

5-BIT PROGRAMMABLE SYNCHRONOUS BUCK CONTROLLER IC WITH TRIPLE LDO CONTROLLER

FEATURES

- Designed to meet VRM 9.0 specification for next generation microprocessors
- On-Board 5-Bit DAC programs the output voltage from 1.075V to 1.850V in 25mV steps
- Linear Regulator Controller On-Board for 1.8V
- Provides single chip solution for Vcore, GTL+, AGP bus, and 1.8V
- Automatic Voltage Selection for AGP Slot VDDQ Supply
- Linear Regulator Controller On-Board for 1.5V GTL+ Supply
- Loss-less Short Circuit Protection for all Outputs
- Synchronous operation allows maximum efficiency
- Patented architecture allows fixed frequency operation as well as 100% duty cycle during dynamic load
- Minimum Part Count
- Soft-Start
- High current totem pole driver for direct driving of the external power MOSFET
- Power Good function monitors all outputs
- Over-Voltage Protection Circuitry protects the switcher output and generates a fault output

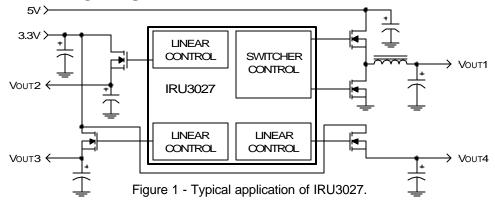
APPLICATIONS

- Total power solution for next generation Intel processor application
- AMD K7 Low Cost Solution

DESCRIPTION

The IRU3027 controller IC is specifically designed to meet VRM 9.0 specification for next generation microprocessor applications requiring multiple on-board regulators. The IRU3027 provides a single chip controller IC for the Vcore, three LDO controllers, one with an automatic select pin that connects to the Type Detect pin of the AGP slot for the AGP VDDQ supply, one for GTL+ and the other for the 1.8V chip set regulator as required for the next generation PC applications. The IRU3027 is designed to use either bipolar transistors for Vout3(1.5V) and Vout4(1.8V). No external resistor divider is necessary for any of the regulators. The switching regulator features a patented topology that in combination with a few external components as shown in the typical application circuit, will provide well in excess of 20A of output current for an on-board DC-DC converter while automatically providing the right output voltage via the 5-bit internal DAC. The IRU3027 also features loss-less current sensing for both switchers by using the RDS(ON) of the high side power MOSFET as the sensing resistor, an output under-voltage shutdown that detects short circuit condition for the linear outputs and latches the system off, and a Power Good window comparator that switches its open collector output low when any one of the outputs is outside of a pre-programmed window.

TYPICAL APPLICATION

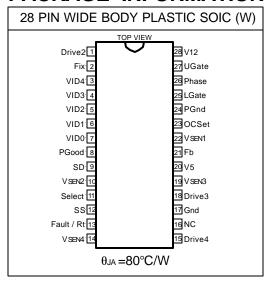


PACKAGE ORDER INFORMATION

T _A (°C)	DEVICE	PACKAGE
0 To 70	IRU3027CW	28-Pin Plastic SOIC WB

ABSOLUTE MAXIMUM RATINGS

PACKAGE INFORMATION



ELECTRICAL SPECIFICATIONS

Unless otherwise specified, these specifications apply over V12=12V, V5=5V and T_A =0 to 70°C. Typical values refer to T_A =25°C. Low duty cycle pulse testing is used which keeps junction and case temperatures equal to the ambient temperature.

PARAMETER	SYM	TEST CONDITION	MIN	TYP	MAX	UNITS
Supply UVLO Section						
UVLO Threshold-12V		Supply Ramping Up		10		V
UVLO Hysteresis-12V				0.6		V
UVLO Threshold-5V		Supply Ramping Up		4.4		V
UVLO Hysteresis-5V				0.3		V
Supply Current						
Operating Supply Current		V12		6		mA
		V5		30		
Switching Controllers; Vcore (V SEN 1) a	nd AGP (Vsen2)				
VID Section (Vcore only)						
DAC Output Voltage (Note 1)			0.99Vs	Vs	1.01Vs	V
DAC Output Line Regulation				0.1		%
DAC Output Temp Variation				0.5		%
VID Input LO					8.0	٧
VID Input HI			2			V
VID Input Internal Pull-Up				27		KΩ
Resistor to V5						
Vsen2 Voltage		Select <0.8V		1.5		V
		Select >2V		3.3		

