to supply current to the attention LEDs.
In addition, when TEST is asserted after PGOOD has changed from 0 to 1 , the AND of Pwrfftx and $\overline{\text { Auxfltx }}$ is present on the ATTLEDx open-drain output, so the attention LEDs can be used to report any faults on each channel.

After PGOOD is asserted with TEST deasserted and the voltage on the SWx pin is greater than V VIGVCC but less than $\mathrm{V}_{\mathrm{V} \text { IIN }}$, $\overline{\text { ATTLEDx }}$ follows the fault status of the main power fault ( $\left.\overline{\mathrm{Pwrfltx}}\right)$. If the SWx input is pulled to $\mathrm{V}_{\mathrm{V} 5 \text { IN }}$ potential, that slot becomes immediately enabled independent of serial interface status and $\overline{\text { ATTLEDx }}$ follows the fault status of the main power fault ( $\overline{\mathrm{Pwrfflt}})$.

Table 1. Fault Reporting Using ATTLED

| SWx | PGOOD | TEST | $\overline{\|c\|} \overline{\text { ATTLEDx }}$ |
| :--- | :---: | :---: | :--- |
| $<$ V DIGVCC | 0 | x | $\overline{\text { Auxfltx }}$ |
| $<$ V DIGVCC | 1 | 1 | $\overline{\text { Pwrfltx }} \bullet \overline{\text { Auxfltx }}$ |
| $>$ V $_{\text {DIGVCC }},<$ V V5IN | 1 | 0 | $\overline{\text { Pwrfltx }}$ |
| $=$ V VIIN | x | x | $\overline{\text { Pwrfltx }}$ |

## PCI-XCAP resolution

Add-in cards indicate their $\mathrm{PCI}-X$ mode and frequency capabilities by connection of the PCI-XCAP pin. The TPS2340A utilizes a three-level comparator to resolve the PCI-XCAP input so as to decode operating mode and generate the PCI-XCAP2 and PCI-XCAP1 internal signals. With PCI-XCAP open, the TPS2340A selects $\mathrm{PCI}-\mathrm{X} 133-\mathrm{MHz}$ operation. $\mathrm{PCI}-\mathrm{X} 66-\mathrm{MHz}$ operation is selected by a $10-\mathrm{k} \Omega$ pull-down resistor on $\mathrm{PCI}-\mathrm{XCAP}$. With PCI-XCAP grounded, conventional PCI-2.2 operation is selected. This decoding conforms to the standard $\mathrm{PCI}-\mathrm{X}$ specification for $\mathrm{PCI}-\mathrm{XCAP}$ resolution.

Table 2. PCI-XCAP Pin State Resolution

| PCI-XCAP | PCI-XCAP2 | PCI-XCAP1 | DESCRIPTION |
| :--- | :---: | :---: | :--- |
| GND | 0 | 0 | Conventional PCI 2.2 operation |
| $10-\mathrm{k} \Omega$ Pull-down Resistor | 0 | 1 | PCI-X 66-MHz operation |
| Open Circuit | 1 | 1 | PCI-X 133-MHz operation |

## APPLICATION INFORMATION

## thermal shutdown

Under normal operating conditions, the power dissipation in the TPS2340A is low enough that junction temperature ( $\mathrm{T}_{\mathrm{J}}$ ) is not more than $15^{\circ} \mathrm{C}$ above air temperature. However, in the case of a load that exceeds PCl specifications but remains under the TPS2340A current-limit threshold, power dissipation can be higher. To prevent any damage from an out-of-specification load or severe rise in ambient temperature, the TPS2340A contains two independent thermal shutdown circuits, one for each slot's main supplies.
The highest power dissipation in the TPS2340A is from the 12-V power FET so the TPS2340A temperature sense elements are integrated closely with these FETs. These sensors indicate when the temperature at these transistors exceeds approximately $150^{\circ} \mathrm{C}$, either due to average TPS2340A power dissipation, $12-\mathrm{V}$ power FET power dissipation, or a combination of both.

