



# P89LPC9381

8-bit microcontroller with accelerated two-clock 80C51 core  
4 kB 3 V byte-erasable flash with 10-bit ADC

Rev. 01 — 8 September 2006

Product data sheet

## 1. General description

---

The P89LPC9381 is a single-chip microcontroller, available in low-cost packages, based on a high performance processor architecture that executes instructions in two to four clocks, six times the rate of standard 80C51 devices. Many system-level functions have been incorporated into the P89LPC9381 in order to reduce component count, board space, and system cost.

## 2. Features

---

### 2.1 Principal features

- 4 kB byte-erasable flash code memory organized into 1 kB sectors and 64 B pages. Single-byte erasing allows any byte(s) to be used as non-volatile data storage.
- 256 B RAM data memory on-chip RAM.
- 8-input multiplexed 10-bit ADC. Two analog comparators with selectable inputs and reference source.
- Two 16-bit counter/timers (each may be configured to toggle a port output upon timer overflow or to become a PWM output) and a 23-bit system timer that can also be used as a RTC.
- Enhanced UART with fractional baud rate generator, break detect, framing error detection, and automatic address detection; 400 kHz byte-wide I<sup>2</sup>C-bus communication port and SPI communication port.
- High-accuracy internal RC oscillator option allows operation without external oscillator components. The RC oscillator option is selectable and fine tunable.
- 2.4 V to 3.6 V  $V_{DD}$  operating range. I/O pins are 5 V tolerant (may be pulled up or driven to 5.5 V).
- 28-pin TSSOP package with 23 I/O pins minimum and up to 26 I/O pins while using on-chip oscillator and reset options.

**PHILIPS**

## 2.2 Additional features

- A high performance 80C51 CPU provides instruction cycle times of 111 ns to 222 ns for all instructions except multiply and divide when executing at 18 MHz. This is six times the performance of the standard 80C51 running at the same clock frequency. A lower clock frequency for the same performance results in power savings and reduced EMI.
- Serial flash ICP allows simple production coding with commercial EPROM programmers. Flash security bits prevent reading of sensitive application programs.
- Serial flash ISP allows coding while the device is mounted in the end application.
- In-Application Programming of the flash code memory. This allows changing the code in a running application.
- Watchdog timer with separate on-chip oscillator, requiring no external components. The watchdog prescaler is selectable from eight values.
- Low voltage reset (brownout detect) allows a graceful system shutdown when power fails. May optionally be configured as an interrupt.
- Idle and two different power-down reduced power modes. Improved wake-up from Power-down mode (a LOW interrupt input starts execution). Typical power-down current is 1  $\mu$ A (total power-down with voltage comparators disabled).
- Active-LOW reset. On-chip power-on reset allows operation without external reset components. A reset counter and reset glitch suppression circuitry prevent spurious and incomplete resets. A software reset function is also available.
- Configurable on-chip oscillator with frequency range options selected by user programmed flash configuration bits. Oscillator options support frequencies from 20 kHz to the maximum operating frequency of 18 MHz.
- Oscillator fail detect. The watchdog timer has a separate fully on-chip oscillator allowing it to perform an oscillator fail detect function.
- Programmable port output configuration options: quasi-bidirectional, open-drain, push-pull, input-only.
- Port 'input pattern match' detect. Port 0 may generate an interrupt when the value of the pins match or do not match a programmable pattern.
- LED drive capability (20 mA) on all port pins. A maximum limit is specified for the entire chip.
- Controlled slew rate port outputs to reduce EMI. Outputs have approximately 10 ns minimum ramp times.
- Only power and ground connections are required to operate the P89LPC9381 when internal reset option is selected.
- Four interrupt priority levels.
- Eight keypad interrupt inputs, plus two additional external interrupt inputs.
- Schmitt trigger port inputs.
- Second data pointer.
- Emulation support.

### 3. Ordering information

Table 1. Ordering information

Type number	Package		
	Name	Description	Version
P89LPC9381FDH	TSSOP28	plastic thin shrink small outline package; 28 leads; body width 4.4 mm	SOT361-1

#### 3.1 Ordering options

Table 2. Ordering options

Type number	Flash memory	Temperature range	Frequency
P89LPC9381FDH	4 kB	-40 °C to +85 °C	0 MHz to 18 MHz

4. Block diagram

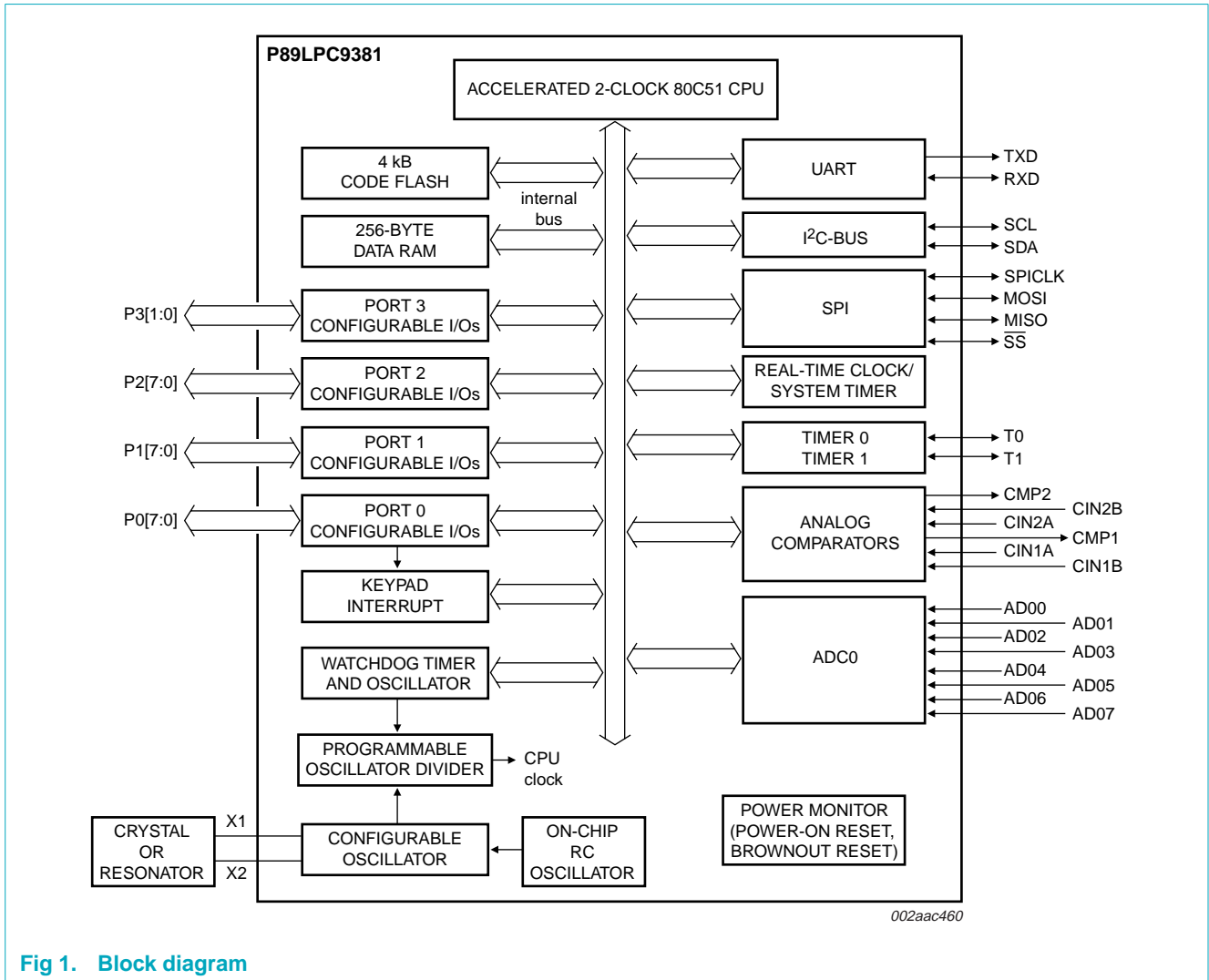
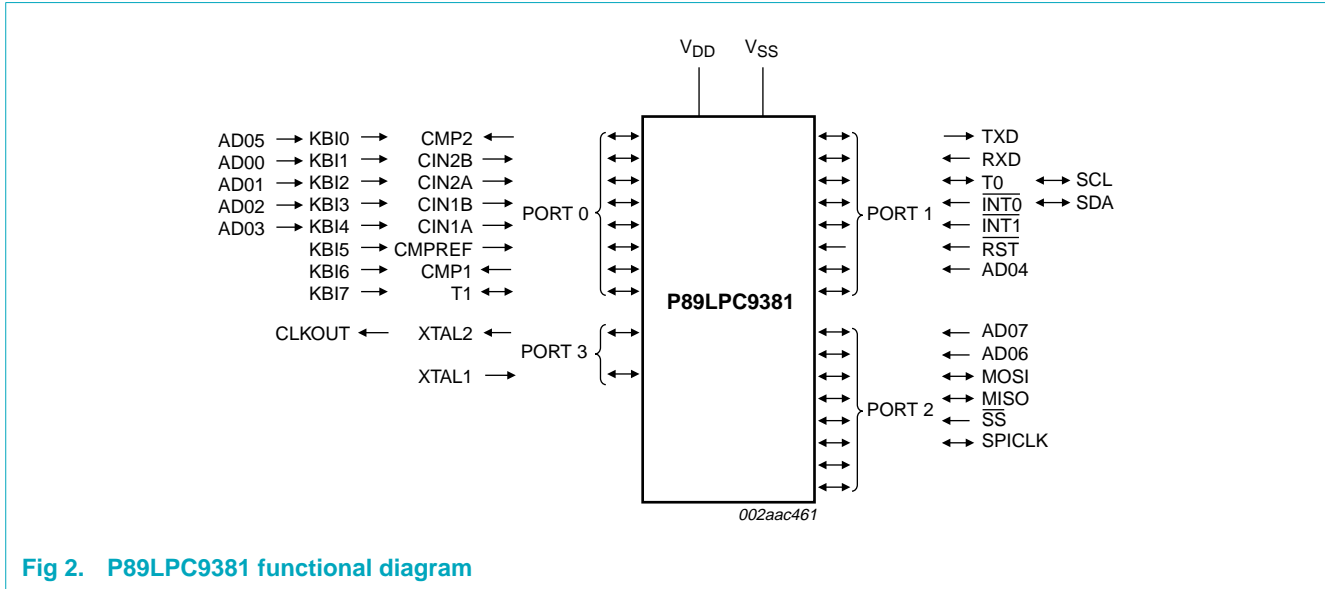


Fig 1. Block diagram

**5. Functional diagram**



**Fig 2. P89LPC9381 functional diagram**

## 6. Pinning information

### 6.1 Pinning

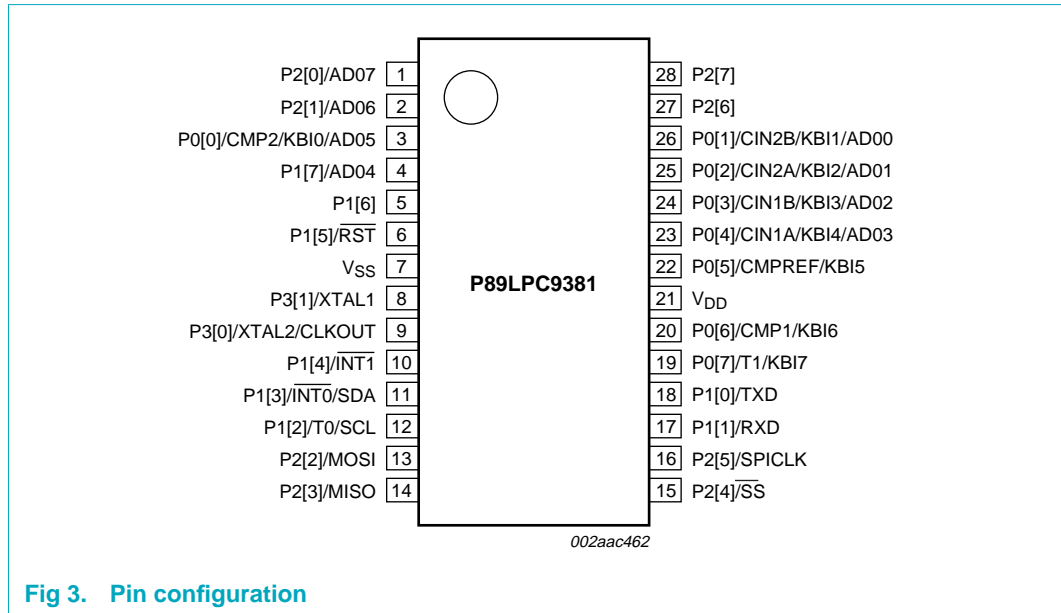


Fig 3. Pin configuration

### 6.2 Pin description

Table 3. Pin description

Symbol	Pin	Type	Description
P0[7:0]		I/O	<p><b>Port 0:</b> Port 0 is an 8-bit I/O port with a user-configurable output type. During reset Port 0 latches are configured in the input only mode with the internal pull-up disabled. The operation of Port 0 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to <a href="#">Section 7.13.1 “Port configurations”</a> and <a href="#">Table 10 “Static characteristics”</a> for details.</p> <p>The Keypad Interrupt feature operates with Port 0 pins.</p> <p>All pins have Schmitt triggered inputs.</p> <p>Port 0 also provides various special functions as described below:</p>
P0[0]/CMP2/ KBI0/AD05	3	I/O	<b>P0[0]</b> — Port 0 bit 0.
		O	<b>CMP2</b> — Comparator 2 output.
		I	<b>KBI0</b> — Keyboard input 0.
		I	<b>AD05</b> — ADC0 channel 5 analog input.
P0[1]/CIN2B/ KBI1/AD00	26	I/O	<b>P0[1]</b> — Port 0 bit 1.
		I	<b>CIN2B</b> — Comparator 2 positive input B.
		I	<b>KBI1</b> — Keyboard input 1.
		I	<b>AD00</b> — ADC0 channel 0 analog input.

Table 3. Pin description ...continued

Symbol	Pin	Type	Description
P0[2]/CIN2A/ KBI2/AD01	25	I/O	<b>P0[2]</b> — Port 0 bit 2.
		I	<b>CIN2A</b> — Comparator 2 positive input A.
		I	<b>KBI2</b> — Keyboard input 2.
		I	<b>AD01</b> — ADC0 channel 1 analog input.
P0[3]/CIN1B/ KBI3/AD02	24	I/O	<b>P0[3]</b> — Port 0 bit 3.
		I	<b>CIN1B</b> — Comparator 1 positive input B.
		I	<b>KBI3</b> — Keyboard input 3.
		I	<b>AD02</b> — ADC0 channel 2 analog input.
P0[4]/CIN1A/ KBI4/AD03	23	I/O	<b>P0[4]</b> — Port 0 bit 4.
		I	<b>CIN1A</b> — Comparator 1 positive input A.
		I	<b>KBI4</b> — Keyboard input 4.
		I	<b>AD03</b> — ADC0 channel 3 analog input.
P0[5]/CMPREF/ KBI5	22	I/O	<b>P0[5]</b> — Port 0 bit 5.
		I	<b>CMPREF</b> — Comparator reference (negative) input.
		I	<b>KBI5</b> — Keyboard input 5.
P0[6]/CMP1/ KBI6	20	I/O	<b>P0[6]</b> — Port 0 bit 6.
		O	<b>CMP1</b> — Comparator 1 output.
		I	<b>KBI6</b> — Keyboard input 6.
P0[7]/T1/KBI7	19	I/O	<b>P0[7]</b> — Port 0 bit 7.
		I/O	<b>T1</b> — Timer/counter 1 external count input or overflow output.
		I	<b>KBI7</b> — Keyboard input 7.
P1[7:0]		I/O, I <a href="#">[1]</a>	<p><b>Port 1:</b> Port 1 is an 8-bit I/O port with a user-configurable output type, except for three pins as noted below. During reset Port 1 latches are configured in the input only mode with the internal pull-up disabled. The operation of the configurable Port 1 pins as inputs and outputs depends upon the port configuration selected. Each of the configurable port pins are programmed independently. Refer to <a href="#">Section 7.13.1 “Port configurations”</a> and <a href="#">Table 10 “Static characteristics”</a> for details. P1[2] to P1[3] are open-drain when used as outputs. P1[5] is input only. All pins have Schmitt triggered inputs.</p> <p>Port 1 also provides various special functions as described below:</p>
P1[0]/TXD	18	I/O	<b>P1[0]</b> — Port 1 bit 0.
		O	<b>TXD</b> — Transmitter output for the serial port.
P1[1]/RXD	17	I/O	<b>P1[1]</b> — Port 1 bit 1.
		I	<b>RXD</b> — Receiver input for the serial port.
P1[2]/T0/SCL	12	I/O	<b>P1[2]</b> — Port 1 bit 2 (open-drain when used as output).
		I/O	<b>T0</b> — Timer/counter 0 external count input or overflow output (open-drain when used as output).
		I/O	<b>SCL</b> — I <sup>2</sup> C-bus serial clock input/output.
P1[3]/INT0/SDA	11	I/O	<b>P1[3]</b> — Port 1 bit 3 (open-drain when used as output).
		I	<b>INT0</b> — External interrupt 0 input.
		I/O	<b>SDA</b> — I <sup>2</sup> C-bus serial data input/output.
P1[4]/INT1	10	I	<b>P1[4]</b> — Port 1 bit 4.
		I	<b>INT1</b> — External interrupt 1 input.