International **IOR** Rectifier

Error Comparator Section 2	PARAMETER	SYM	TEST CONDITION	MIN	TYP	MAX	UNITS
Input Disas Current	Error Comparator Section						
Input Offset Voltage	-					2	μA
Current Limit Section 200 μA CS Tomp Offset Voltage -5 +5 mV Hiccup Duty Cycle Css=0.1μF 10 % Output Drivers Section Rise Time C.=3000pF 70 ns Fall Time C.=3000pF 70 ns Dead Band Time Between High Side and Synch Drive (Vcore Switcher Only) 200 ns High Side and Synch Drive (Vcore Switcher Only) Rt=Open 217 KHz Osc Frequency Rt=Open 217 KHz 1.80 Sec Frequency Rt=Open 217 KHz 1.80 Vosa Voltage Vo4 Tx=25°C, Drive4=Vssv4 1.800 V Vssv Voltage Vo4 Tx=25°C, Drive4=Vssv4 1.800 V Vssv Voltage Vo3 Tx=25°C, Drive3=Vssv3 1.500	_ •			-2			
Current Limit Section 200 μΑ CS Tomp Offset Vollage -5 +5 mV Hiccup Duty Cycle Css=0.1μF 10 % Output Drivers Section Rise Time C.=3000pF 70 ns Fall Time C.=3000pF 70 ns Dead Band Time Between C.=3000pF 70 ns High Side and Synch Drive (Voors Switcher Only) Voor Switcher Only) 200 ns Oscillator Section (Internal) Scs Frequency Rt=Open 217 KHz Osc Frequency Rt=Open 217 KHz 1.800 V Osc Frequency Rt=Open 217 KHz 1.800 V Osc Frequency Rt=Open 217 KHz 1.800 V VssN Voltage Voltage 1.800 V V 1.800 V Input Bias Current 50 mA 1.500 V V 1.500 V V VssN Voltage 1.500 V Input Bias Current 50 <td></td> <td></td> <td>V_{DIFF}=10mV</td> <td></td> <td></td> <td>100</td> <td>ns</td>			V _{DIFF} =10mV			100	ns
CS Comp Offset Voltage -5 +5 mV Hiccup Duty Cycle Css=0.1μF 10 % Output Drivers Section Css=0.1μF 10 % Rise Time Css=0.000pF 70 ns Fall Time Css=0.000pF 70 ns Dead Band Time Between Css=0.000pF 200 ns High Side and Synch Drive (Vcore Switcher Only) Css=0.00pF 200 ns Osc Frequency Rt=Open 217 KHz 1.87 Regulator (Vsex4) Sex Voltage 1.800 V Vsen Voltage 1.800 V Instance V Vsen Voltage 1.500 V Instance Instance Instance Instance Instance V Vsen Voltage Vo3 Tx=25°C, Drive3=Vsen3 1.500 V							
CS Comp Offset Voltage -5 +5 mV Hiccup Duty Cycle Css=0.1μF 10 % Output Drivers Section Css=0.1μF 10 % Rise Time Css=0.000pF 70 ns Fall Time Css=0.000pF 70 ns Dead Band Time Between Css=0.000pF 200 ns High Side and Synch Drive (Vcore Switcher Only) Css=0.00pF 200 ns Osc Frequency Rt=Open 217 KHz 1.87 Regulator (Vsex4) Sex Voltage 1.800 V Vsen Voltage 1.800 V Instance V Vsen Voltage 1.500 V Instance Instance Instance Instance Instance V Vsen Voltage Vo3 Tx=25°C, Drive3=Vsen3 1.500 V					200		μA
Hiccup Duty Cycle				-5		+5	
Output Drivers Section CL=3000pF 70 ns Rise Time CL=3000pF 70 ns Dead Band Time Between CL=3000pF 200 ns High Side and Synch Drive (Vcore Switcher Only) Ns Notage 200 ns High Side and Synch Drive Rt=Open 217 KHz Ns Notage 217 KHz KHz Ns Notage Vod Ts=25°C, Drive4=Vssv4 1.800 V V Vssv Voltage Vod Ts=25°C, Drive4=Vssv4 1.800 V V Vssv Voltage Vod Ts=25°C, Drive4=Vssv4 1.800 V V Vssv Voltage Vod Ts=25°C, Drive4=Vssv4 1.800 V V Nsv Voltage 1.500 V Ts Ts Nsv Voltage 1.500 V Nsv Voltage Vssv Voltage Vssv Voltage Vssv Voltage Vssv Voltage Vssv Voltage	· · · · · · · · · · · · · · · · · · ·		Css=0.1µF		10		%
Rise Time	<u> </u>						
Fall Time	=		CL=3000pF		70		ns
Dead Band Time Between High Side and Synch Drive (Vcore Switcher Only)	Fall Time				70		ns
High Side and Synch Drive (Vcore Switcher Only)	Dead Band Time Between		·		200		ns
(Vcore Switcher Only) Oscillator Section (Internal) Osc Frequency Rt=Open 217 KHz	High Side and Synch Drive						
Oscillator Section (Internal) Rt=Open 217 KHz 1.8V Regulator (Vsex4) Vsex Ovitage Vo4 Tx=25°C, Drive4=Vsex4 1.800 V Vsex Voltage 1.800 V Incompage Vsex Vsex Vsex Vsex Vsex Vsex Vsex Vse	•						