When excessive junction temperature is detected on one slot, that slot fault latch is set and remains set until junction temperature drops by approximately $10^{\circ} \mathrm{C}$ and the slot is turned off, then on again through the serial interface. The other slot is not affected by this event. Also, 3.3-Vaux is not affected by thermal events.

## digital circuits

The TPS2340A implements two independent slots of hot-plug power management. In addition, the TPS2340A contains digital circuits to allow communication between these two slots and the system's hot-plug controller. Also, the TPS2340A contains inputs for connection to mechanical switches and outputs for driving indicator LEDs to directly communicate with users and service.
This digital logic implements a serial-to-parallel converter that accepts data from the serial interface and presents that data to each slot. Data from the serial interface drives the Pwrenx internal main power control signal. Additional data from the serial interface includes $\overline{B U S E N x}$ and $\overline{\text { CLKENx }}$ outputs to activate bus switches, $\overline{\text { RESETx }}$ outputs to allow reset to each slot, and $\overline{\text { ATTLEDx }}$ and $\overline{\text { PWRLEDx }}$ outputs to indicate slot status to users.

The serial-to-parallel converter also accepts data from the slot and sends that data back on the serial interface. Data to the serial interface includes SWx and BUTTONx switch state, PRSNTnx bits, M66ENx and PCI-XCAPx bits to indicate board type, and internal fault signals Auxfltx and Faultx.

When the digital circuits are operating in normal mode, the serial interface is enabled and controls the slots. The TPS2340A also operates in various test modes, allowing board testing and system development.
Each TPS2340A controls two PCI/PCI-X slots. For systems requiring more than two slots, additional TPS2340A devices can be cascaded without requiring an additional hot-plug controller. To maintain system reliability and data integrity at full speed, bus length is often limited to six PCl slots. When higher speed $\mathrm{PCI}-\mathrm{X}$ protocols are active, bus length may be limited to even fewer active slots. However, the TPS2340A is not limited in cascade capability or bus length.

## APPLICATION INFORMATION

## power on configuration

Table 3. lists the various modes of operation and the proper pin states that are necessary to achieve these modes. The shaded rows indicate test modes.

Table 3. Power On Configuration

| PGOOD | TEST | SODO/MODE1 | SIDO/MODEO | OPERATING MODE |
| :---: | :---: | :---: | :---: | :--- |
| $\uparrow$ | 0 | 1 | 1 | Operational mode |
| $\uparrow$ | 1 | 0 | 0 | NAND tree test mode |
| $\uparrow$ | 1 | 0 | 1 | Tri-state test mode (all pins tri-stated) |
| $\uparrow$ | 1 | 1 | X | Reserved |
| $\uparrow$ | 0 | 0 | X | Reserved |
| $\uparrow$ | 0 | X | 0 | Reserved |
| 1 | 1 | SODO | SIDO | Normal operation, but Pwrenx driven on $\overline{\text { PWRLEDx }}$, Faultx <br> driven on $\overline{\text { ATTLEDx. }}$ |
| 1 | 0 | SODO | SIDO | Normal operation |

NOTE: $X=$ Don't care, $\mathrm{x}=$ slot A or B

## operational mode functional description

When both mode pins are 1 and the test pin is 0 , on the rising edge of PGOOD, the TPS2340A enters operational mode. In this mode, the PCI hot-plug controller is able to address multiple TPS2340A ICs and multiple slots. Input status for all slots is grouped into four channels that can be requested by the PCl hot-plug controller.

## channel selection

The PCI hot-plug controller indicates to the TPS2340A which channel to read back via signaling on the $\overline{\text { SIL }}$ pin. The $\overline{\text { SIL }}$ pin asserts low for one SIC clock to indicate a start bit, followed by three SIC clocks of channel address information in the order of channel address LSB followed by channel address MSB and a reserved bit. So as an example, if the hot-plug controller desires to read non-interrupt capable input data from the TPS2340A (channel 01 binary), the $\overline{\text { SIL }}$ sequence is shown in Table 3.