Table 3. Pin description ...continued

Symbol	Pin	Type	Description
P1[5]/RST	6	I	<b>P1[5]</b> — Port 1 bit 5 (input only).
		I	<b>RST</b> — External Reset input during power-on or if selected via UCFG1. When functioning as a reset input, a LOW on this pin resets the microcontroller, causing I/O ports and peripherals to take on their default states, and the processor begins execution at address 0. Also used during a power-on sequence to force In-System Programming mode. <b>When using an oscillator frequency above 12 MHz, the reset input function of P1[5] must be enabled. An external circuit is required to hold the device in reset at power-up until V<sub>DD</sub> has reached its specified level. When system power is removed V<sub>DD</sub> will fall below the minimum specified operating voltage. When using an oscillator frequency above 12 MHz, in some applications, an external brownout detect circuit may be required to hold the device in reset when V<sub>DD</sub> falls below the minimum specified operating range.</b>
P1[6]	5	I/O	<b>P1[6]</b> — Port 1 bit 6.
P1[7]/AD04	4	I/O	<b>P1[7]</b> — Port 1 bit 7.
		I	<b>AD04</b> — ADC0 channel 4 analog input.
P2[0] to P2[7]		I/O	<b>Port 2:</b> Port 2 is an 8-bit I/O port with a user-configurable output type. During reset Port 2 latches are configured in the input only mode with the internal pull-up disabled. The operation of Port 2 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to <a href="#">Section 7.13.1 “Port configurations”</a> and <a href="#">Table 10 “Static characteristics”</a> for details.  All pins have Schmitt triggered inputs.  Port 2 also provides various special functions as described below:
P2[0]/AD07	1	I/O	<b>P2[0]</b> — Port 2 bit 0.
		I	<b>AD07</b> — ADC0 channel 7 analog input.
P2[1]/AD06	2	I/O	<b>P2[1]</b> — Port 2 bit 1.
		I	<b>AD06</b> — ADC0 channel 6 analog input.
P2[2]/MOSI	13	I/O	<b>P2[2]</b> — Port 2 bit 2.
		I/O	<b>MOSI</b> — SPI master out slave in. When configured as master, this pin is output; when configured as slave, this pin is input.
P2[3]/MISO	14	I/O	<b>P2[3]</b> — Port 2 bit 3.
		I/O	<b>MISO</b> — When configured as master, this pin is input, when configured as slave, this pin is output.
P2[4]/SS	15	I/O	<b>P2[4]</b> — Port 2 bit 4.
		I	<b>SS</b> — SPI Slave select.
P2[5]/SPICLK	16	I/O	<b>P2[5]</b> — Port 2 bit 5.
		I/O	<b>SPICLK</b> — SPI clock. When configured as master, this pin is output; when configured as slave, this pin is input.
P2[6]	27	I/O	<b>P2[6]</b> — Port 2 bit 6.
P2[7]	28	I/O	<b>P2[7]</b> — Port 2 bit 7.



Table 3. Pin description ...continued

Symbol	Pin	Type	Description
P3[1:0]		I/O	<p><b>Port 3:</b> Port 3 is a 2-bit I/O port with a user-configurable output type. During reset Port 3 latches are configured in the input only mode with the internal pull-up disabled. The operation of Port 3 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to <a href="#">Section 7.13.1 “Port configurations”</a> and <a href="#">Table 10 “Static characteristics”</a> for details.</p> <p>All pins have Schmitt triggered inputs.</p> <p>Port 3 also provides various special functions as described below:</p>
P3[0]/XTAL2/ CLKOUT	9	I/O	<b>P3[0]</b> — Port 3 bit 0.
		O	<b>XTAL2</b> — Output from the oscillator amplifier (when a crystal oscillator option is selected via the flash configuration).
		O	<b>CLKOUT</b> — CPU clock divided by 2 when enabled via SFR bit (ENCLK -TRIM.6). It can be used if the CPU clock is the internal RC oscillator, watchdog oscillator or external clock input, except when XTAL1/XTAL2 are used to generate clock source for the RTC/system timer.
P3[1]/XTAL1	8	I/O	<b>P3[1]</b> — Port 3 bit 1.
		I	<b>XTAL1</b> — Input to the oscillator circuit and internal clock generator circuits (when selected via the flash configuration). It can be a port pin if internal RC oscillator or watchdog oscillator is used as the CPU clock source, <b>and</b> if XTAL1/XTAL2 are not used to generate the clock for the RTC/system timer.
V <sub>SS</sub>	7	I	<b>ground:</b> 0 V reference.
V <sub>DD</sub>	21	I	<b>power supply:</b> This is the power supply voltage for normal operation as well as Idle and Power-down modes.

[1] Input/output for P1[0] to P1[4], P1[6], P1[7]. Input for P1[5].

## 7. Functional description

---

**Remark:** Please refer to the P89LPC9381 *User's Manual* for a more detailed functional description.

### 7.1 Special function registers

**Remark:** SFR accesses are restricted in the following ways:

- User must **not** attempt to access any SFR locations not defined.
- Accesses to any defined SFR locations must be strictly for the functions for the SFRs.
- SFR bits labeled '-', '0' or '1' can **only** be written and read as follows:
  - '-' Unless otherwise specified, **must** be written with '0', but can return any value when read (even if it was written with '0'). It is a reserved bit and may be used in future derivatives.
  - '0' **must** be written with '0', and will return a '0' when read.
  - '1' **must** be written with '1', and will return a '1' when read.

Table 4. P89LPC9381 Special function registers

\* indicates SFRs that are bit addressable.

Name	Description	SFR addr.	Bit functions and addresses								Reset value	
			MSB							LSB	Hex	Binary
Bit address			E7	E6	E5	E4	E3	E2	E1	E0		
ACC*	Accumulator	E0H									00	0000 0000
AD0CON	ADC0 control register	97H	ENBIO	ENADCI 0	TMM0	EDGE0	ADCIO	ENADC0	ADCS01	ADCS00	00	0000 0000
AD0INS	ADC0 input select	A3H	ADI07	ADI06	ADI05	ADI04	ADI03	ADI02	ADI01	ADI00	00	0000 0000
AD0MODA	ADC0 mode register A	C0H	BNDIO	BURST0	SCC0	SCAN0	-	-	-	-	00	0000 0000
AD0MODB	ADC0 mode register B	A1H	CLK2	CLK1	CLK0	-	-	-	-	-	00	000x 0000
AUXR1	Auxiliary function register	A2H	CLKLP	EBRR	ENT1	ENT0	SRST	0	-	DPS	00	0000 00x0
Bit address			F7	F6	F5	F4	F3	F2	F1	F0		
B*	B register	F0H									00	0000 0000
BRGR0 <sup>[1]</sup>	Baud rate generator rate low	BEH									00	0000 0000
BRGR1 <sup>[1]</sup>	Baud rate generator rate high	BFH									00	0000 0000
BRGCON	Baud rate generator control	BDH	-	-	-	-	-	-	SBRGS	BRGEN	00 <sup>[1]</sup>	xxxx xx00
CMP1	Comparator 1 control register	ACH	-	-	CE1	CP1	CN1	OE1	CO1	CMF1	00 <sup>[2]</sup>	xx00 0000
CMP2	Comparator 2 control register	ADH	-	-	CE2	CP2	CN2	OE2	CO2	CMF2	00 <sup>[2]</sup>	xx00 0000
DIVM	CPU clock divide-by-M control	95H									00	0000 0000
DPTR	Data pointer (2 B)											
DPH	Data pointer high	83H									00	0000 0000
DPL	Data pointer low	82H									00	0000 0000
FMADRH	Program flash address high	E7H									00	0000 0000
FMADRL	Program flash address low	E6H									00	0000 0000

**Table 4. P89LPC9381 Special function registers ...continued**

\* indicates SFRs that are bit addressable.

Name	Description	SFR addr.	Bit functions and addresses								Reset value	
			MSB								LSB	Hex
FMCON	Program flash control (Read)	E4H	BUSY	-	-	-	HVA	HVE	SV	OI	70	0111 0000
	Program flash control (Write)	E4H										
FMDATA	Program flash data	E5H									00	0000 0000
I2ADR	I <sup>2</sup> C slave address register	DBH	I2ADR.6	I2ADR.5	I2ADR.4	I2ADR.3	I2ADR.2	I2ADR.1	I2ADR.0	GC	00	0000 0000
		<b>Bit address</b>	<b>DF</b>	<b>DE</b>	<b>DD</b>	<b>DC</b>	<b>DB</b>	<b>DA</b>	<b>D9</b>	<b>D8</b>		
I2CON*	I <sup>2</sup> C control register	D8H	-	I2EN	STA	STO	SI	AA	-	CRSEL	00	x000 00x0
I2DAT	I <sup>2</sup> C data register	DAH										
I2SCLH	Serial clock generator/SCL duty cycle register high	DDH									00	0000 0000
I2SCLL	Serial clock generator/SCL duty cycle register low	DCH									00	0000 0000
I2STAT	I <sup>2</sup> C status register	D9H	STA.4	STA.3	STA.2	STA.1	STA.0	0	0	0	F8	1111 1000
ICRAH	Input capture A register high	ABH									00	0000 0000
ICRAL	Input capture A register low	AAH									00	0000 0000
ICRBH	Input capture B register high	AFH									00	0000 0000
ICRBL	Input capture B register low	AEH									00	0000 0000
		<b>Bit address</b>	<b>AF</b>	<b>AE</b>	<b>AD</b>	<b>AC</b>	<b>AB</b>	<b>AA</b>	<b>A9</b>	<b>A8</b>		
IEN0*	Interrupt enable 0	A8H	EA	EWDRT	EBO	ES/ESR	ET1	EX1	ET0	EX0	00	0000 0000
		<b>Bit address</b>	<b>EF</b>	<b>EE</b>	<b>ED</b>	<b>EC</b>	<b>EB</b>	<b>EA</b>	<b>E9</b>	<b>E8</b>		
IEN1*	Interrupt enable 1	E8H	EIEE	EST	-	-	ESPI	EC	EKBI	EI2C	00 <sup>[2]</sup>	00x0 0000
IEN2	Interrupt enable 2	D5H	-	-	-	-	-	-	EADC	-	00 <sup>[2]</sup>	00x0 0000
		<b>Bit address</b>	<b>BF</b>	<b>BE</b>	<b>BD</b>	<b>BC</b>	<b>BB</b>	<b>BA</b>	<b>B9</b>	<b>B8</b>		
IPO*	Interrupt priority 0	B8H	-	PWDRT	PBO	PS/PSR	PT1	PX1	PT0	PX0	00 <sup>[2]</sup>	x000 0000

**Table 4. P89LPC9381 Special function registers ...continued**

\* indicates SFRs that are bit addressable.

Name	Description	SFR addr.	Bit functions and addresses								Reset value	
			MSB								LSB	
IP0H	Interrupt priority 0 high	B7H	-	PWDRT H	PBOH	PSH/ PSRH	PT1H	PX1H	PT0H	PX0H	00[2]	x000 0000
	<b>Bit address</b>		<b>FF</b>	<b>FE</b>	<b>FD</b>	<b>FC</b>	<b>FB</b>	<b>FA</b>	<b>F9</b>	<b>F8</b>		
IP1*	Interrupt priority 1	F8H	-	PST	-	-	PSPI	PC	PKBI	PI2C	00[2]	00x0 0000
IP1H	Interrupt priority 1 high	F7H	-	PSTH	-	-	PSPIH	PCH	PKBIH	PI2CH	00[2]	00x0 0000
IP2	Interrupt priority 2	D6H	-	-	-	-	-	-	PADC	-	00[2]	00x0 0000
IP2H	Interrupt priority 2 high	D7H	-	-	-	-	-	-	PADCH	-	00[2]	00x0 0000
KBCON	Keypad control register	94H	-	-	-	-	-	-	PATN _SEL	KBIF	00[2]	xxxx xx00
KBMASK	Keypad interrupt mask register	86H									00	0000 0000
KBPATN	Keypad pattern register										FF	1111 1111
	<b>Bit address</b>		<b>87</b>	<b>86</b>	<b>85</b>	<b>84</b>	<b>83</b>	<b>82</b>	<b>81</b>	<b>80</b>		
P0*	Port 0	80H	T1/KB7	CMP1 /KB6	CMPREF /KB5	CIN1A /KB4	CIN1B /KB3	CIN2A /KB2	CIN2B /KB1	CMP2 /KB0		[2]
	<b>Bit address</b>		<b>97</b>	<b>96</b>	<b>95</b>	<b>94</b>	<b>93</b>	<b>92</b>	<b>91</b>	<b>90</b>		
P1*	Port 1	90H	-	-	RST	INT1	INT0/ SDA	T0/SCL	RXD	TXD		[2]
	<b>Bit address</b>		<b>97</b>	<b>96</b>	<b>95</b>	<b>94</b>	<b>93</b>	<b>92</b>	<b>91</b>	<b>90</b>		
P2*	Port 2	A0H	-	-	SPICLK	SS	MISO	MOSI	-	-		[2]
	<b>Bit address</b>		<b>B7</b>	<b>B6</b>	<b>B5</b>	<b>B4</b>	<b>B3</b>	<b>B2</b>	<b>B1</b>	<b>B0</b>		
P3*	Port 3	B0H	-	-	-	-	-	-	XTAL1	XTAL2		[2]
P0M1	Port 0 output mode 1	84H	(P0M1.7)	(P0M1.6)	(P0M1.5)	(P0M1.4)	(P0M1.3)	(P0M1.2)	(P0M1.1)	(P0M1.0)	FF[2]	1111 1111
P0M2	Port 0 output mode 2	85H	(P0M2.7)	(P0M2.6)	(P0M2.5)	(P0M2.4)	(P0M2.3)	(P0M2.2)	(P0M2.1)	(P0M2.0)	00[2]	0000 0000
P1M1	Port 1 output mode 1	91H	(P1M1.7)	(P1M1.6)	-	(P1M1.4)	(P1M1.3)	(P1M1.2)	(P1M1.1)	(P1M1.0)	D3[2]	11x1 xx11
P1M2	Port 1 output mode 2	92H	(P1M2.7)	(P1M2.6)	-	(P1M2.4)	(P1M2.3)	(P1M2.2)	(P1M2.1)	(P1M2.0)	00[2]	00x0 xx00
P2M1	Port 2 output mode 1	A4H	(P2M1.7)	(P2M1.6)	(P2M1.5)	(P2M1.4)	(P2M1.3)	(P2M1.2)	(P2M1.1)	(P2M1.0)	FF[2]	1111 1111
P2M2	Port 2 output mode 2	A5H	(P2M2.7)	(P2M2.6)	(P2M2.5)	(P2M2.4)	(P2M2.3)	(P2M2.2)	(P2M2.1)	(P2M2.0)	00[2]	0000 0000
P3M1	Port 3 output mode 1	B1H	-	-	-	-	-	-	(P3M1.1)	(P3M1.0)	03[2]	xxxx xx11
P3M2	Port 3 output mode 2	B2H	-	-	-	-	-	-	(P3M2.1)	(P3M2.0)	00[2]	xxxx xx00

**Table 4. P89LPC9381 Special function registers ...continued**

\* indicates SFRs that are bit addressable.