1.8V Regulator (Vsen4) Vo4 TA=25°C, Drive4=Vsen4 1.800 V Vsen Voltage 1.800 V Input Bias Current 2 μA Output Drive Current 50 mA 1.5V Regulator (Vsen3) Vsen Voltage Vo3 Ta=25°C, Drive3=Vsen3 1.500 V Vsen Voltage Vo3 Ta=25°C, Drive3=Vsen3 1.500 V Vsen Voltage 1.500 V V Usen Voltage 1.500 V Usen Good Section Vsen Ramping Up 0.920 section V	<u>, </u>						
1.80	Osc Frequency		Rt=Open		217		KHz
VSEN Voltage Vo4 Ta=25°C, Drive4=Vsen4 1.800 V VSEN Voltage 1.800 V Input Bias Current 2 μA Output Drive Current 50 mA 1.5V Regulator (Vsen3) V Ta=25°C, Drive3=Vsen3 1.500 V Vsen Voltage 1.500 V V Vsen Voltage 1.500 V Input Bias Current 2 μA Output Drive Current 50 mA Power Good Section Vsen1 Ramping Down 0.90Vs V Vsen1 UV Lower Trip Point Vsen1 Ramping Up 0.92Vs V Vsen1 HV Upper Trip Point Vsen1 Ramping Up 1.10Vs V Vsen1 HV Upper Trip Point Vsen1 Ramping Down 1.08Vs V Vsen1 HV Upper Trip Point Vsen1 Ramping Down 1.08Vs V Vsen1 HV Hysterises 0.02Vs V Vsen2 Trip Point Select <0.8V							
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Vose	Output Drive Current			50			
Vase Voltage	1.5V Regulator (V sen 3)						
Vase Voltage	Vsen Voltage	Vo3	Ta=25°C, Drive3=Vsen3		1.500		V
Output Drive Current 50 mA Power Good Section Vsen1 UV Lower Trip Point Vsen1 Ramping Down 0.90Vs V Vsen1 UV Lower Trip Point Vsen1 Ramping Up 0.92Vs V Vsen1 UV Hysterises 0.02Vs V Vsen1 HV Upper Trip Point Vsen1 Ramping Up 1.10Vs V Vsen1 HV Lower Trip Point Vsen1 Ramping Down 1.08Vs V Vsen2 Trip Point Select <0.8V					1.500		V
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VSEN2 Trip Point Select <0.8V Select >2V 1.100 V VSEN3 Trip Point Fix=Gnd Fix=Open 0.920 V VSEN4 Trip Point Fix=Gnd Fix=Open 0.920 V Power Good Output LO RL=3mA 0.4 V Power Good Output HI RL=5K, Pull-Up to 5V 4.8 V Fault (Overvoltage) Section Core OV Upper Trip Point VSEN1 Ramping Up 1.17Vs V Core OV Lower Trip Point VSEN1 Ramping Down 1.15Vs V Fault Output HI Io=3mA 10 V	Vsen1 HV Lower Trip Point		V _{SEN} 1 Ramping Down		1.08Vs		V
Select >2V 2.560	V _{SEN} 1 HV Hysterises				0.02Vs		V
VSEN3 Trip Point Fix=Gnd	Vsen2 Trip Point		Select <0.8V		1.100		V
Fix=Open 1.320			Select >2V		2.560		
Vsen4 Trip Point Fix=Gnd Fix=Open 0.920 V V Power Good Output LO RL=3mA 0.4 V Power Good Output HI RL=5K, Pull-Up to 5V 4.8 V Fault (Overvoltage) Section Vsen1 Ramping Up 1.17Vs V Core OV Upper Trip Point Vsen1 Ramping Up 1.15Vs V Core OV Lower Trip Point Vsen1 Ramping Down 1.15Vs V Fault Output HI Io=3mA 10 V Soft-Start Section 10 V	Vsen3 Trip Point		Fix=Gnd		0.920		V
Fix=Open 1.140			Fix=Open		1.320		
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Power Good Output HI RL=5K, Pull-Up to 5V 4.8 V Fault (Overvoltage) Section Vsen1 Ramping Up 1.17Vs V Core OV Upper Trip Point Vsen1 Ramping Up 1.15Vs V Core OV Lower Trip Point Vsen1 Ramping Down 1.15Vs V Fault Output HI Io=3mA 10 V Soft-Start Section 10 V			Fix=Open		1.140		
Fault (Overvoltage) Section Core OV Upper Trip Point Vsen1 Ramping Up 1.17Vs V Core OV Lower Trip Point Vsen1 Ramping Down 1.15Vs V Fault Output HI Io=3mA 10 V Soft-Start Section 0 0 0	Power Good Output LO		RL=3mA		0.4		V
Core OV Upper Trip Point Vsen1 Ramping Up 1.17Vs V Core OV Lower Trip Point Vsen1 Ramping Down 1.15Vs V Fault Output HI Io=3mA 10 V Soft-Start Section 0 0 0			RL=5K, Pull-Up to 5V		4.8		V
Core OV Lower Trip Point Vsen1 Ramping Down 1.15Vs V Fault Output HI Io=3mA 10 V Soft-Start Section							
Fault Output HI Io=3mA 10 V Soft-Start Section							
Soft-Start Section					1.15Vs		
			Io=3mA		10		V
Pull-Up Resistor to 5V OCSet=0V, Phase=5V 20 μA							
	Pull-Up Resistor to 5V		OCSet=0V, Phase=5V		20		μΑ