Table 4. Input Channel Selection Example - Channel 01 Selected

| CLOCK | $\overline{\text { SIL }}$ STATUS |
| :---: | :--- |
| 0 | Start bit (low) |
| 1 | Channel address LSB (high) |
| 2 | Channel address MSB (low) |
| 3 | Reserved (high) |
| $4-47$ | High |

## APPLICATION INFORMATION

## operational mode input channel address grouping

Table 5. lists the functions assigned to the channel addresses.
Table 5. Operational Mode Channel Address Grouping

| CHANNEL <br> ADDRESS | FUNCTIONAL CHANNEL GROUP |
| :---: | :--- |
| 00 | Interrupt capable inputs |
| 01 | Non-interrupt capable inputs |
| 10 | Diagnostic data \#1 |
| 11 | Diagnostic data \#2 |

As devices are cascaded, SIDI pin data is received from devices down-stream and passes through the device, shift-register style, to eventually reach the SIDO pin and the PCI hot-plug controller. The data contained in each input channel group is described in Tables 6, 7, 8, and 9.

Table 6. Channel 00 Data Group - Interrupt Capable Inputs

| BIT <br> NUMBER | FUNCTION |
| :---: | :--- |
| 0 | Slot A SW (1 = interlock A open $)$ |
| 1 | Slot A button state (1 = button A pushed) |
| 2 | Slot A power fault state ( $\overline{\text { PwrfltA } \& ~ \overline{\text { AuxfltA }})}$ |
| 3 | Slot A $\overline{\text { PRSNT2 }}$ |
| 4 | Slot A $\overline{\text { PRSNT1 }}$ |
| 5 | Reserved |
| 6 | Reserved |
| 7 | Reserved |
| 8 | Slot B SW |
| 9 | Slot B button state |
| 10 | Slot B power fault state ( $\overline{\text { PwrfltB } \& \overline{\text { AuxfltB }})}$ |
| 11 | Slot B $\overline{\text { PRSNT2 }}$ |
| 12 | Slot B $\overline{\text { PRSNT1 }}$ |
| 13 | Reserved |
| 14 | Reserved |
| 15 | Reserved |
| $16 \ldots 47$ | SIDI pin data for slots C, D, E and F follows |

## APPLICATION INFORMATION

Table 7. Channel 01 Data Group - Non-Interrupt Capable Inputs

| BIT <br> NUMBER | FUNCTION |
| :---: | :--- |
| 0 | Slot A M66EN |
| 1 | Slot A PCI-XCAP1 |
| 2 | Slot A PCI-XCAP2 |
| 3 | Slot A Aux Power Fault state ( $\overline{\text { AuxfltA }})$ |
| 4 | Reserved |
| 5 | Reserved |
| 6 | Reserved |
| 7 | Reserved |
| 8 | Slot B M66EN |
| 9 | Slot B PCI-XCAP1 |
| 10 | Slot B PCI-XCAP2 |
| 11 | Slot B Aux Power Fault state ( $\overline{\text { AuxfltB }})$ |
| 12 | Reserved |
| 13 | Reserved |
| 14 | Reserved |
| 15 | Reserved |
| $16 \ldots 47$ | SIDI pin data for slots C, D, E and F follows |

Table 8. Channel 10 Data Group - Diagnostic Channel \#1

| BIT <br> NUMBER | FUNCTION |
| :---: | :--- |
| 0 | Device Present <br> $1=$ Power controller is present <br> $0=$ Power controller is not installed |
| 1 | Slot A Pwren state |
| 2 | Slot A $\overline{\text { CLKEN }}$ state $(0=$ clock enabled $)$ |
| 3 | Slot A $\overline{\text { BUSEN }}$ state $(0=$ bus enabled $)$ |
| 4 | Slot A $\overline{\text { PCIRST state }(0=\text { reset asserted })}$ |
| 5 | Slot A PWRLED state $(1=$ power LED on $)$ |
| 6 | Slot A $\overline{\text { ATTLED state }(1=\text { attention LED on })}$ |
| 7 | Reserved |
| 8 | Device Present <br> $1=$ Power controller is present <br> $0=$ Power controller is not installed |
| 9 | Slot B Pwren state |
| 10 | Slot B $\overline{\text { CLKEN }}$ state |
| 11 | Slot B $\overline{\text { BUSEN }}$ state |
| 12 | Slot B $\overline{\text { PCIRST } \text { state }}$ |
| 13 | Slot B $\overline{\text { PWRLED } \text { state }}$ |
| 14 | Slot B $\overline{\text { ATTLED state }}$ |
| 15 | Reserved |
| $16 \ldots 47$ | SIDI pin data for slots C, D, E and F follows |