Name	Description	SFR addr.	Bit functions and addresses								Reset value	
			MSB							LSB	Hex	Binary
PCON	Power control register	87H	SMOD1	SMOD0	BOPD	BOI	GF1	GF0	PMOD1	PMOD0	00	0000 0000
PCONA	Power control register A	B5H	RTCPD	-	VCPD	ADPD	I2PD	SPPD	SPD	-	00 <sup>[2]</sup>	0000 0000
		<b>Bit address</b>	<b>D7</b>	<b>D6</b>	<b>D5</b>	<b>D4</b>	<b>D3</b>	<b>D2</b>	<b>D1</b>	<b>D0</b>		
PSW*	Program status word	D0H	CY	AC	F0	RS1	RS0	OV	F1	P	00	0000 0000
PT0AD	Port 0 digital input disable	F6H	-	-	PT0AD.5	PT0AD.4	PT0AD.3	PT0AD.2	PT0AD.1	-	00	xx00 000x
RSTSRC	Reset source register	DFH	-	-	BOF	POF	R_BK	R_WD	R_SF	R_EX		<sup>[3]</sup>
RTCCON	RTC control	D1H	RTCF	RTCS1	RTCS0	-	-	-	ERTC	RTCEN	60 <sup>[2][4]</sup>	011x xx00
RTCH	RTC register high	D2H									00 <sup>[4]</sup>	0000 0000
RTCL	RTC register low	D3H									00 <sup>[4]</sup>	0000 0000
SADDR	Serial port address register	A9H									00	0000 0000
SADEN	Serial port address enable	B9H									00	0000 0000
SBUF	Serial Port data buffer register	99H									xx	xxxx xxxx
		<b>Bit address</b>	<b>9F</b>	<b>9E</b>	<b>9D</b>	<b>9C</b>	<b>9B</b>	<b>9A</b>	<b>99</b>	<b>98</b>		
SCON*	Serial port control	98H	SM0/FE	SM1	SM2	REN	TB8	RB8	TI	RI	00	0000 0000
SSTAT	Serial port extended status register	BAH	DBMOD	INTLO	CIDIS	DBISEL	FE	BR	OE	STINT	00	0000 0000
SP	Stack pointer	81H									07	0000 0111
SPCTL	SPI control register	E2H	SSIG	SPEN	DORD	MSTR	CPOL	CPHA	SPR1	SPR0	04	0000 0100
SPSTAT	SPI status register	E1H	SPIF	WCOL	-	-	-	-	-	-	00	00xx xxxx
SPDAT	SPI data register	E3H									00	0000 0000
TAMOD	Timer 0 and 1 auxiliary mode	8FH	-	-	-	T1M2	-	-	-	T0M2	00	xxx0 xxx0
		<b>Bit address</b>	<b>8F</b>	<b>8E</b>	<b>8D</b>	<b>8C</b>	<b>8B</b>	<b>8A</b>	<b>89</b>	<b>88</b>		
TCON*	Timer 0 and 1 control	88H	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00	0000 0000
TH0	Timer 0 high	8CH									00	0000 0000
TH1	Timer 1 high	8DH									00	0000 0000
TL0	Timer 0 low	8AH									00	0000 0000
TL1	Timer 1 low	8BH									00	0000 0000

**Table 4. P89LPC9381 Special function registers ...continued**

\* indicates SFRs that are bit addressable.

Name	Description	SFR addr.	Bit functions and addresses								Reset value	
			MSB				LSB				Hex	Binary
TMOD	Timer 0 and 1 mode	89H	T1GATE	T1C/T	T1M1	T1M0	T0GATE	T0C/T	T0M1	T0M0	00	0000 0000
TPCR2H	Prescaler control register high	CBH	-	-	-	-	-	-	TPCR2H.1	TPCR2H.0	00	xxxx xx00
TPCR2L	Prescaler control register low	CAH									00	0000 0000
TRIM	Internal oscillator trim register	96H	RCCLK	ENCLK	TRIM.5	TRIM.4	TRIM.3	TRIM.2	TRIM.1	TRIM.0		<a href="#">[4][5]</a>
WDCON	Watchdog control register	A7H	PRE2	PRE1	PRE0	-	-	WDRUN	WDTOF	WDCLK		<a href="#">[4][6]</a>
WDL	Watchdog load	C1H									FF	1111 1111
WFEEED1	Watchdog feed 1	C2H										
WFEEED2	Watchdog feed 2	C3H										

- [1] BRGR1 and BRGR0 must only be written if BRGEN in BRGCON SFR is logic 0. If any are written while BRGEN = 1, the result is unpredictable.
- [2] All ports are in input only (high-impedance) state after power-up.
- [3] The RSTSRC register reflects the cause of the P89LPC9381 reset. Upon a power-up reset, all reset source flags are cleared except POF and BOF; the power-on reset value is xx11 0000.
- [4] The only reset source that affects these SFRs is power-on reset.
- [5] On power-on reset, the TRIM SFR is initialized with a factory preprogrammed value. Other resets will not cause initialization of the TRIM register.
- [6] After reset, the value is 1110 01x1, i.e., PRE2 to PRE0 are all logic 1, WDRUN = 1 and WDCLK = 1. WDTOF bit is logic 1 after watchdog reset and is logic 0 after power-on reset. Other resets will not affect WDTOF.

Table 5. P89LPC9381 extended special function registers

Name	Description	SFR addr.	Bit functions and addresses		Reset value	
			MSB	LSB	Hex	Binary
ADC0HBND	ADC0 high_boundary register, left (MSB)	FFEFH			FF	1111 1111
ADC0LBND	ADC0 low_boundary register (MSB)	FFEEH			00	0000 0000
AD0DAT0R	ADC0 data register 0, right (LSB)	FFFEH		AD0DAT0[7:0]	00	0000 0000
AD0DAT0L	ADC0 data register 0, left (MSB)	FFFFH		AD0DAT0[9:2]	00	0000 0000
AD0DAT1R	ADC0 data register 1, right (LSB)	FFFCH		AD0DAT1[7:0]	00	0000 0000
AD0DAT1L	ADC0 data register 1, left (MSB)	FFFDH		AD0DAT1[9:2]	00	0000 0000
AD0DAT2R	ADC0 data register 2, right (LSB)	FFFAH		AD0DAT2[7:0]	00	0000 0000
AD0DAT2L	ADC0 data register 2, left (MSB)	FFFBH		AD0DAT2[9:2]	00	0000 0000
AD0DAT3R	ADC0 data register 3, right (LSB)	FFF8H		AD0DAT3[7:0]	00	0000 0000
AD0DAT3L	ADC0 data register 3, left (MSB)	FFF9H		AD0DAT3[9:2]	00	0000 0000
AD0DAT4R	ADC0 data register 4, right (LSB)	FFF6H		AD0DAT4[7:0]	00	0000 0000
AD0DAT4L	ADC0 data register 4, left (MSB)	FFF7H		AD0DAT4[9:2]	00	0000 0000
AD0DAT5R	ADC0 data register 5, right (LSB)	FFF4H		AD0DAT5[7:0]	00	0000 0000
AD0DAT5L	ADC0 data register 5, left (MSB)	FFF5H		AD0DAT5[9:2]	00	0000 0000
AD0DAT6R	ADC0 data register 6, right (LSB)	FFF2H		AD0DAT6[7:0]	00	0000 0000
AD0DAT6L	ADC0 data register 6, left (MSB)	FFF3H		AD0DAT6[9:2]	00	0000 0000
AD0DAT7R	ADC0 data register 7, right (LSB)	FFF0H		AD0DAT7[7:0]		
AD0DAT7L	ADC0 data register 7, left (MSB)	FFF1H		AD0DAT7[9:2]		
BNDSTA0	ADC0 boundary status register	FFEDH				



## 7.2 Enhanced CPU

The P89LPC9381 uses an enhanced 80C51 CPU which runs at six times the speed of standard 80C51 devices. A machine cycle consists of two CPU clock cycles, and most instructions execute in one or two machine cycles.

## 7.3 Clocks

### 7.3.1 Clock definitions

The P89LPC9381 device has several internal clocks as defined below:

**OSCCLK** — Input to the DIVM clock divider. OSCCLK is selected from one of four clock sources (see [Figure 4](#)) and can also be optionally divided to a slower frequency (see [Section 7.8 “CCLK modification: DIVM register”](#)).

**Note:**  $f_{osc}$  is defined as the OSCCLK frequency.

**CCLK** — CPU clock; output of the clock divider. There are two CCLK cycles per machine cycle, and most instructions are executed in one to two machine cycles (two or four CCLK cycles).

**RCCLK** — The internal 7.373 MHz RC oscillator output.

**PCLK** — Clock for the various peripheral devices and is  $CCLK/2$ .

### 7.3.2 CPU clock (OSCCLK)

The P89LPC9381 provides several user-selectable oscillator options in generating the CPU clock. This allows optimization for a range of needs from high precision to lowest possible cost. These options are configured when the flash is programmed and include an on-chip watchdog oscillator, an on-chip RC oscillator, an oscillator using an external crystal, or an external clock source. The crystal oscillator can be optimized for low, medium, or high frequency crystals covering a range from 20 kHz to 18 MHz.

### 7.3.3 Low speed oscillator option

This option supports an external crystal in the range of 20 kHz to 100 kHz. Ceramic resonators are also supported in this configuration.

### 7.3.4 Medium speed oscillator option

This option supports an external crystal in the range of 100 kHz to 4 MHz. Ceramic resonators are also supported in this configuration.

### 7.3.5 High speed oscillator option

This option supports an external crystal in the range of 4 MHz to 18 MHz. Ceramic resonators are also supported in this configuration.

### 7.3.6 Clock output

The P89LPC9381 supports a user-selectable clock output function on the XTAL2/CLKOUT pin when crystal oscillator is not being used. This condition occurs if another clock source has been selected (on-chip RC oscillator, watchdog oscillator, external clock input on X1) and if the RTC is not using the crystal oscillator as its clock source. This allows external devices to synchronize to the P89LPC9381. This output is enabled by the ENCLK bit in the TRIM register.

The frequency of this clock output is  $\frac{1}{2}$  that of the CCLK. If the clock output is not needed in Idle mode, it may be turned off prior to entering Idle, saving additional power.

#### 7.4 On-chip RC oscillator option

The P89LPC9381 has a 6-bit TRIM register that can be used to tune the frequency of the RC oscillator. During reset, the TRIM value is initialized to a factory pre-programmed value to adjust the oscillator frequency to  $7.373 \text{ MHz} \pm 1\%$  at room temperature. End-user applications can write to the TRIM register to adjust the on-chip RC oscillator to other frequencies.

#### 7.5 Watchdog oscillator option

The watchdog has a separate oscillator which has a frequency of 400 kHz. This oscillator can be used to save power when a high clock frequency is not needed.

#### 7.6 External clock input option

In this configuration, the processor clock is derived from an external source driving the XTAL1/P3[1] pin. The rate may be from 0 Hz up to 18 MHz. The XTAL2/P3[0] pin may be used as a standard port pin or a clock output. **When using an oscillator frequency above 12 MHz, the reset input function of P1[5] must be enabled. An external circuit is required to hold the device in reset at power-up until  $V_{DD}$  has reached its specified level. When system power is removed  $V_{DD}$  will fall below the minimum specified operating voltage. When using an oscillator frequency above 12 MHz, in some applications, an external brownout detect circuit may be required to hold the device in reset when  $V_{DD}$  falls below the minimum specified operating voltage.**

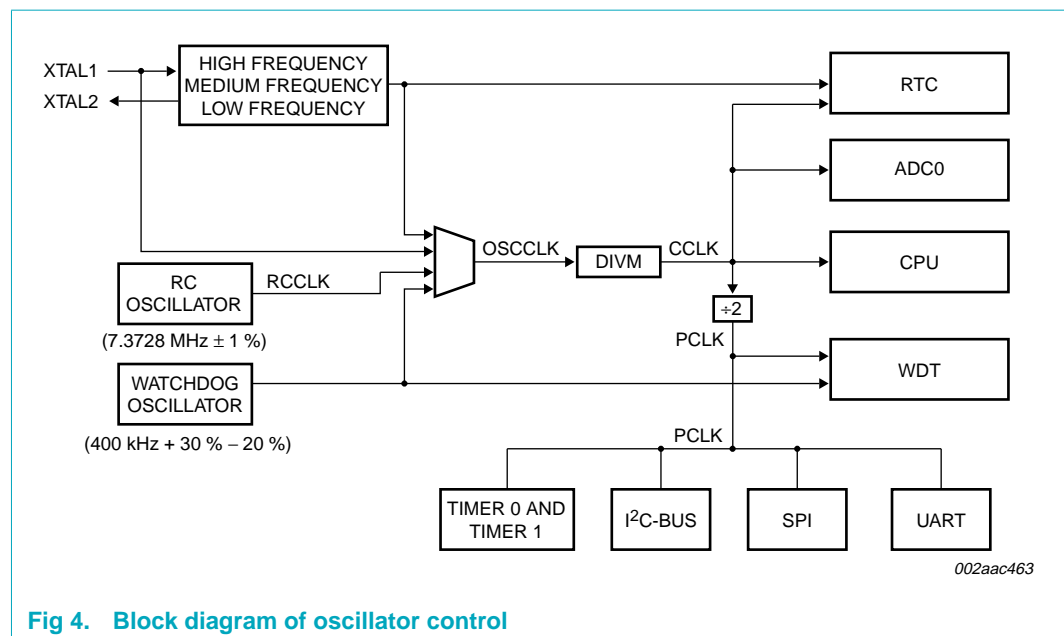


Fig 4. Block diagram of oscillator control

### 7.7 CCLK wake-up delay

The P89LPC9381 has an internal wake-up timer that delays the clock until it stabilizes depending on the clock source used. If the clock source is any of the three crystal selections (low, medium and high frequencies) the delay is 992 OSCCLK cycles plus 60 to 100  $\mu$ s. If the clock source is either the internal RC oscillator, watchdog oscillator, or external clock, the delay is 224 OSCCLK cycles plus 60  $\mu$ s to 100  $\mu$ s.

### 7.8 CCLK modification: DIVM register

The OSCCLK frequency can be divided down up to 510 times by configuring a dividing register, DIVM, to generate CCLK. This feature makes it possible to temporarily run the CPU at a lower rate, reducing power consumption. By dividing the clock, the CPU can retain the ability to respond to events that would not exit Idle mode by executing its normal program at a lower rate. This can also allow bypassing the oscillator start-up time in cases where Power-down mode would otherwise be used. The value of DIVM may be changed by the program at any time without interrupting code execution.

### 7.9 Low power select

The P89LPC9381 is designed to run at 12 MHz (CCLK) maximum. However, if CCLK is 8 MHz or slower, the CLKLP SFR bit (AUXR1.7) can be set to '1' to lower the power consumption further. On any reset, CLKLP is '0' allowing highest performance access. This bit can then be set in software if CCLK is running at 8 MHz or slower.

### 7.10 Memory organization

The various P89LPC9381 memory spaces are as follows:

- DATA  
128 B of internal data memory space (00H:7FH) accessed via direct or indirect addressing, using instructions other than MOVX and MOVC. All or part of the Stack may be in this area.
- IDATA  
Indirect Data. 256 B of internal data memory space (00H:FFH) accessed via indirect addressing using instructions other than MOVX and MOVC. All or part of the Stack may be in this area. This area includes the DATA area and the 128 B immediately above it.
- SFR  
Special Function Registers. Selected CPU registers and peripheral control and status registers, accessible only via direct addressing.

- CODE  
64 kB of Code memory space, accessed as part of program execution and via the MOVC instruction. The P89LPC9381 has 4 kB of on-chip Code memory.

## 7.11 Data RAM arrangement

The 256 B of on-chip RAM are organized as shown in [Table 6](#).

**Table 6. On-chip data memory usages**

Type	Data RAM	Size
DATA	Memory that can be addressed directly and indirectly	128 B
IDATA	Memory that can be addressed indirectly	256 B

## 7.12 Interrupts

The P89LPC9381 uses a four priority level interrupt structure. This allows great flexibility in controlling the handling of the many interrupt sources. The P89LPC9381 supports 15 interrupt sources: external interrupts 0 and 1, timers 0 and 1, serial port TX, serial port RX, combined serial port RX/TX, brownout detect, watchdog/RTC, I<sup>2</sup>C, keyboard, comparators 1 and 2, SPI, write/ADC completion.