Note 1: Vs refers to the set point voltage given in Table 1.

D4	D3	D2	D1	D0	Vs
1	1	1	1	1	1.075
1	1	1	1	0	1.100
1	1	1	0	1	1.125
1	1	1	0	0	1.150
1	1	0	1	1	1.175
1	1	0	1	0	1.200
1	1	0	0	1	1.225
1	1	0	0	0	1.250
1	0	1	1	1	1.275
1	0	1	1	0	1.300
1	0	1	0	1	1.325
1	0	1	0	0	1.350
1	0	0	1	1	1.375
1	0	0	1	0	1.400
1	0	0	0	1	1.425
1	0	0	0	0	1.450

D4	D3	D2	D1	D0	Vs
0	1	1	1	1	1.475
0	1	1	1	0	1.500
0	1	1	0	1	1.525
0	1	1	0	0	1.550
0	1	0	1	1	1.575
0	1	0	1	0	1.600
0	1	0	0	1	1.625
0	1	0	0	0	1.650
0	0	1	1	1	1.675
0	0	1	1	0	1.700
0	0	1	0	1	1.725
0	0	1	0	0	1.750
0	0	0	1	1	1.775
0	0	0	1	0	1.800
0	0	0	0	1	1.825
0	0	0	0	0	1.850

Table 1 - Set point voltage vs. VID codes.