Table 9. Channel 11 Data Group - Diagnostic Channel \#2

| BIT <br> NUMBER | FUNCTION |
| :---: | :--- |
| 0 | Slot A latched Mode 0 bit state. This bit is latched at PGOOD |
| 1 | Slot A latched Mode 1 bit state. This bit is latched at PGOOD. |
| 2 | Must be set to 1. |
| 3 | Reserved |
| 4 | Reserved |
| 5 | Reserved |
| 6 | Reserved |
| 7 | Reserved |
| 8 | Slot B latched Mode 0 bit state. This bit is latched at PGOOD. |
| 9 | Slot B latched Mode 1 bit state. This bit is latched at PGOOD. |
| 10 | Must be set to 1. |
| 11 | Reserved |
| 12 | Reserved |
| 13 | Reserved |
| 14 | Reserved |
| 15 | Reserved |
| $16 \ldots 47$ | SIDI pin data for slots C, D, E and F follows |

Figure 3. illustrates the input scan chain timing for operational mode.


Figure 3. Serial Input Data Timing

## APPLICATION INFORMATION

Similar to input data, serial output data has been ordered to facilitate cascading TPS2340A devices. Assuming a six-slot implementation, as input data arrives in the order of slot A first, followed by data for slots B through F, output data is delivered in reverse order, slot F data first, then data for slots E through A. For a PCI hot-plug controller that supports 6 slots, 48 bits ( 8 per slot) are shifted out by the controller any time a slot control state needs to be changed. At the conclusion of a shift-out process, output data for slots $A$ and $B$ is contained in the device (electrically) closest to the PCI hot-plug controller. Output data for each slot is delivered in the bit order described in Table 10.

## Table 10. Operational Mode Output Data Bit Order

| BIT <br> NUMBER | FUNCTION |
| :---: | :--- |
| 0 | Slot F Pwren state ( $1=$ turn on power $)$ |
| 1 | Slot F CLKEN state $(1=$ enable the clock $)$ |
| 2 | Slot F BUSEN state $(1=$ enable the bus $)$ |
| 3 | Slot F $\overline{\text { PCIRST state }(0=\text { assert } \overline{\text { PCIRST }})}$ |
| 4 | Slot F PWRLED state $(1=$ turn on power LED $)$ |
| 5 | Slot F ATTLED state $(1=$ turn on attention LED $)$ |
| 6 | Reserved |
| 7 | Reserved |
| 8 | Slot E Pwren state |
| 9 | Slot E CLKEN state |
| 10 | Slot E BUSEN state |
| 11 | Slot E $\overline{\text { PCIRST state }}$ |
| 12 | Slot E PWRLED state |
| 13 | Slot E ATTLED state |
| 14 | Reserved |
| 15 | Reserved |
| $16 \ldots 47$ | SODO pin data for slots D, C, B and A follows |

The output data for the six slots is shifted out using the SOC clock. After the appropriate data pattern is established in the shift registers, this data is parallel clocked into registers via latch clocks. These registers drive the output pins and controls. The SOLC pin serves as the latch clock for the Pwrenx, $\bar{B} U S E N x, \overline{C L K E N x}$, $\overline{\text { PWRLEDx }}$, and $\overline{\text { ATTLEDx }}$ outputs.