Each interrupt source can be individually enabled or disabled by setting or clearing a bit in the interrupt enable registers IEN0 or IEN1. The IEN0 register also contains a global disable bit, EA, which disables all interrupts.

Each interrupt source can be individually programmed to one of four priority levels by setting or clearing bits in the interrupt priority registers IP0, IP0H, IP1, and IP1H. An interrupt service routine in progress can be interrupted by a higher priority interrupt, but not by another interrupt of the same or lower priority. The highest priority interrupt service cannot be interrupted by any other interrupt source. If two requests of different priority levels are pending at the start of an instruction, the request of higher priority level is serviced.

If requests of the same priority level are pending at the start of an instruction, an internal polling sequence determines which request is serviced. This is called the arbitration ranking. Note that the arbitration ranking is only used to resolve pending requests of the same priority level.

### 7.12.1 External interrupt inputs

The P89LPC9381 has two external interrupt inputs as well as the Keypad Interrupt function. The two interrupt inputs are identical to those present on the standard 80C51 microcontrollers.

These external interrupts can be programmed to be level-triggered or edge-triggered by setting or clearing bit IT1 or IT0 in Register TCON.

In edge-triggered mode, if successive samples of the  $\overline{\text{INTn}}$  pin show a HIGH in one cycle and a LOW in the next cycle, the interrupt request flag IEn in TCON is set, causing an interrupt request.

If an external interrupt is enabled when the P89LPC9381 is put into Power-down or Idle mode, the interrupt will cause the processor to wake-up and resume operation. Refer to [Section 7.15 "Power reduction modes"](#) for details.

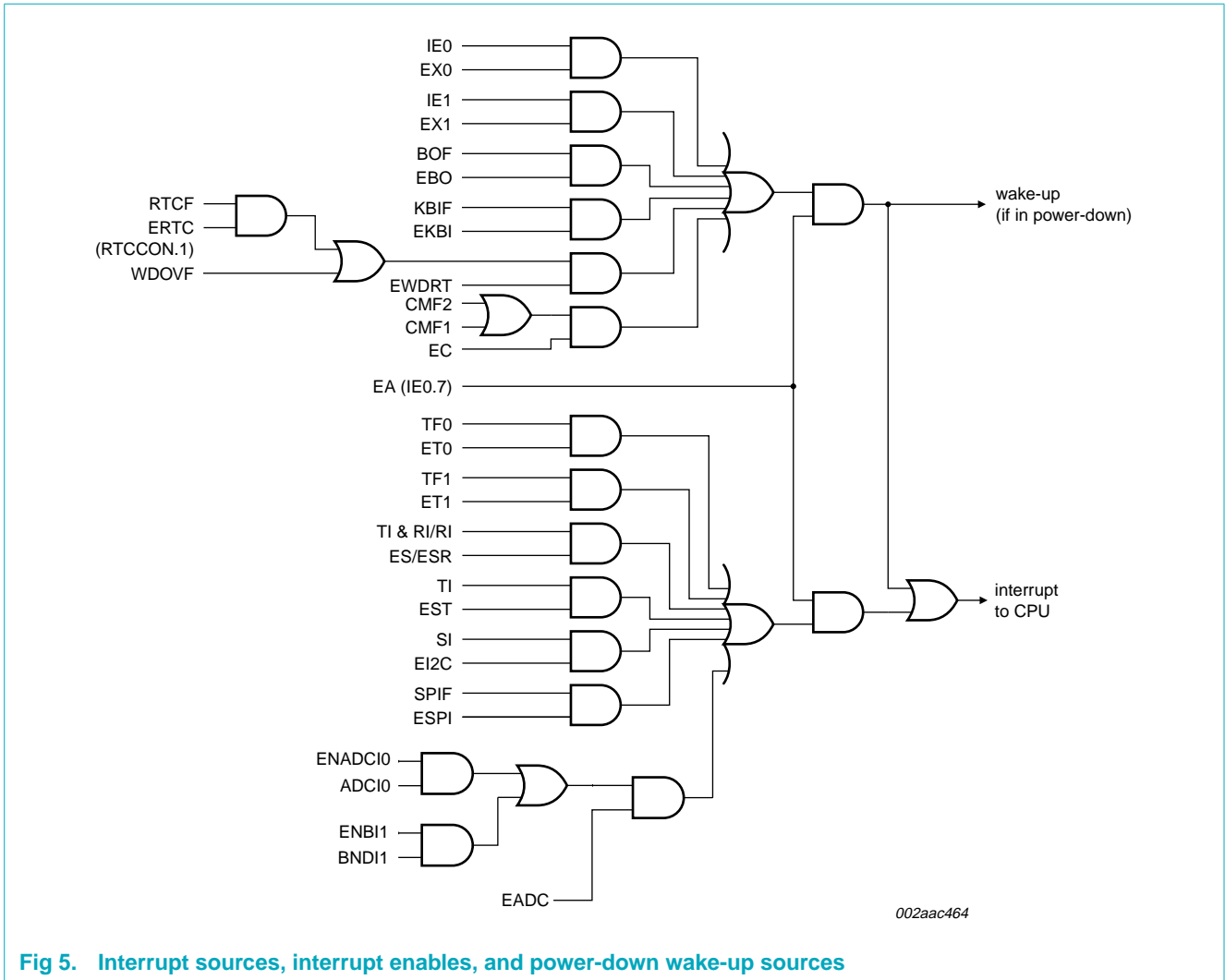


Fig 5. Interrupt sources, interrupt enables, and power-down wake-up sources

## 7.13 I/O ports

The P89LPC9381 has four I/O ports: Port 0, Port 1, Port 2, and Port 3. Ports 0, 1 and 2 are 8-bit ports, and Port 3 is a 2-bit port. The exact number of I/O pins available depends upon the clock and reset options chosen, as shown in [Table 7](#).

**Table 7. Number of I/O pins available**

Clock source	Reset option	Number of I/O pins (28-pin package)
On-chip oscillator or watchdog oscillator	No external reset (except during power-up)	26
	External $\overline{\text{RST}}$ pin supported	25
External clock input	No external reset (except during power-up)	25
	External $\overline{\text{RST}}$ pin supported <sup>[1]</sup>	24
Low/medium/high speed oscillator (external crystal or resonator)	No external reset (except during power-up)	24
	External $\overline{\text{RST}}$ pin supported <sup>[1]</sup>	23

[1] Required for operation above 12 MHz.

### 7.13.1 Port configurations

All but three I/O port pins on the P89LPC9381 may be configured by software to one of four types on a bit-by-bit basis. These are: quasi-bidirectional (standard 80C51 port outputs), push-pull, open-drain, and input-only. Two configuration registers for each port select the output type for each port pin.

1. P1[5] ( $\overline{\text{RST}}$ ) can only be an input and cannot be configured.
2. P1[2] (SCL/T0) and P1[3] (SDA/ $\overline{\text{INT0}}$ ) may only be configured to be either input-only or open-drain.

#### 7.13.1.1 Quasi-bidirectional output configuration

Quasi-bidirectional output type can be used as both an input and output without the need to reconfigure the port. This is possible because when the port outputs a logic HIGH, it is weakly driven, allowing an external device to pull the pin LOW. When the pin is driven LOW, it is driven strongly and able to sink a fairly large current. These features are somewhat similar to an open-drain output except that there are three pull-up transistors in the quasi-bidirectional output that serve different purposes.

The P89LPC9381 is a 3 V device, but the pins are 5 V tolerant. In quasi-bidirectional mode, if a user applies 5 V on the pin, there will be a current flowing from the pin to  $V_{\text{DD}}$ , causing extra power consumption. Therefore, applying 5 V in quasi-bidirectional mode is discouraged.

A quasi-bidirectional port pin has a Schmitt triggered input that also has a glitch suppression circuit.

#### 7.13.1.2 Open-drain output configuration

The open-drain output configuration turns off all pull-ups and only drives the pull-down transistor of the port driver when the port latch contains a logic 0. To be used as a logic output, a port configured in this manner must have an external pull-up, typically a resistor tied to  $V_{\text{DD}}$ .

An open-drain port pin has a Schmitt triggered input that also has a glitch suppression circuit.

#### 7.13.1.3 Input-only configuration

The input-only port configuration has no output drivers. It is a Schmitt triggered input that also has a glitch suppression circuit.

#### 7.13.1.4 Push-pull output configuration

The push-pull output configuration has the same pull-down structure as both the open-drain and the quasi-bidirectional output modes, but provides a continuous strong pull-up when the port latch contains a logic 1. The push-pull mode may be used when more source current is needed from a port output. A push-pull port pin has a Schmitt triggered input that also has a glitch suppression circuit.

### 7.13.2 Port 0 analog functions

The P89LPC9381 incorporates two Analog Comparators. In order to give the best analog function performance and to minimize power consumption, pins that are being used for analog functions must have the digital outputs and digital inputs disabled.

Digital outputs are disabled by putting the port output into the Input-Only (high-impedance) mode.

Digital inputs on Port 0 may be disabled through the use of the PT0AD register, bits 5:1. On any reset, PT0AD[5:1] defaults to '0's to enable digital functions.

### 7.13.3 Additional port features

After power-up, all pins are in Input-Only mode. **Please note that this is different from the LPC76x series of devices.**

- After power-up, all I/O pins except P1[5], may be configured by software.
- Pin P1[5] is input only. Pins P1[2] and P1[3] and are configurable for either input-only or open-drain.

Every output on the P89LPC9381 has been designed to sink typical LED drive current. However, there is a maximum total output current for all ports which must not be exceeded. Please refer to [Table 10 "Static characteristics"](#) for detailed specifications.

All ports pins that can function as an output have slew rate controlled outputs to limit noise generated by quickly switching output signals. The slew rate is factory-set to approximately 10 ns rise and fall times.

## 7.14 Power monitoring functions

The P89LPC9381 incorporates power monitoring functions designed to prevent incorrect operation during initial power-up and power loss or reduction during operation. This is accomplished with two hardware functions: Power-on detect and brownout detect.

### 7.14.1 Brownout detection

The brownout detect function determines if the power supply voltage drops below a certain level. The default operation is for a brownout detection to cause a processor reset, however it may alternatively be configured to generate an interrupt.

Brownout detection may be enabled or disabled in software.

If brownout detection is enabled, the brownout condition occurs when  $V_{DD}$  falls below the brownout trip voltage,  $V_{bo}$  (see [Table 10 "Static characteristics"](#)), and is negated when  $V_{DD}$  rises above  $V_{bo}$ . If the P89LPC9381 device is to operate with a power supply that can be below 2.7 V, BOE should be left in the unprogrammed state so that the device can operate at 2.4 V, otherwise continuous brownout reset may prevent the device from operating.

For correct activation of brownout detect, the  $V_{DD}$  rise and fall times must be observed. Please see [Table 10 "Static characteristics"](#) for specifications.

#### 7.14.2 Power-on detection

The Power-on detect has a function similar to the brownout detect, but is designed to work as power comes up initially, before the power supply voltage reaches a level where brownout detect can work. The POF flag in the RSTSRC register is set to indicate an initial power-up condition. The POF flag will remain set until cleared by software.

### 7.15 Power reduction modes

The P89LPC9381 supports three different power reduction modes. These modes are Idle mode, Power-down mode, and Total Power-down mode.

#### 7.15.1 Idle mode

Idle mode leaves peripherals running in order to allow them to activate the processor when an interrupt is generated. Any enabled interrupt source or reset may terminate Idle mode.

#### 7.15.2 Power-down mode

The Power-down mode stops the oscillator in order to minimize power consumption. The P89LPC9381 exits Power-down mode via any reset, or certain interrupts. In Power-down mode, the power supply voltage may be reduced to the data retention voltage ( $V_{DDR}$ ). This retains the RAM contents at the point where Power-down mode was entered. SFR contents are not guaranteed after  $V_{DD}$  has been lowered to  $V_{DDR}$ , therefore it is highly recommended to wake up the processor via reset in this case.  $V_{DD}$  must be raised to within the operating range before the Power-down mode is exited.

Some chip functions continue to operate and draw power during Power-down mode, increasing the total power used during power-down. These include: brownout detect, watchdog timer, comparators (note that comparators can be powered-down separately), and RTC/system timer. The internal RC oscillator is disabled unless both the RC oscillator has been selected as the system clock and the RTC is enabled.

#### 7.15.3 Total Power-down mode

This is the same as Power-down mode except that the brownout detection circuitry and the voltage comparators are also disabled to conserve additional power. The internal RC oscillator is disabled unless both the RC oscillator has been selected as the system clock **and** the RTC is enabled. If the internal RC oscillator is used to clock the RTC during power-down, there will be high power consumption. Please use an external low frequency clock to achieve low power with the RTC running during power-down.



## 7.16 Reset

The P1[5]/ $\overline{\text{RST}}$  pin can function as either an active-LOW reset input or as a digital input, P1[5]. The RPE (Reset Pin Enable) bit in UCFG1, when set to '1', enables the external reset input function on P1[5]. When cleared, P1[5] may be used as an input pin.

**Remark:** During a power-up sequence, The RPE selection is overridden and this pin will always function as a reset input. **An external circuit connected to this pin should not hold this pin LOW during a power-on sequence as this will keep the device in reset.** After power-up this input will function either as an external reset input or as a digital input as defined by the RPE bit. Only a power-up reset will temporarily override the selection defined by RPE bit. Other sources of reset will not override the RPE bit.

Reset can be triggered from the following sources:

- External reset pin (during power-up or if user configured via UCFG1);
- Power-on detect;
- Brownout detect;
- Watchdog timer;
- Software reset;
- UART break character detect reset.

For every reset source, there is a flag in the Reset Register, RSTSRC. The user can read this register to determine the most recent reset source. These flag bits can be cleared in software by writing a '0' to the corresponding bit. More than one flag bit may be set:

- During a power-on reset, both POF and BOF are set but the other flag bits are cleared.
- For any other reset, previously set flag bits that have not been cleared will remain set.

### 7.16.1 Reset vector

Following reset, the P89LPC9381 will fetch instructions from either address 0000H or the Boot address. The Boot address is formed by using the Boot Vector as the high byte of the address and the low byte of the address = 00H.

The Boot address will be used if a UART break reset occurs, or the non-volatile Boot Status bit (BOOTSTAT.0) = 1, or the device is forced into ISP mode during power-on (see P89LPC9381 *User's Manual*). Otherwise, instructions will be fetched from address 0000H.

## 7.17 Timers/counters 0 and 1

The P89LPC9381 has two general purpose counter/timers which are upward compatible with the standard 80C51 Timer 0 and Timer 1. Both can be configured to operate either as timers or event counter. An option to automatically toggle the T0 and/or T1 pins upon timer overflow has been added.

In the 'Timer' function, the register is incremented every machine cycle.

In the 'Counter' function, the register is incremented in response to a 1-to-0 transition at its corresponding external input pin, T0 or T1. In this function, the external input is sampled once during every machine cycle.

Timer 0 and Timer 1 have five operating modes (modes 0, 1, 2, 3 and 6). Modes 0, 1, 2 and 6 are the same for both Timers/Counters. Mode 3 is different.

#### 7.17.1 Mode 0

Putting either Timer into Mode 0 makes it look like an 8048 Timer, which is an 8-bit Counter with a divide-by-32 prescaler. In this mode, the Timer register is configured as a 13-bit register. Mode 0 operation is the same for Timer 0 and Timer 1.

#### 7.17.2 Mode 1

Mode 1 is the same as Mode 0, except that all 16 bits of the timer register are used.

#### 7.17.3 Mode 2

Mode 2 configures the Timer register as an 8-bit Counter with automatic reload. Mode 2 operation is the same for Timer 0 and Timer 1.

#### 7.17.4 Mode 3

When Timer 1 is in Mode 3 it is stopped. Timer 0 in Mode 3 forms two separate 8-bit counters and is provided for applications that require an extra 8-bit timer. When Timer 1 is in Mode 3 it can still be used by the serial port as a baud rate generator.

#### 7.17.5 Mode 6

In this mode, the corresponding timer can be changed to a PWM with a full period of 256 timer clocks.

#### 7.17.6 Timer overflow toggle output

Timers 0 and 1 can be configured to automatically toggle a port output whenever a timer overflow occurs. The same device pins that are used for the T0 and T1 count inputs are also used for the timer toggle outputs. The port outputs will be a logic 1 prior to the first timer overflow when this mode is turned on.