PIN DESCRIPTIONS

PIN#	PIN SYMBOL	PIN DESCRIPTION
1	Drive2	This pin controls the gate of an external MOSFET for the AGP linear regulator.
2	Fix	Leaving this pin open provides fixed output voltages of the 1.5V and 1.8V for the #3 and
		#4 linear regulators. When this pin is grounded the reference to the linear regulators are
		set to 1.26V and therefore the output of the regulators can be programmed to any volt-
		ages above the 1.26V using: $V_{OUT} = 1.26 \times (1 + R_{TOP}/R_{BOT})$
		Where:
		RTOP=Top resistor connected from the output to the VSENSE pin
		RBOT=Bottom resistor connected from the VSENSE pin to ground.
3	VID4	This pin selects a range of output voltages for the DAC. When in the LOW state the
		range is 1.3V to 2.05V and when it switches to HI state the range is 2V to 3.5V. This pin
		is TTL compatible that realizes a logic "1" as either HI or Open. When left open, this pin
		is pulled up internally by a 27K Ω resistor to 5V supply.
4	VID3	MSB input to the DAC that programs the output voltage. This pin is TTL compatible that
		realizes a logic "1" as either HI or Open. When left open, this pin is pulled up internally by
		a 27K Ω resistor to 5V supply.
5	VID2	Input to the DAC that programs the output voltage. This pin is TTL compatible that real-
		izes a logic "1" as either HI or Open. When left open, this pin is pulled up internally by a
		27K Ω resistor to 5V supply.
6	VID1	Input to the DAC that programs the output voltage. This pin is TTL compatible that real-
		izes a logic "1" as either HI or Open. When left open, this pin is pulled up internally by a
		27K Ω resistor to 5V supply.
7	VID0	LSB input to the DAC that programs the output voltage. This pin is TTL compatible that
		realizes a logic "1" as either HI or Open. When left open, this pin is pulled up internally by
		a 27K Ω resistor to 5V supply.
8	PGood	This pin is an open collector output that switches LO when any of the outputs are outside
		of the specified under-voltage trip point. It also switches low when V _{SEN} 1 pin is more than
		10% above the DAC voltage setting.



PIN#	PIN SYMBOL	PIN DESCRIPTION
9	SD	This pin provides shutdown for all the regulators. A TTL compatible, logic level high applied
		to this pin disables all the outputs and discharges the soft-start capacitor. The SD signal
		turns off the synchronous allowing the body diode to conduct and discharge the output
		capacitor.
10	Vsen2	This pin provides the feedback for the AGP linear regulator. The Select pin when con-
		nected to the "Type Detect" pin of the AGP slot automatically selects the right voltage for
	0.1.	the AGP VDDQ.
11	Select	This pin provides automatic voltage selection for the AGP switching regulator. When it is pulled LO, the voltage is 1.5V and when left open or pulled to HI, the voltage is 3.3V.
12	SS	This pin provides the soft-start for all the regulators. An internal current source charges an
		external capacitor that is connected from this pin to ground which ramps up the outputs of
		the regulators, preventing the outputs from overshooting as well as limiting the input cur-
		rent. The second function of the Soft-Start cap is to provide long off time (HICCUP) for the
	E 1/5/	synchronous MOSFET during current limiting.
13	Fault / Rt	This pin has dual function. It acts as an output of the over-voltage protection circuitry or it
		can be used to program the frequency using an external resistor. When used as a fault
		detector, if any of the switcher outputs exceed the OVP trip point, the Fault pin switches
		to 12V and the soft-start cap is discharged. If the Fault pin is to be connected to any
14	Vsen4	external circuitry, it needs to be buffered. This pin provides the feedback for the linear regulator that its output drive is Drive4.
15	Drive4	This pin controls the gate of an external MOSFET for the 1.8V chip set linear regulator.
16	NC	This pin has no connection.
17	Gnd	This pin reas no connection. This pin serves as the ground pin and must be connected directly to the ground plane.
18	Drive3	This pin controls the gate of an external transistor for the 1.5V GTL+ linear regulator.
19	V _{SEN} 3	This pin provides the feedback for the linear regulator that its output drive is Drive3.
20	V5	5V supply voltage. A high frequency capacitor (0.1 to 1μF) must be placed close to this
		pin and connected from this pin to the ground plane for noise free operation.
21	Fb	This pin provides the feedback for the synchronous switching regulator. Typically this pin
		can be connected directly to the output of the switching regulator. However, a resistor
		divider is recommended to be connected from this pin to Vout1 and ground to adjust the
		output voltage for any drop in the output voltage that is caused by the trace resistance.
		The value of the resistor connected from Vout1 to Fb1 must be less than 1000Ω .
22	Vsen1	This pin is internally connected to the under-voltage and over-voltage comparators sens-
		ing the Vcore status. It must be connected directly to the Vcore supply.
23	OCSet	This pin is connected to the Drain of the power MOSFET of the Core supply and it provides
		the positive sensing for the internal current sensing circuitry. An external resistor pro-
		grams the current sense threshold depending on the Ros of the power MOSFET. An
		external capacitor is placed in parallel with the programming resistor to provide high fre-
24	PGnd	quency noise filtering. This pin serves as the Power ground pin and must be connected directly to the ground
24	rGilu	plane close to the source of the synchronous MOSFET. A high frequency capacitor (typi-
		cally 1µF) must be connected from V12 pin to this pin for noise free operation.
25	LGate	Output driver for the synchronous power MOSFET for the Core supply.
26	Phase	This pin is connected to the Source of the power MOSFET for the Core supply and it
20	1 11400	provides the negative sensing for the internal current sensing circuitry.
27	UGate	Output driver for the high side power MOSFET for the Core supply.
28	V12	This pin is connected to the 12V supply and serves as the power Vcc pin for the output
	. —	drivers. A high frequency capacitor (typically 1µF) must be placed close to this pin and
		PGnd pin and be connected directly from this pin to the ground plane for the noise free
		operation.