The SORLC pin serves as the latch clock for the RESETx pins. Two latch clocks are used so that the PCI bus timing requirements between $\overline{\text { REQ64, }} \overline{\text { TRDY }}, \overline{\text { DEVSEL, }}$ and $\overline{\text { STOP }}$ and the slot's $\overline{\text { RST }}$ signal can be more easily controlled by the hot-plug controller.
If a fault occurs, as indicated by the assertion of the $\overline{\text { Faultx }}$ signal (where $\overline{\text { Faultx }}=\overline{\text { Pwrfltx }}$ and $\overline{\text { Auxfltx }}$ ), the RESETx pin asynchronously asserts and the CLKENx and BUSENx pins are asynchronously deasserted.

## APPLICATION INFORMATION



UDG-01013
Figure 4. Serial Output Data Timing

## APPLICATION INFORMATION

## package information

The TPS2340A dissipates very little heat under normal operating conditions. However, when the TPS2340A is used with a load that exceeds PCI specifications either due to a fault or a specification violation, power dissipation increases as the square of load current. To allow reliable operation with high power dissipation, the TPS2340A is packaged in a PowerPAD ${ }^{m M}$ molded quad flat pack with heat-conducting tab on the underside. This package offers thermal resistance from junction to case of approximately $8^{\circ} \mathrm{C} / \mathrm{W}$. In a typical layout, thermal resistance from junction to ambient is approximately $35^{\circ} \mathrm{C} / \mathrm{W}$.

For optimum heat conduction and system reliability, the heat-conducting tab should be soldered directly to a $10-\mathrm{mm}$ square copper area on the circuit board tied to M12VIN voltage potential. To increase heat conduction, add plated vias to the copper area directly under the part and connect these vias to copper planes around $100 \%$ of their perimeter. These vias conduct heat away from the top layer of the circuit board and into other layers and should not use thermal reliefs (also known as thermals or webs). To prevent solder wicking into the vias, the vias should either be drilled to a size small enough to allow complete filling of the vias during hole plating or should be covered by solder mask on the back of the circuit board.

The TPS2340A TQFP-80 package conforms to JEDEC MS-026. For more information on the PowerPAD package, consult TI PowerPAD Technical Brief (SLMA002).

## device testability

Table 11. Test Mode Configuration

| PGOOD | TEST | SODO/MODE1 | SIDO/MODE0 | OPERATING MODE |
| :---: | :---: | :---: | :---: | :--- |
| $\uparrow$ | 0 | MODE1 | MODE0 | Normal operation mode, latch MODE inputs |
| 1 | 0 | SODO | SIDO | Normal operation mode, drive SODO and SIDO |
| $\uparrow$ | 1 | 0 | 0 | NAND tree test mode |
| $\uparrow$ | 1 | 0 | 1 | Tri-state test mode (all pins tri-stated) |
| $\uparrow$ | 1 | 1 | X | Reserved |
| 1 | 1 | SODO | SIDO | Normal operation, but Pwrenx driven on $\overline{\text { PWRLEDx }}, \overline{\text { Faultx }}$ <br> driven on $\overline{\text { ATTLEDx. }}$ |

NOTE: X = Don't care, $\mathrm{x}=$ slot A or B; shaded cells signify test modes.
When TEST is asserted after PGOOD is asserted, the device enters a run-time test mode. When this test mode is enabled, the slot-specific PWRLEDx output asserts when the slot's internal Pwrenx signal is asserted. Similarly, the $\overline{\text { ATTLEDx }}$ output asserts when the corresponding ANDed slot Faultx signal is asserted.

## APPLICATION INFORMATION

## NAND tree test

All bi-directional pins are tri-stated for input mode during the NAND Tree test except SIDO. All inputs except TEST should be forced low then forced high, one signal at a time, in the order listed below. The TEST pin is the last signal in the chain and is forced high then low. The NAND result is driven on the SIDO output pin as shown in the timing diagram in Figure 5.