### 7.18 RTC/system timer

The P89LPC9381 has a simple RTC that allows a user to continue running an accurate timer while the rest of the device is powered-down. The RTC can be a wake-up or an interrupt source. The RTC is a 23-bit down counter comprised of a 7-bit prescaler and a 16-bit loadable down counter. When it reaches all '0's, the counter will be reloaded again and the RTCF flag will be set. The clock source for this counter can be either the CPU clock (CCLK) or the XTAL oscillator, provided that the XTAL oscillator is not being used as the CPU clock. If the XTAL oscillator is used as the CPU clock, then the RTC will use CCLK as its clock source. Only power-on reset will reset the RTC and its associated SFRs to the default state.

### 7.19 UART

The P89LPC9381 has an enhanced UART that is compatible with the conventional 80C51 UART except that Timer 2 overflow cannot be used as a baud rate source. The P89LPC9381 does include an independent Baud Rate Generator. The baud rate can be selected from the oscillator (divided by a constant), Timer 1 overflow, or the independent Baud Rate Generator. In addition to the baud rate generation, enhancements over the

standard 80C51 UART include Framing Error detection, automatic address recognition, selectable double buffering and several interrupt options. The UART can be operated in 4 modes: shift register, 8-bit UART, 9-bit UART, and CPU clock/32 or CPU clock/16.

### 7.19.1 Mode 0

Serial data enters and exits through RXD. TXD outputs the shift clock. 8 bits are transmitted or received, LSB first. The baud rate is fixed at  $\frac{1}{16}$  of the CPU clock frequency.

### 7.19.2 Mode 1

10 bits are transmitted (through TXD) or received (through RXD): a start bit (logic 0), 8 data bits (LSB first), and a stop bit (logic 1). When data is received, the stop bit is stored in RB8 in Special Function Register SCON. The baud rate is variable and is determined by the Timer 1 overflow rate or the Baud Rate Generator (described in [Section 7.19.5 "Baud rate generator and selection"](#)).

### 7.19.3 Mode 2

11 bits are transmitted (through TXD) or received (through RXD): start bit (logic 0), 8 data bits (LSB first), a programmable 9<sup>th</sup> data bit, and a stop bit (logic 1). When data is transmitted, the 9<sup>th</sup> data bit (TB8 in SCON) can be assigned the value of '0' or '1'. Or, for example, the parity bit (P, in the PSW) could be moved into TB8. When data is received, the 9<sup>th</sup> data bit goes into RB8 in Special Function Register SCON, while the stop bit is not saved. The baud rate is programmable to either  $\frac{1}{16}$  or  $\frac{1}{32}$  of the CPU clock frequency, as determined by the SMOD1 bit in PCON.

### 7.19.4 Mode 3

11 bits are transmitted (through TXD) or received (through RXD): a start bit (logic 0), 8 data bits (LSB first), a programmable 9<sup>th</sup> data bit, and a stop bit (logic 1). In fact, Mode 3 is the same as Mode 2 in all respects except baud rate. The baud rate in Mode 3 is variable and is determined by the Timer 1 overflow rate or the Baud Rate Generator (described in [Section 7.19.5 "Baud rate generator and selection"](#)).

### 7.19.5 Baud rate generator and selection

The P89LPC9381 enhanced UART has an independent Baud Rate Generator. The baud rate is determined by a baud-rate preprogrammed into the BRGR1 and BRGR0 SFRs which together form a 16-bit baud rate divisor value that works in a similar manner as Timer 1 but is much more accurate. If the baud rate generator is used, Timer 1 can be used for other timing functions.

The UART can use either Timer 1 or the baud rate generator output (see [Figure 6](#)). Note that Timer T1 is further divided by 2 if the SMOD1 bit (PCON.7) is cleared. The independent Baud Rate Generator uses OSCCLK.

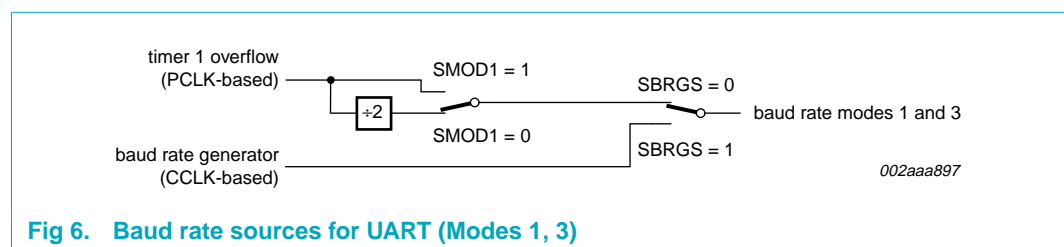


Fig 6. Baud rate sources for UART (Modes 1, 3)

### 7.19.6 Framing error

Framing error is reported in the status register (SSTAT). In addition, if SMOD0 (PCON.6) is '1', framing errors can be made available in SCON.7 respectively. If SMOD0 is '0', SCON.7 is SM0. It is recommended that SM0 and SM1 (SCON.7:6) are set up when SMOD0 is '0'.

### 7.19.7 Break detect

Break detect is reported in the status register (SSTAT). A break is detected when 11 consecutive bits are sensed LOW. The break detect can be used to reset the device and force the device into ISP mode.

### 7.19.8 Double buffering

The UART has a transmit double buffer that allows buffering of the next character to be written to SBUF while the first character is being transmitted. Double buffering allows transmission of a string of characters with only one stop bit between any two characters, as long as the next character is written between the start bit and the stop bit of the previous character.

Double buffering can be disabled. If disabled (DBMOD, i.e., SSTAT.7 = 0), the UART is compatible with the conventional 80C51 UART. If enabled, the UART allows writing to SBUF while the previous data is being shifted out. Double buffering is only allowed in Modes 1, 2 and 3. When operated in Mode 0, double buffering must be disabled (DBMOD = 0).

### 7.19.9 Transmit interrupts with double buffering enabled (modes 1, 2 and 3)

Unlike the conventional UART, in double buffering mode, the TX interrupt is generated when the double buffer is ready to receive new data.

### 7.19.10 The 9<sup>th</sup> bit (bit 8) in double buffering (modes 1, 2 and 3)

If double buffering is disabled TB8 can be written before or after SBUF is written, as long as TB8 is updated some time before that bit is shifted out. TB8 must not be changed until the bit is shifted out, as indicated by the TX interrupt.

If double buffering is enabled, TB8 **must** be updated before SBUF is written, as TB8 will be double-buffered together with SBUF data.

## 7.20 I<sup>2</sup>C-bus serial interface

I<sup>2</sup>C-bus uses two wires (SDA and SCL) to transfer information between devices connected to the bus, and it has the following features:

- Bidirectional data transfer between masters and slaves
- Multi master bus (no central master)
- Arbitration between simultaneously transmitting masters without corruption of serial data on the bus
- Serial clock synchronization allows devices with different bit rates to communicate via one serial bus
- Serial clock synchronization can be used as a handshake mechanism to suspend and resume serial transfer
- The I<sup>2</sup>C-bus may be used for test and diagnostic purposes.

A typical I<sup>2</sup>C-bus configuration is shown in [Figure 7](#). The P89LPC9381 device provides a byte-oriented I<sup>2</sup>C-bus interface that supports data transfers up to 400 kHz.

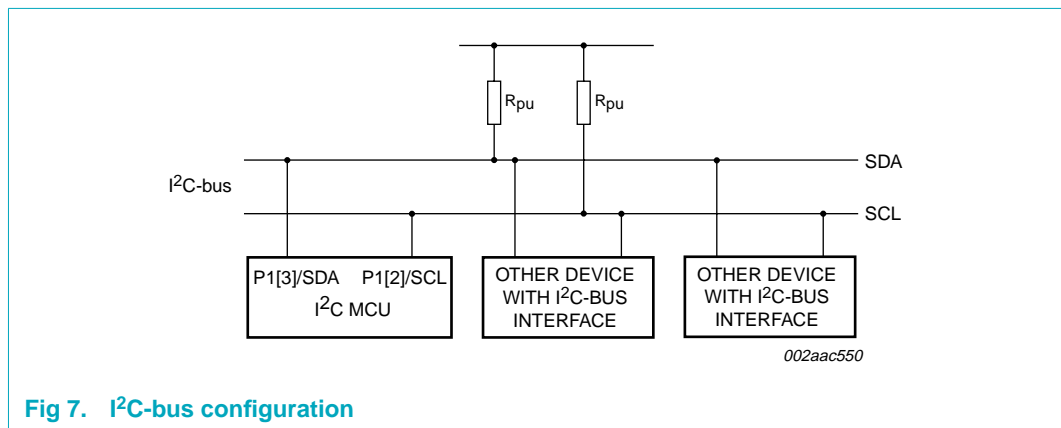


Fig 7. I<sup>2</sup>C-bus configuration

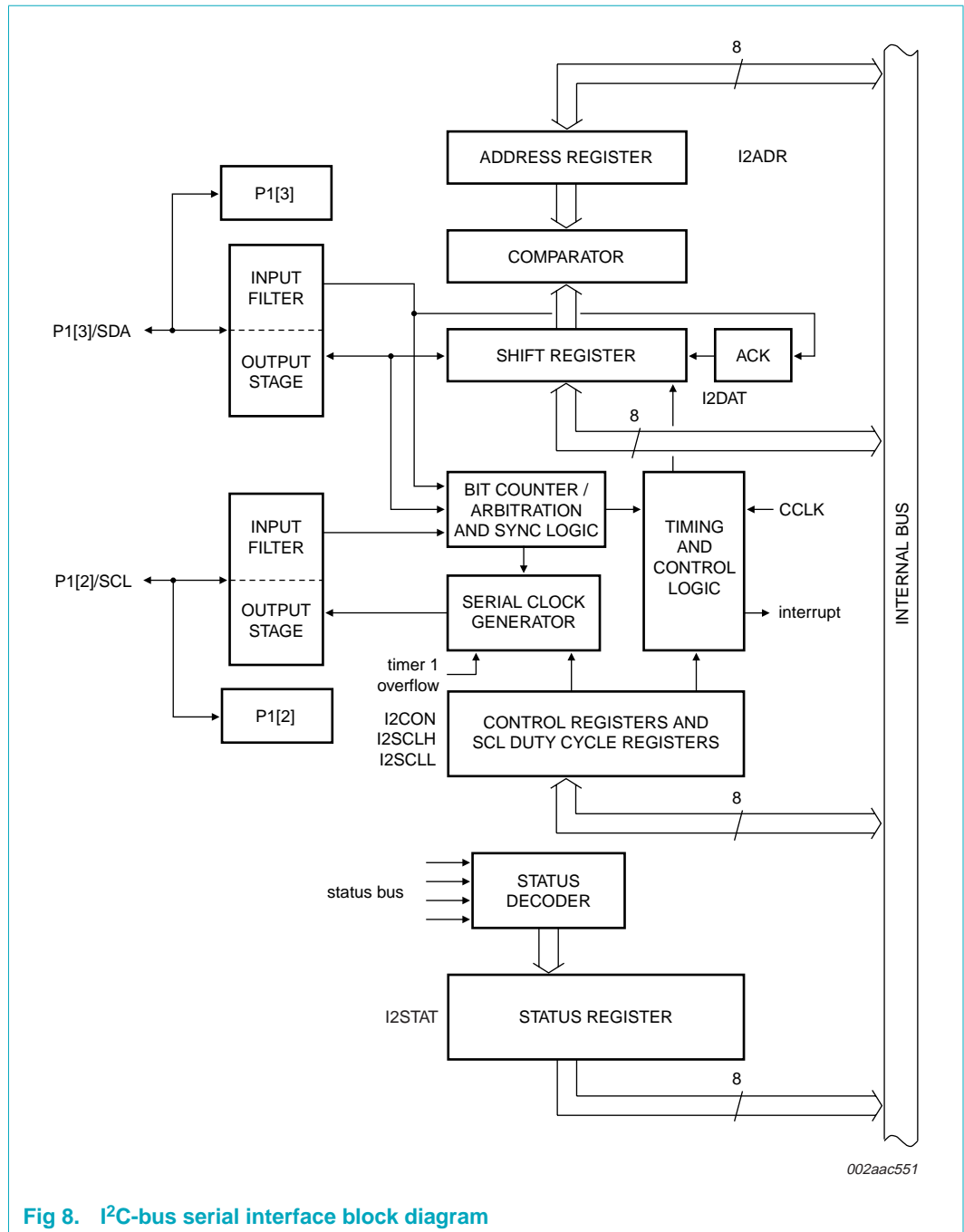


Fig 8. I<sup>2</sup>C-bus serial interface block diagram

7.21 SPI

The P89LPC9381 provides another high-speed serial communication interface—the SPI interface. SPI is a full-duplex, high-speed, synchronous communication bus with two operation modes: Master mode and Slave mode. Up to 3 Mbit/s can be supported in either Master or Slave mode. It has a Transfer Completion Flag and Write Collision Flag Protection.

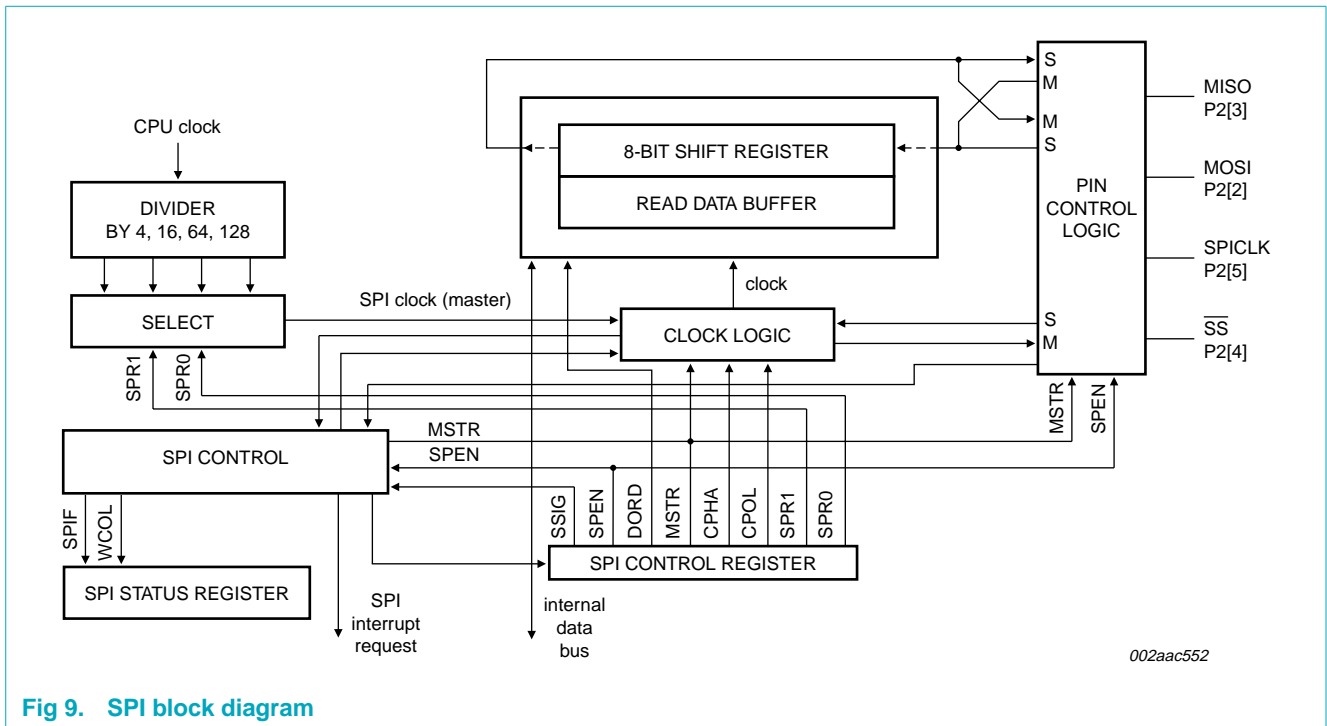


Fig 9. SPI block diagram

The SPI interface has four pins: SPICLK, MOSI, MISO and  $\overline{SS}$ :

- SPICLK, MOSI and MISO are typically tied together between two or more SPI devices. Data flows from master to slave on MOSI (Master Out Slave In) pin and flows from slave to master on MISO (Master In Slave Out) pin. The SPICLK signal is output in the master mode and is input in the slave mode. If the SPI system is disabled, i.e., SPEN (SPCTL.6) = 0 (reset value), these pins are configured for port functions.
- $\overline{SS}$  is the optional slave select pin. In a typical configuration, an SPI master asserts one of its port pins to select one SPI device as the current slave. An SPI slave device uses its  $\overline{SS}$  pin to determine whether it is selected.