BLOCK DIAGRAM

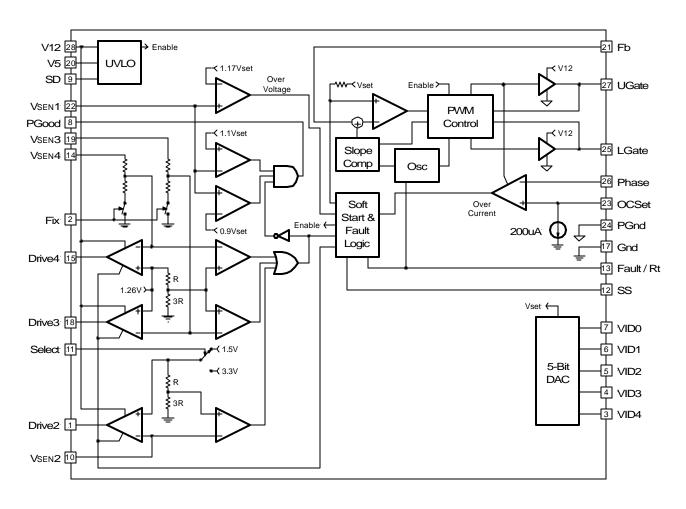


Figure 2 - Simplified block diagram of the IRU3027.

TYPICAL APPLICATION

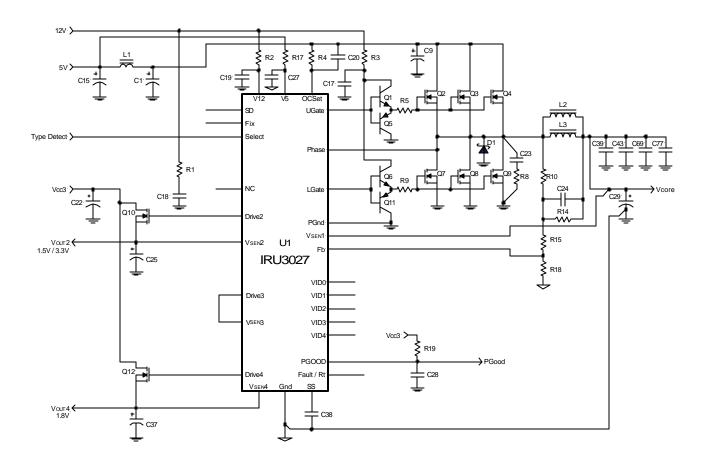


Figure 3 - Typical application of IRU3027 for the AMD Slot A socket.