Table 12. NAND Tree Order

|  | NAND TREE ORDER | PIN NO. |
| :---: | :--- | :---: |
| 1 | SIDI | 20 |
| 2 | SODO | 19 |
| 3 | SWA | 46 |
| 4 | SWB | 14 |
| 5 | $\overline{\text { PRSNT1A }}$ | 37 |
| 6 | $\overline{\text { PRSNT1B }}$ | 24 |
| 7 | $\overline{\text { PRSNT2A }}$ | 36 |
| 8 | $\overline{\text { PRSNT2B }}$ | 25 |
| 9 | $\overline{\text { BUTTONA }}$ | 45 |
| 10 | $\overline{\text { BUTTONB }}$ | 15 |
| 11 | M66ENA | 34 |
| 12 | M66ENB | 27 |
| 13 | PCI-XCAPA | 50 |
| 14 | PCI-XCAPB | 11 |
| 15 | SODI | 42 |
| 16 | SOC | 17 |
| 17 | $\overline{\text { SOR }}$ | 40 |
| 18 | $\overline{\text { SORR }}$ | 21 |
| 19 | SOLC | 39 |
| 20 | SORLC | 22 |
| 21 | $\overline{\text { BUSENA }}$ | 32 |
| 22 | $\overline{\text { BUSENB }}$ | 29 |
| 23 | $\overline{\text { CLKENA }}$ | 33 |
| 24 | $\overline{\text { CLKENB }}$ | 28 |
| 25 | $\overline{\text { RESETA }}$ | 35 |
| 26 | $\overline{\text { RESETB }}$ | 26 |
| 27 | $\overline{\text { PWRLEDA }}$ | 48 |
| 28 | $\overline{\text { PWRLEDB }}$ | 12 |
| 29 | $\overline{\text { ATTLEDA }}$ | 47 |
| 30 | $\overline{\text { ATTLEDB }}$ | 13 |
| 31 | $\overline{\text { SIL }}$ | 23 |
| 32 | SIC |  |
| 33 | TEST | 24 |
|  |  | 23 |



Table 13. NAND Tree Order

| OUTPUT PIN | POWER-ON <br> STATE |
| :--- | :---: |
| $\overline{\text { PWRLEDx }}$ | 1 |
| $\overline{\text { ATTLEDx }}$ | 1 |
| $\overline{\text { BUSENx }}$ | 1 |
| $\overline{\text { CLKENx }}$ | 1 |
| $\overline{\text { RESETx }}$ | 0 |

The TPS2340A supports tri-state and NAND-tree test functions.

## package dimensions

This package has an exposed copper heat-conducting pad on the bottom of the mold. This pad is approximately $6 \mathrm{~mm}^{2}$, with maximum dimensions $6.07 \mathrm{~mm} \times 6.07 \mathrm{~mm}$


Figure 6. PowerPAD Package Details Bottom View


NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body dimensions do not include mold flash or protrusion.
D. The package thermal performance may be enhanced by bonding the thermal pad to an external thermal plane. This pad is electrically and thermally connected to the backside of the die and possibly selected leads.
E. Falls within JEDEC MS-026

## PACKAGING INFORMATION

| Orderable Device | Status ${ }^{(1)}$ | Package <br> Type | Package <br> Drawing | Pins Package <br> Qty | Eco Plan ${ }^{(2)}$ | Lead/Ball Finish | MSL Peak Temp ${ }^{(3)}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TPS2340APFP | ACTIVE | HTQFP | PFP | 80 | 96 |  <br> no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR |
| TPS2340APFPG4 | ACTIVE | HTQFP | PFP | 80 | 96 |  <br> no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR |
| TPS2340APFPR | ACTIVE | HTQFP | PFP | 80 | 1000 |  <br> no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR |
| TPS2340APFPRG4 | ACTIVE | HTQFP | PFP | 80 | 1000 |  <br> no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR |

${ }^{(1)}$ The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but Tl does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.
${ }^{(2)}$ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS \& no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb -Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed $0.1 \%$ by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Green (RoHS \& no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants ( Br or Sb do not exceed $0.1 \%$ by weight in homogeneous material)
${ }^{(3)}$ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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