Typical connections are shown in [Figure 10](#) through [Figure 12](#).

7.21.1 Typical SPI configurations

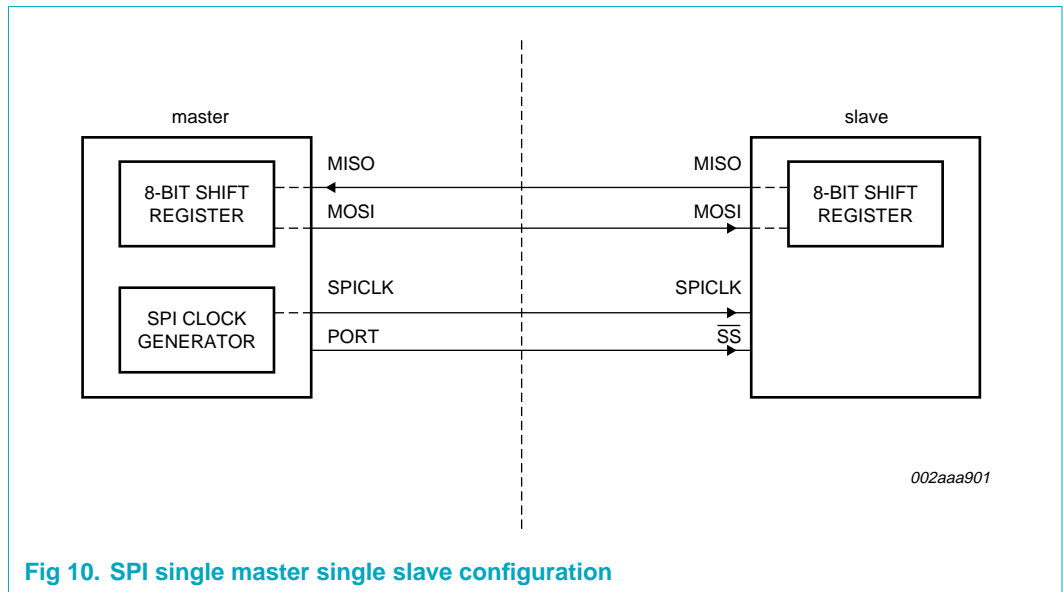


Fig 10. SPI single master single slave configuration

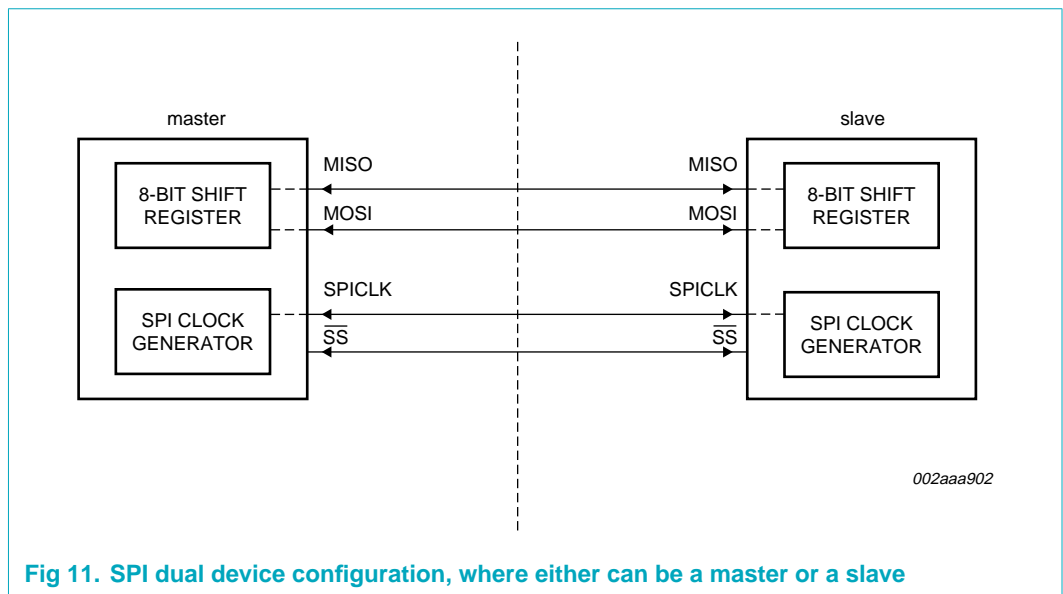
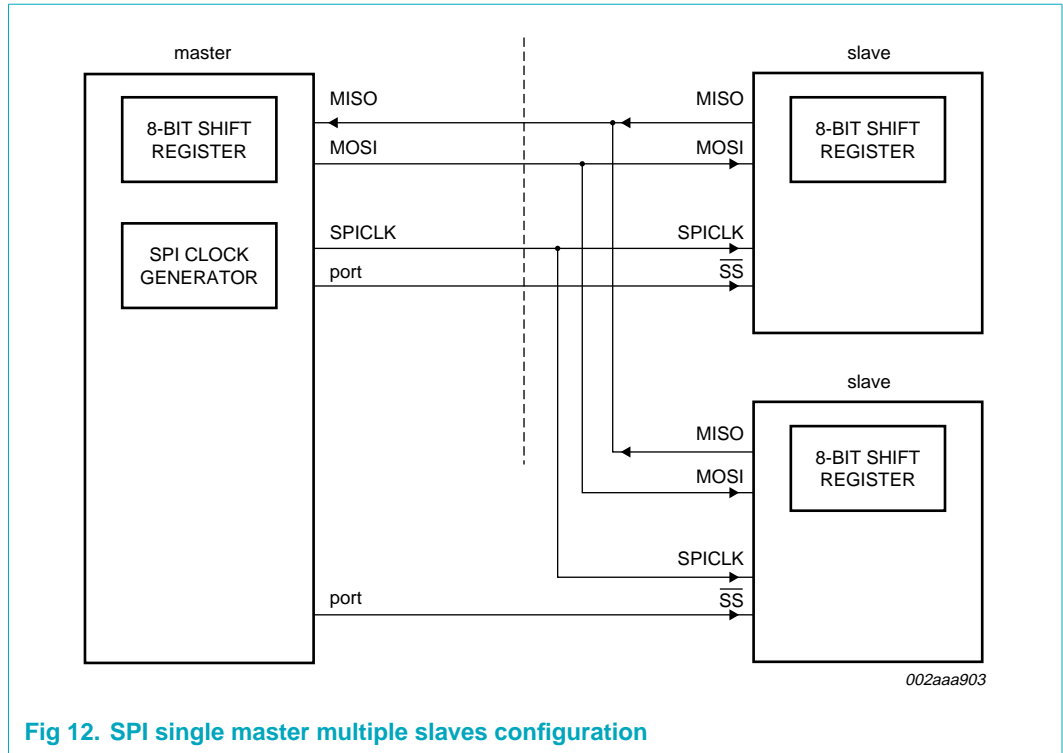


Fig 11. SPI dual device configuration, where either can be a master or a slave





## 7.22 Analog comparators

Two analog comparators are provided on the P89LPC9381. Input and output options allow use of the comparators in a number of different configurations. Comparator operation is such that the output is a logical one (which may be read in a register and/or routed to a pin) when the positive input (one of two selectable pins) is greater than the negative input (selectable from a pin or an internal reference voltage). Otherwise the output is a zero. Each comparator may be configured to cause an interrupt when the output value changes.

The overall connections to both comparators are shown in [Figure 13](#). The comparators function to  $V_{DD} = 2.4$  V.

When each comparator is first enabled, the comparator output and interrupt flag are not guaranteed to be stable for 10 microseconds. The corresponding comparator interrupt should not be enabled during that time, and the comparator interrupt flag must be cleared before the interrupt is enabled in order to prevent an immediate interrupt service.

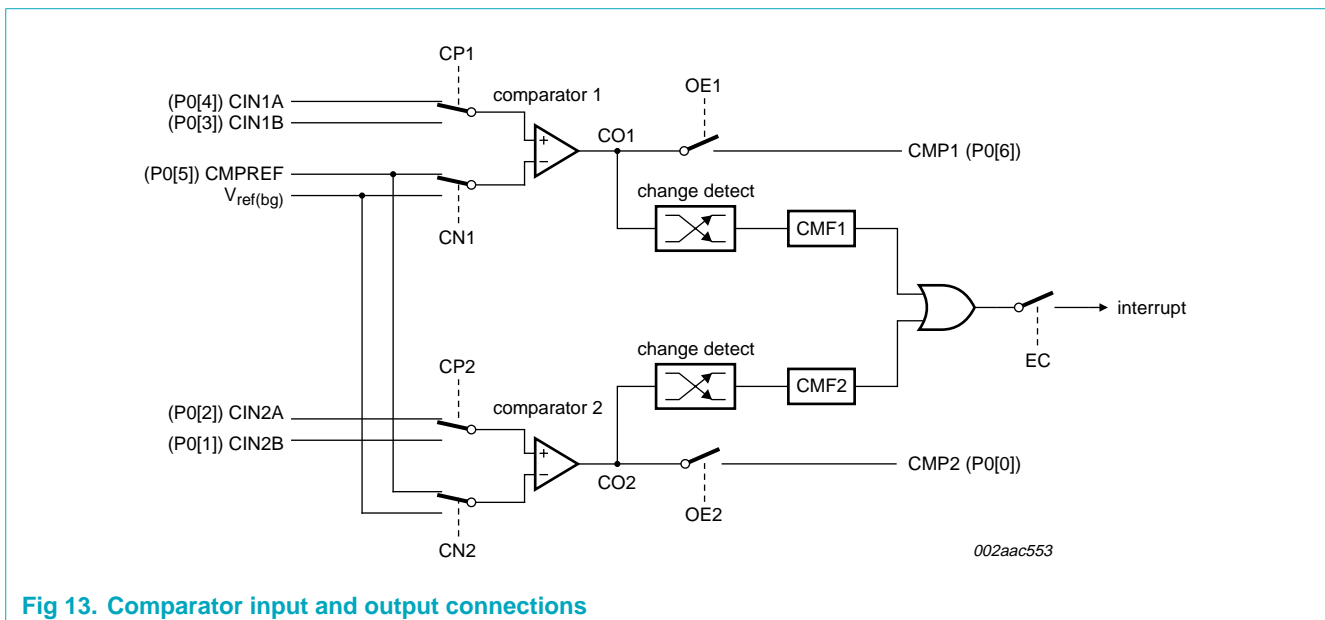


Fig 13. Comparator input and output connections

### 7.22.1 Internal reference voltage

An internal reference voltage generator may supply a default reference when a single comparator input pin is used. The value of the internal reference voltage, referred to as  $V_{ref(bg)}$ , is  $1.23\text{ V} \pm 10\%$ .

### 7.22.2 Comparator interrupt

Each comparator has an interrupt flag contained in its configuration register. This flag is set whenever the comparator output changes state. The flag may be polled by software or may be used to generate an interrupt. The two comparators use one common interrupt vector. If both comparators enable interrupts, after entering the interrupt service routine, the user needs to read the flags to determine which comparator caused the interrupt.

### 7.22.3 Comparators and power reduction modes

Either or both comparators may remain enabled when Power-down or Idle mode is activated, but both comparators are disabled automatically in Total Power-down mode.

If a comparator interrupt is enabled (except in Total Power-down mode), a change of the comparator output state will generate an interrupt and wake up the processor. If the comparator output to a pin is enabled, the pin should be configured in the push-pull mode in order to obtain fast switching times while in Power-down mode. The reason is that with the **oscillator** stopped, the temporary strong pull-up that normally occurs during switching on a quasi-bidirectional port pin does not take place.

Comparators consume power in Power-down and Idle modes, as well as in the normal operating mode. This fact should be taken into account when system power consumption is an issue. To minimize power consumption, the user can disable the comparators via PCONA.5, or put the device in Total Power-down mode.

### 7.23 KBI

The Keypad Interrupt function is intended primarily to allow a single interrupt to be generated when Port 0 is equal to or not equal to a certain pattern. This function can be used for bus address recognition or keypad recognition. The user can configure the port via SFRs for different tasks.

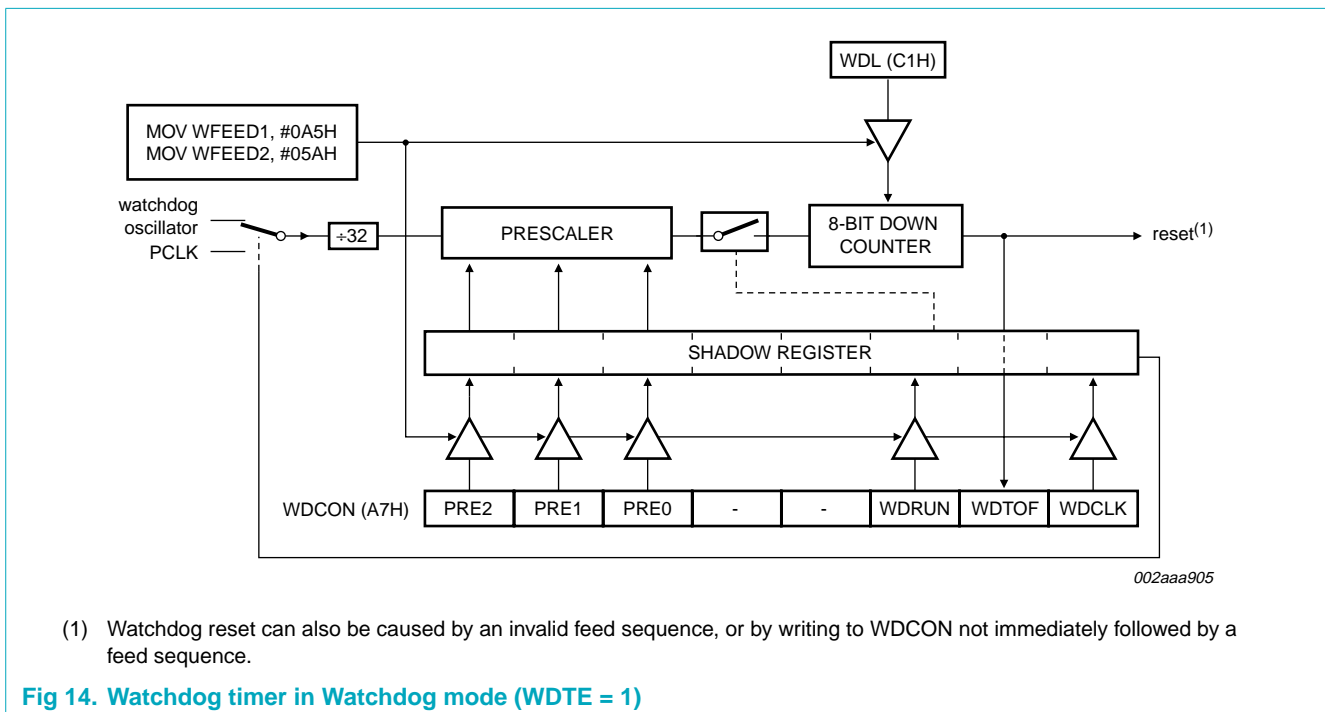
The Keypad Interrupt Mask Register (KBMASK) is used to define which input pins connected to Port 0 can trigger the interrupt. The Keypad Pattern Register (KBPATN) is used to define a pattern that is compared to the value of Port 0. The Keypad Interrupt Flag (KBIF) in the Keypad Interrupt Control Register (KBCON) is set when the condition is matched while the Keypad Interrupt function is active. An interrupt will be generated if enabled. The PATN\_SEL bit in the Keypad Interrupt Control Register (KBCON) is used to define equal or not-equal for the comparison.

In order to use the Keypad Interrupt as an original KBI function like in 87LPC76x series, the user needs to set KBPATN = 0FFH and PATN\_SEL = 1 (not equal), then any key connected to Port 0 which is enabled by the KBMASK register will cause the hardware to set KBIF and generate an interrupt if it has been enabled. The interrupt may be used to wake up the CPU from Idle or Power-down modes. This feature is particularly useful in handheld, battery-powered systems that need to carefully manage power consumption yet also need to be convenient to use.

In order to set the flag and cause an interrupt, the pattern on Port 0 must be held longer than six CCLKs.