IRU3027 APPLICATION PARTS LIST

Dual Layout with HIP6019

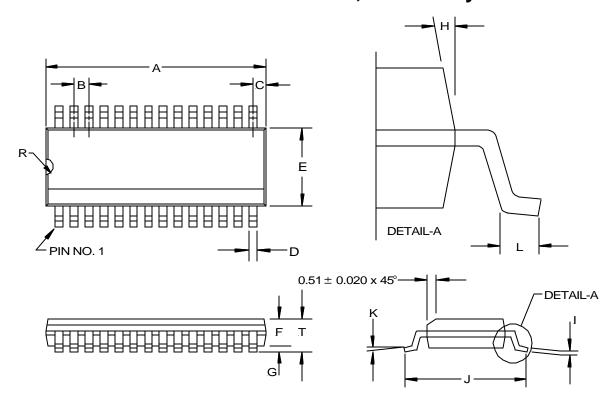
Ref Desig	Description	Qty	Part #	Manuf
Q1, 6	Transistor	2	2SD882, TO-226 package	Fairchild
Q2, 3, 4	MOSFET	3	IRF3706S, TO-263 package	IR
Q5, 11	Transistor	2	2SB772, TO-226 package	Fairchild
Q7, 8, 9	MOSFET	3	IRL2203NS, TO-263 package	IR
Q10	MOSFET	1	IRLR3103S, TO-252 package	IR
Q12	MOSFET	1	IRLR024, TO-252 package	IR
D1	Diode	1	MBR1535CT, TO-220 package	IR
L1	Inductor	1	L=1μH, 5052B core with 5 turns of triple 0.8mm wire	Micro Metal
L2, 3	Inductor	2	L=1.8μH, 6018 core with 6 turns of triple 0.8mm wire	Micro Metal
C1	Capacitor, Ceramic	8	1μF, 0603	
C9	Capacitor, Electrolytic	6	10MV1500GX, 1500μF, 10V	Sanyo
C15	Capacitor, Electrolytic	1	10MV1500GX, 1500μF, 10V	Sanyo
C16	Capacitor, Ceramic	1	1μF, 0603	,
C17, 18,	Capacitor, Ceramic	4	1μF, 0805	
19, 21	·		·	
C20	Capacitor, Ceramic	1	220pF, 0603	
C22, 26	Capacitor, Electrolytic	2	6MV1000GX, 1000μF, 6.3V	Sanyo
C23	Capacitor, Ceramic	1	1000pF, 0805	
C24, 27	Capacitor, Ceramic	2	1μF, 0603	
C25, 37	Capacitor, Electrolytic	2	6MV1500GX, 1500μF, 6.3V	Sanyo
C28, 38	Capacitor, Ceramic	2	0.1μF, 0603	
C29	Capacitor, Electrolytic	8	6MV2200GX, 2200μF, 6.3V	Sanyo
C39	Capacitor, Ceramic	4	4.7μF, 0805	
C43	Capacitor, Ceramic	26	1μF, 0603	
C69	Capacitor, Ceramic	8	0.01μF, 0603	
C77	Capacitor, Ceramic	8	39pF, 0603	
R1, 2, 3,	Resistor	6	10Ω, 5%, 0603	
7, 16, 21				
R4	Resistor	1	2KΩ, 5%, 0603	
R5, 9	Resistor	4	1Ω, 5%, 0805	
R8	Resistor	1	4.7Ω, 5%, 0805	
R10, 14	Resistor	2	3.3KΩ, 1%, 0603	
R11, 20	Resistor	2	0Ω, 0603	
R12	Resistor	1	47KΩ, 5%, 0603	
R15	Resistor	1	2.2KΩ, 1%, 0603	
R17	Resistor	1	1Ω, 0603	
R18	Resistor	1	100ΚΩ, 1%, 0603	
R19	Resistor	1	10KΩ, 5%, 0603	



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(W) SOIC Package 28-Pin Surface Mount, Wide Body

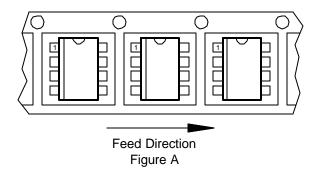


SYMBOL	28-PIN		
	MIN	MAX	
Α	17.73	17.93	
В	1.27	BSC	
С	0.66	REF	
D	0.36	0.46	
Е	7.40	7.60	
F	2.44	2.64	
G	0.10	0.30	
I	0.23	0.32	
J	10.11	10.51	
K	0°	8°	
L	0.51	1.01	
R	0.63	0.89	
T	2.44	2.64	

NOTE: ALL MEASUREMENTS ARE IN MILLIMETERS.

PACKAGE SHIPMENT METHOD

PKG	PACKAGE	PIN	PARTS	PARTS	T & R
DESIG	DESCRIPTION	COUNT	PER TUBE	PER REEL	Orientation
W	SOIC, Wide Body	28	27	1000	Fig A





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