## 7.24 Watchdog timer

The watchdog timer causes a system reset when it underflows as a result of a failure to feed the timer prior to the timer reaching its terminal count. It consists of a programmable 12-bit prescaler, and an 8-bit down counter. The down counter is decremented by a tap taken from the prescaler. The clock source for the prescaler is either the PCLK or the nominal 400 kHz Watchdog oscillator. The watchdog timer can only be reset by a power-on reset. When the watchdog feature is disabled, it can be used as an interval timer and may generate an interrupt. Figure 14 shows the watchdog timer in Watchdog mode. Feeding the watchdog requires a two-byte sequence. If PCLK is selected as the watchdog clock and the CPU is powered-down, the watchdog is disabled. The watchdog timer has a time-out period that ranges from a few  $\mu\text{s}$  to a few seconds. Please refer to the P89LPC9381 *User's Manual* for more details.



## 7.25 Additional features

### 7.25.1 Software reset

The SRST bit in AUXR1 gives software the opportunity to reset the processor completely, as if an external reset or watchdog reset had occurred. Care should be taken when writing to AUXR1 to avoid accidental software resets.

### 7.25.2 Dual data pointers

The dual Data Pointers (DPTR) provides two different Data Pointers to specify the address used with certain instructions. The DPS bit in the AUXR1 register selects one of the two Data Pointers. Bit 2 of AUXR1 is permanently wired as a logic 0 so that the DPS bit may be toggled (thereby switching Data Pointers) simply by incrementing the AUXR1 register, without the possibility of inadvertently altering other bits in the register.

## 7.26 Flash program memory

### 7.26.1 General description

The P89LPC9381 flash memory provides in-circuit electrical erasure and programming. The flash can be erased, read, and written as bytes. The Sector and Page Erase functions can erase any flash sector (1 kB) or page (64 B). The Chip Erase operation will erase the entire program memory. In-Circuit Programming using standard commercial programmers is available. In addition, IAP and byte-erase allows code memory to be used for non-volatile data storage. On-chip erase and write timing generation contribute to a user-friendly programming interface. The P89LPC9381 flash reliably stores memory contents even after 100000 erase and program cycles. The cell is designed to optimize the erase and programming mechanisms. The P89LPC9381 uses  $V_{DD}$  as the supply voltage to perform the Program/Erase algorithms.

### 7.26.2 Features

- Programming and erase over the full operating voltage range.
- Byte erase allows code memory to be used for data storage.
- Read/Programming/Erase using ISP/IAP/ICP.
- Internal fixed boot ROM, containing low-level IAP routines available to user code.
- Default loader providing In-System Programming via the serial port, located in upper end of user program memory.
- Boot vector allows user-provided flash loader code to reside anywhere in the flash memory space, providing flexibility to the user.
- Any flash program/erase operation in 2 ms.
- Programming with industry-standard commercial programmers.
- Programmable security for the code in the flash for each sector.
- 400000 typical erase/program cycles for each byte.
- 20 year minimum data retention.

### 7.26.3 Flash organization

The program memory consists of four 1 kB sectors on the P89LPC9381 devices. Each sector can be further divided into 64 B pages. In addition to sector erase, page erase, and byte erase, a 64 B page register is included which allows from 1 B to 64 B of a given page to be programmed at the same time, substantially reducing overall programming time.

### 7.26.4 Using flash as data storage

The flash code memory array of this device supports individual byte erasing and programming. Any byte in the code memory array may be read using the MOV<sub>C</sub> instruction, provided that the sector containing the byte has not been secured (a MOV<sub>C</sub> instruction is not allowed to read code memory contents of a secured sector). Thus any byte in a non-secured sector may be used for non-volatile data storage.

### 7.26.5 Flash programming and erasing

Four different methods of erasing or programming of the flash are available. The flash may be programmed or erased in the end-user application (IAP) under control of the application's firmware. Another option is to use the ICP mechanism. This ICP system

provides for programming through a serial clock/serial data interface. As shipped from the factory, the upper 512 B of user code space contains a serial ISP routine allowing for the device to be programmed in circuit through the serial port. The flash may also be programmed or erased using a commercially available EPROM programmer which supports this device. This device does not provide for direct verification of code memory contents. Instead, this device provides a 32-bit CRC result on either a sector or the entire user code space.

#### 7.26.6 ICP

In-Circuit Programming is performed without removing the microcontroller from the system. The In-Circuit Programming facility consists of internal hardware resources to facilitate remote programming of the P89LPC9381 through a two-wire serial interface. The Philips In-Circuit Programming facility has made in-circuit programming in an embedded application—using commercially available programmers—possible with a minimum of additional expense in components and circuit board area. The ICP function uses five pins. Only a small connector needs to be available to interface your application to a commercial programmer in order to use this feature. Additional details may be found in the P89LPC9381 *User's Manual*.

#### 7.26.7 IAP

In-Application Programming is performed in the application under the control of the microcontroller's firmware. The IAP facility consists of internal hardware resources to facilitate programming and erasing. The Philips In-Application Programming has made in-application programming in an embedded application possible without additional components. Two methods are available to accomplish IAP. A set of predefined IAP functions are provided in a Boot ROM and can be called through a common interface, PGM\_MTP. Several IAP calls are available for use by an application program to permit selective erasing and programming of flash sectors, pages, security bits, configuration bytes, and device ID. These functions are selected by setting up the microcontroller's registers before making a call to PGM\_MTP at FF03H. The Boot ROM occupies the program memory space at the top of the address space from FF00 to FEFF hex, thereby not conflicting with the user program memory space.

In addition, IAP operations can be accomplished through the use of four SFRs consisting of a control/status register, a data register, and two address registers. Additional details may be found in the P89LPC9381 *User's Manual*.

#### 7.26.8 ISP

In-System Programming is performed without removing the microcontroller from the system. The In-System Programming facility consists of a series of internal hardware resources coupled with internal firmware to facilitate remote programming of the P89LPC9381 through the serial port. This firmware is provided by Philips and embedded within each P89LPC9381 device. The Philips In-System Programming facility has made in-system programming in an embedded application possible with a minimum of additional expense in components and circuit board area. The ISP function uses five pins ( $V_{DD}$ ,  $V_{SS}$ , TXD, RXD, and  $\overline{RST}$ ). Only a small connector needs to be available to interface your application to an external circuit in order to use this feature.

### 7.26.9 Power-on reset code execution

The P89LPC9381 contains two special flash elements: the Boot Vector and the Boot Status Bit. Following reset, the P89LPC9381 examines the contents of the Boot Status Bit. If the Boot Status Bit is set to zero, power-up execution starts at location 0000H, which is the normal start address of the user's application code. When the Boot Status Bit is set to a value other than zero, the contents of the Boot Vector is used as the high byte of the execution address and the low byte is set to 00H.

[Table 8](#) shows the factory default Boot Vector setting for this device. A factory-provided bootloader is pre-programmed into the address space indicated and uses the indicated boot loader entry point to perform ISP functions. This code can be erased by the user. **Users who wish to use this loader should take precautions to avoid erasing the 1 kB sector that contains this boot loader. Instead, the page erase function can be used to erase the first eight 64 B pages located in this sector.** A custom bootloader can be written with the Boot Vector set to the custom boot loader, if desired.

**Table 8. Default Boot Vector values and ISP entry points**

Device	Default boot vector	Default bootloader entry point	Default bootloader code range	1 kB sector range
P89LPC9381	0FH	0F00H	0E00H to 0FFFH	0C00H to 0FFFH

### 7.26.10 Hardware activation of the bootloader

The bootloader can also be executed by forcing the device into ISP mode during a power-on sequence (see the P89LPC9381 *User's Manual* for specific information). This has the same effect as having a non-zero status byte. This allows an application to be built that will normally execute user code but can be manually forced into ISP operation. If the factory default setting for the Boot Vector (0FH) is changed, it will no longer point to the factory pre-programmed ISP bootloader code. After programming the flash, the status byte should be programmed to zero in order to allow execution of the user's application code beginning at address 0000H.

### 7.27 User configuration bytes

Some user-configurable features of the P89LPC9381 must be defined at power-up and therefore cannot be set by the program after start of execution. These features are configured through the use of the flash byte UCFG1. Please see the P89LPC9381 *User's Manual* for additional details.

### 7.28 User sector security bytes

There are four User Sector Security Bytes on the P89LPC9381. Each byte corresponds to one sector. Please see the P89LPC9381 *User's Manual* for additional details.

## 8. ADC

### 8.1 General description

The P89LPC9381 has a 10-bit, 8-channel multiplexed successive approximation analog-to-digital converter module. A block diagram of the ADC is shown in [Figure 15](#). The ADC consists of an 8-input multiplexer which feeds a sample-and-hold circuit

providing an input signal to one of two comparator inputs. The control logic in combination with the SAR drives a digital-to-analog converter which provides the other input to the comparator. The output of the comparator is fed to the SAR.

## 8.2 Features

- 10-bit, 8-channel multiplexed input, successive approximation ADC.
- Eight result register pairs.
- Six operating modes
  - ◆ Fixed channel, single conversion mode
  - ◆ Fixed channel, continuous conversion mode
  - ◆ Auto scan, single conversion mode
  - ◆ Auto scan, continuous conversion mode
  - ◆ Dual channel, continuous conversion mode
  - ◆ Single step mode
- Three conversion start modes
  - ◆ Timer triggered start
  - ◆ Start immediately
  - ◆ Edge triggered
- 10-bit conversion time of 4  $\mu$ s at an ADC clock of 9 MHz
- Interrupt or polled operation
- High and Low Boundary limits interrupt; selectable in or out-of-range
- Clock divider
- Power-down mode

## 8.3 Block diagram

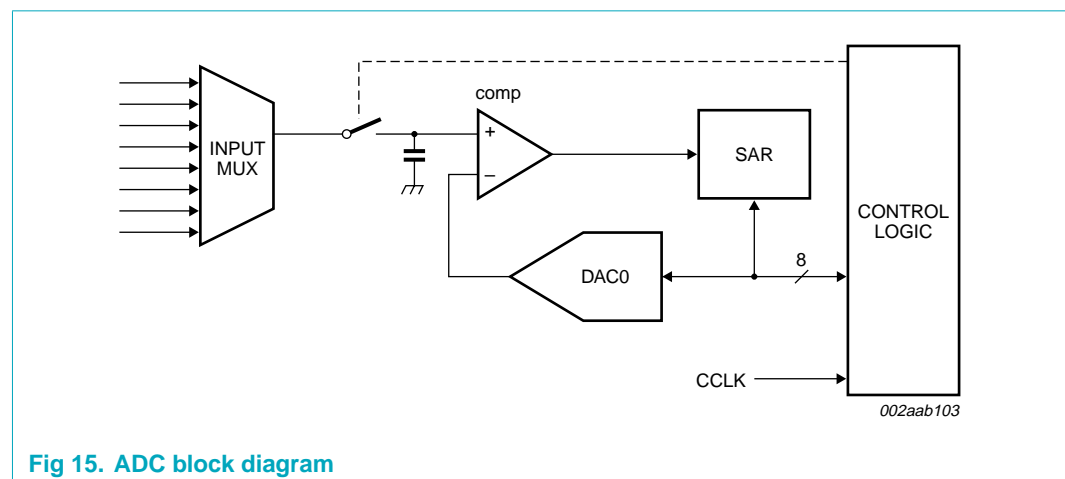


Fig 15. ADC block diagram



## 8.4 ADC operating modes

### 8.4.1 Fixed channel, single conversion mode

A single input channel can be selected for conversion. A single conversion will be performed and the result placed in the result register pair which corresponds to the selected input channel. An interrupt, if enabled, will be generated after the conversion completes.

### 8.4.2 Fixed channel, continuous conversion mode

A single input channel can be selected for continuous conversion. The results of the conversions will be sequentially placed in the eight result register pairs. The user may select whether an interrupt can be generated after every four or every eight conversions. Additional conversion results will again cycle through the result register pairs, overwriting the previous results. Continuous conversions continue until terminated by the user.

### 8.4.3 Auto scan, single conversion mode

Any combination of the eight input channels can be selected for conversion. A single conversion of each selected input will be performed and the result placed in the result register pair which corresponds to the selected input channel. The user may select whether an interrupt, if enabled, will be generated after either the first four conversions have occurred or all selected channels have been converted. If the user selects to generate an interrupt after the four input channels have been converted, a second interrupt will be generated after the remaining input channels have been converted. If only a single channel is selected this is equivalent to single channel, single conversion mode.

### 8.4.4 Auto scan, continuous conversion mode

Any combination of the eight input channels can be selected for conversion. A conversion of each selected input will be performed and the result placed in the result register pair which corresponds to the selected input channel. The user may select whether an interrupt, if enabled, will be generated after either the first four conversions have occurred or all selected channels have been converted. If the user selects to generate an interrupt after the four input channels have been converted, a second interrupt will be generated after the remaining input channels have been converted. After all selected channels have been converted, the process will repeat starting with the first selected channel. Additional conversion results will again cycle through the eight result register pairs, overwriting the previous results. Continuous conversions continue until terminated by the user.

### 8.4.5 Dual channel, continuous conversion mode

This is a variation of the auto scan continuous conversion mode where conversion occurs on two user-selectable inputs. The result of the conversion of the first channel is placed in the result register pair, AD0DAT0R and AD0DAT0L. The result of the conversion of the second channel is placed in result register pair, AD0DAT1R and AD0DAT1L. The first channel is again converted and its result stored in AD0DAT2R and AD0DAT2L. The second channel is again converted and its result placed in AD0DAT3R and AD0DAT3L, etc. An interrupt is generated, if enabled, after every set of four or eight conversions (user selectable).

#### 8.4.6 Single step mode

This special mode allows 'single-stepping' in an auto scan conversion mode. Any combination of the eight input channels can be selected for conversion. After each channel is converted, an interrupt is generated, if enabled, and the ADC waits for the next start condition. May be used with any of the start modes.

### 8.5 Conversion start modes

#### 8.5.1 Timer triggered start

The ADC is started by the overflow of Timer 0. Once a conversion has started, additional Timer 0 triggers are ignored until the conversion has completed. The Timer triggered start mode is available in all ADC operating modes.

#### 8.5.2 Start immediately

Programming this mode immediately starts a conversion. This start mode is available in all ADC operating modes.

#### 8.5.3 Edge triggered

The ADC is started by rising or falling edge of P1[4]. Once a conversion has started, additional edge triggers are ignored until the conversion has completed. The edge triggered start mode is available in all ADC operating modes.

### 8.6 Boundary limits interrupt

The ADC has both a high and low boundary limit register. The user may select whether an interrupt is generated when the conversion result is within (or equal to) the high and low boundary limits or when the conversion result is outside the boundary limits. An interrupt will be generated, if enabled, if the result meets the selected interrupt criteria. The boundary limit may be disabled by clearing the boundary limit interrupt enable.

An early detection mechanism exists when the interrupt criteria has been selected to be outside the boundary limits. In this case, after the four MSBs have been converted, these four bits are compared with the four MSBs of the boundary high and low registers. If the four MSBs of the conversion meet the interrupt criteria (i.e., outside the boundary limits) an interrupt will be generated, if enabled. If the four MSBs do not meet the interrupt criteria, the boundary limits will again be compared after all 8 MSBs have been converted. A boundary status register (BNDSTAO) flags the channels which caused a boundary interrupt.

### 8.7 Clock divider

The ADC requires that its internal clock source be in the range of 500 kHz to 3 MHz to maintain accuracy. A programmable clock divider that divides the clock from 1 to 8 is provided for this purpose.

### 8.8 Power-down and Idle mode

In Idle mode the ADC, if enabled, will continue to function and can cause the device to exit Idle mode when the conversion is completed if the ADC interrupt is enabled. In Power-down mode or Total Power-down mode, the ADC does not function. If the ADC is enabled, it will consume power. Power can be reduced by disabling the ADC.

## 9. Limiting values

**Table 9. Limiting values**

In accordance with the Absolute Maximum Rating System (IEC 60134).<sup>[1]</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
$T_{amb(bias)}$	bias ambient temperature		-55	+125	°C
$T_{stg}$	storage temperature		-65	+150	°C
$I_{OH(I/O)}$	HIGH-state output current per input/output pin		-	20	mA
$I_{OL(I/O)}$	LOW-state output current per input/output pin		-	20	mA
$I_{I/Otot(max)}$	maximum total input/output current		-	100	mA
$V_n$	voltage on any other pin	except $V_{SS}$ , with respect to $V_{DD}$	-	3.5	V
$P_{tot(pack)}$	total power dissipation (per package)	based on package heat transfer, not device power consumption	-	1.5	W

[1] The following applies to [Table 9](#):

- a) This product includes circuitry specifically designed for the protection of its internal devices from the damaging effects of excessive static charge. Nonetheless, it is suggested that conventional precautions be taken to avoid applying greater than the rated maximum.
- b) Parameters are valid over operating temperature range unless otherwise specified. All voltages are with respect to  $V_{SS}$  unless otherwise noted.

## 10. Static characteristics

**Table 10. Static characteristics**

$V_{DD} = 2.4\text{ V to }3.6\text{ V}$  unless otherwise specified.

$T_{amb} = -40\text{ °C to }+85\text{ °C}$  for industrial applications, unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ <sup>[1]</sup>	Max	Unit
$I_{DD(oper)}$	operating supply current	$V_{DD} = 3.6\text{ V}; f_{osc} = 12\text{ MHz}$	[2] -	14	23	mA
$I_{DD(idle)}$	Idle mode supply current	$V_{DD} = 3.6\text{ V}; f_{osc} = 12\text{ MHz}$	[2] -	5	7	mA
$I_{DD(pd)}$	Power-down mode supply current	$V_{DD} = 3.6\text{ V}$ ; voltage comparators powered-down	[2] -	55	80	$\mu\text{A}$
$I_{DD(tpd)}$	total Power-down mode supply current	$V_{DD} = 3.6\text{ V}$	[3] -	0.5	5	$\mu\text{A}$
$(dV/dt)_r$	rise rate	of $V_{DD}$	-	-	2	mV/ $\mu\text{s}$
$(dV/dt)_f$	fall rate	of $V_{DD}$	-	-	50	mV/ $\mu\text{s}$
$V_{DDR}$	data retention supply voltage		1.5	-	-	V
$V_{th(HL)}$	HIGH-LOW threshold voltage	except SCL, SDA	$0.22V_{DD}$	$0.4V_{DD}$	-	V
$V_{IL}$	LOW-level input voltage	SCL, SDA only	-0.5	-	$+0.3V_{DD}$	V
$V_{th(LH)}$	LOW-HIGH threshold voltage	except SCL, SDA	-	$0.6V_{DD}$	$0.7V_{DD}$	V
$V_{IH}$	HIGH-level input voltage	SCL, SDA only	$0.7V_{DD}$	-	5.5	V
$V_{hys}$	hysteresis voltage	port 1	-	$0.2V_{DD}$	-	V
$V_{OL}$	LOW-level output voltage	$I_{OL} = 20\text{ mA}$ ; $V_{DD} = 2.4\text{ V to }3.6\text{ V}$ ; all ports, all modes except high-Z	[4] -	0.6	1.0	V
		$I_{OL} = 3.2\text{ mA}$ ; $V_{DD} = 2.4\text{ V to }3.6\text{ V}$ ; all ports, all modes except high-Z	[4] -	0.2	0.3	V
$V_{OH}$	HIGH-level output voltage	$I_{OH} = -20\text{ }\mu\text{A}$ ; $V_{DD} = 2.4\text{ V to }3.6\text{ V}$ ; quasi-bidirectional mode, all ports	$V_{DD} - 0.3$	$V_{DD} - 0.2$	-	V
		$I_{OH} = -3.2\text{ mA}$ ; $V_{DD} = 2.4\text{ V to }3.6\text{ V}$ ; push-pull mode, all ports	$V_{DD} - 0.7$	$V_{DD} - 0.4$	-	V
		$I_{OH} = -20\text{ mA}$ ; $V_{DD} = 2.4\text{ V to }3.6\text{ V}$ ; push-pull mode, all ports	$V_{DD} - 1.0$	-	-	V
$V_{xtal}$	crystal voltage	on XTAL1, XTAL2 pins; with respect to $V_{SS}$	-0.5	-	+4.0	V
$V_n$	voltage on any other pin	except XTAL1, XTAL2, $V_{DD}$ ; with respect to $V_{SS}$	-0.5	-	+5.5	V
$C_{iss}$	input capacitance		[5] -	-	15	pF
$I_{IL}$	LOW-level input current	$V_I = 0.4\text{ V}$	[6] -	-	-80	$\mu\text{A}$
$I_{LI}$	input leakage current	$V_I = V_{IL}, V_{IH}, V_{th(HL)}$	[7] -	-	$\pm 1$	$\mu\text{A}$
$I_{THL}$	HIGH-LOW transition current	$V_I = 1.5\text{ V}$ at $V_{DD} = 3.6\text{ V}$	[8] -30	-	-450	$\mu\text{A}$

**Table 10. Static characteristics ...continued** $V_{DD} = 2.4\text{ V to }3.6\text{ V unless otherwise specified.}$  $T_{amb} = -40\text{ °C to }+85\text{ °C for industrial applications, unless otherwise specified.}$ 

Symbol	Parameter	Conditions	Min	Typ <sup>[1]</sup>	Max	Unit
$R_{RST\_N(int)}$	internal pull-up resistance on pin $\overline{RST}$		10	-	30	$k\Omega$
$V_{bo}$	brownout trip voltage	$2.4\text{ V} < V_{DD} < 3.6\text{ V}$ ; with $BOV = 1$ , $BOPD = 0$	2.40	-	2.70	V
$V_{ref(bg)}$	band gap reference voltage		1.19	1.23	1.27	V
$TC_{bg}$	band gap temperature coefficient		-	10	20	ppm/°C

- [1] Typical ratings are not guaranteed. The values listed are at room temperature,  $V_{DD} = 3\text{ V}$ .
- [2] The  $I_{DD(oper)}$ ,  $I_{DD(idle)}$ , and  $I_{DD(pd)}$  specifications are measured using an external clock with the following functions disabled: comparators, real-time clock, and watchdog timer.
- [3] The  $I_{DD(tpd)}$  specification is measured using an external clock with the following functions disabled: comparators, real-time clock, brownout detect, and watchdog timer.
- [4] See [Section 9 "Limiting values" on page 43](#) for steady state (non-transient) limits on  $I_{OL}$  or  $I_{OH}$ . If  $I_{OL}/I_{OH}$  exceeds the test condition,  $V_{OL}/V_{OH}$  may exceed the related specification.
- [5] Pin capacitance is characterized but not tested.
- [6] Measured with port in quasi-bidirectional mode.
- [7] Measured with port in high-impedance mode.
- [8] Port pins source a transition current when used in quasi-bidirectional mode and externally driven from logic 1 to logic 0. This current is highest when  $V_I$  is approximately 2 V.

## 11. Dynamic characteristics

**Table 11. Dynamic characteristics (12 MHz)**
 $V_{DD} = 2.4\text{ V to }3.6\text{ V}$ , unless otherwise specified.

 $T_{amb} = -40\text{ °C to }+85\text{ °C}$  for industrial applications, unless otherwise specified.<sup>[1][2]</sup>

Symbol	Parameter	Conditions	Variable clock		$f_{osc} = 12\text{ MHz}$		Unit
			Min	Max	Min	Max	
$f_{osc(RC)}$	internal RC oscillator frequency		7.189	7.557	7.189	7.557	MHz
$f_{osc(WD)}$	internal watchdog oscillator frequency		320	520	320	520	kHz
$f_{osc}$	oscillator frequency		0	12	-	-	MHz
$T_{cy(clk)}$	clock cycle time	see <a href="#">Figure 17</a>	83	-	-	-	ns
$f_{CLKLP}$	low-power select clock frequency	on pin CLKLP	0	8	-	-	MHz
<b>Glitch filter</b>							
$t_{gr}$	glitch rejection time	P1[5]/ $\overline{RST}$ pin	-	50	-	50	ns
		any pin except P1[5]/ $\overline{RST}$	-	15	-	15	ns
$t_{sa}$	signal acceptance time	P1[5]/ $\overline{RST}$ pin	125	-	125	-	ns
		any pin except P1[5]/ $\overline{RST}$	50	-	50	-	ns
<b>External clock</b>							
$t_{CHCX}$	clock HIGH time	see <a href="#">Figure 17</a>	33	$T_{cy(clk)} - t_{CLCX}$	33	-	ns
$t_{CLCX}$	clock LOW time	see <a href="#">Figure 17</a>	33	$T_{cy(clk)} - t_{CHCX}$	33	-	ns
$t_{CLCH}$	clock rise time	see <a href="#">Figure 17</a>	-	8	-	8	ns
$t_{CHCL}$	clock fall time	see <a href="#">Figure 17</a>	-	8	-	8	ns
<b>Shift register (UART mode 0)</b>							
$T_{XLXL}$	serial port clock cycle time	see <a href="#">Figure 16</a>	$16T_{cy(clk)}$	-	1333	-	ns
$t_{QVXH}$	output data setup to clock rising edge time	see <a href="#">Figure 16</a>	$13T_{cy(clk)}$	-	1083	-	ns
$t_{XHQX}$	output data hold after clock rising edge time	see <a href="#">Figure 16</a>	-	$T_{cy(clk)} + 20$	-	103	ns
$t_{XHDX}$	input data hold after clock rising edge time	see <a href="#">Figure 16</a>	-	0	-	0	ns
$t_{XHDV}$	input data valid to clock rising edge time	see <a href="#">Figure 16</a>	150	-	150	-	ns
<b>SPI interface</b>							
$f_{SPI}$	SPI operating frequency						
	slave		0	$CCLK/6$	0	2.0	MHz
	master		-	$CCLK/4$	-	3.0	MHz
$T_{SPICYC}$	SPI cycle time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>					
	slave		$6/CCLK$	-	500	-	ns
	master		$4/CCLK$	-	333	-	ns
$t_{SPILEAD}$	SPI enable lead time	see <a href="#">Figure 20</a> and <a href="#">21</a>					
	slave		250	-	250	-	ns

**Table 11. Dynamic characteristics (12 MHz) ...continued**

$V_{DD} = 2.4\text{ V to }3.6\text{ V}$ , unless otherwise specified.

$T_{amb} = -40\text{ }^{\circ}\text{C to }+85\text{ }^{\circ}\text{C}$  for industrial applications, unless otherwise specified.<sup>[1][2]</sup>

Symbol	Parameter	Conditions	Variable clock		$f_{osc} = 12\text{ MHz}$		Unit
			Min	Max	Min	Max	
tSPILAG	SPI enable lag time	see <a href="#">Figure 20</a> and <a href="#">21</a>	250	-	250	-	ns
	slave						
tSPICLK <sub>H</sub>	SPICLK HIGH time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	$\frac{2}{CCLK}$	-	165	-	ns
	master						
	slave						
tSPICLK <sub>L</sub>	SPICLK LOW time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	$\frac{2}{CCLK}$	-	165	-	ns
	master						
	slave						
tSPID <sub>SU</sub>	SPI data set-up time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	100	-	100	-	ns
tSPID <sub>H</sub>	SPI data hold time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	100	-	100	-	ns
tSPIA	SPI access time	see <a href="#">Figure 20</a> and <a href="#">21</a>	0	120	0	120	ns
	slave						
tSPID <sub>IS</sub>	SPI disable time	see <a href="#">Figure 20</a> and <a href="#">21</a>	0	240	-	240	ns
	slave						
tSPID <sub>V</sub>	SPI enable to output data valid time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	-	240	-	240	ns
	slave						
	master						
tSPIO <sub>H</sub>	SPI output data hold time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	0	-	0	-	ns
tSPIR	SPI rise time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	-	100	-	100	ns
	SPI outputs (SPICLK, MOSI, MISO)						
	SPI inputs (SPICLK, MOSI, MISO, SS)						
tSPI <sub>F</sub>	SPI fall time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	-	100	-	100	ns
	SPI outputs (SPICLK, MOSI, MISO)						
	SPI inputs (SPICLK, MOSI, MISO, SS)						

[1] Parameters are valid over operating temperature range unless otherwise specified.

[2] Parts are tested to 2 MHz, but are guaranteed to operate down to 0 Hz.

**Table 12. Dynamic characteristics (18 MHz)** $V_{DD} = 3.0\text{ V to }3.6\text{ V}$ , unless otherwise specified. $T_{amb} = -40\text{ °C to }+85\text{ °C}$  for industrial applications, unless otherwise specified.<sup>[1][2]</sup>

Symbol	Parameter	Conditions	Variable clock		$f_{osc} = 18\text{ MHz}$		Unit
			Min	Max	Min	Max	
$f_{osc(RC)}$	internal RC oscillator frequency		7.189	7.557	7.189	7.557	MHz
$f_{osc(WD)}$	internal watchdog oscillator frequency		320	520	320	520	kHz
$f_{osc}$	oscillator frequency		0	18	-	-	MHz
$T_{cy(clk)}$	clock cycle time	see <a href="#">Figure 17</a>	55	-	-	-	ns
$f_{CLKLP}$	low-power select clock frequency		0	8	-	-	MHz

**Glitch filter**

$t_{gr}$	glitch rejection time	P1[5]/RST pin	-	50	-	50	ns
		any pin except P1[5]/RST	-	15	-	15	ns
$t_{sa}$	signal acceptance time	P1[5]/RST pin	125	-	125	-	ns
		any pin except P1[5]/RST	50	-	50	-	ns

**External clock**

$t_{CHCX}$	clock HIGH time	see <a href="#">Figure 17</a>	22	$T_{cy(clk)} - t_{CLCX}$	22	-	ns
$t_{CLCX}$	clock LOW time	see <a href="#">Figure 17</a>	22	$T_{cy(clk)} - t_{CHCX}$	22	-	ns
$t_{CLCH}$	clock rise time	see <a href="#">Figure 17</a>	-	5	-	5	ns
$t_{CHCL}$	clock fall time	see <a href="#">Figure 17</a>	-	5	-	5	ns

**Shift register (UART mode 0)**

$T_{XLXL}$	serial port clock cycle time	see <a href="#">Figure 16</a>	$16T_{cy(clk)}$	-	888	-	ns
$t_{QVXH}$	output data setup to clock rising edge time	see <a href="#">Figure 16</a>	$13T_{cy(clk)}$	-	722	-	ns
$t_{XHQX}$	output data hold after clock rising edge time	see <a href="#">Figure 16</a>	-	$T_{cy(clk)} + 20$	-	75	ns
$t_{XHDX}$	input data hold after clock rising edge time	see <a href="#">Figure 16</a>	-	0	-	0	ns
$t_{XHDX}$	input data valid to clock rising edge time	see <a href="#">Figure 16</a>	150	-	150	-	ns

**SPI interface**

$f_{SPI}$	SPI operating frequency						
	slave		0	$CCLK/6$	0	3.0	MHz
	master		-	$CCLK/4$	-	4.5	MHz
$T_{SPICYC}$	SPI cycle time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> , <a href="#">21</a>					
	slave		$6/CCLK$	-	333	-	ns
	master		$4/CCLK$	-	222	-	ns
$t_{SPILEAD}$	SPI enable lead time	see <a href="#">Figure 20</a> , <a href="#">21</a>					
	slave		250	-	250	-	ns
$t_{SPILAG}$	SPI enable lag time	see <a href="#">Figure 20</a> , <a href="#">21</a>					
	slave		250	-	250	-	ns



**Table 12. Dynamic characteristics (18 MHz) ...continued** $V_{DD} = 3.0\text{ V to }3.6\text{ V}$ , unless otherwise specified. $T_{amb} = -40\text{ °C to }+85\text{ °C}$  for industrial applications, unless otherwise specified.<sup>[1][2]</sup>

Symbol	Parameter	Conditions	Variable clock		$f_{osc} = 18\text{ MHz}$		Unit
			Min	Max	Min	Max	
t <sub>SPICLK<sup>H</sup></sub>	SPICLK HIGH time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>					
	master		$\frac{2}{CCLK}$	-	111	-	ns
	slave		$\frac{3}{CCLK}$	-	167	-	ns
t <sub>SPICLK<sup>L</sup></sub>	SPICLK LOW time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>					
	master		$\frac{2}{CCLK}$	-	111	-	ns
	slave		$\frac{3}{CCLK}$	-	167	-	ns
t <sub>SPID<sup>SU</sup></sub>	SPI data set-up time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	100	-	100	-	ns
t <sub>SPID<sup>H</sup></sub>	SPI data hold time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	100	-	100	-	ns
t <sub>SPIA</sub>	SPI access time	see <a href="#">Figure 20</a> and <a href="#">21</a>	0	80	0	80	ns
	slave						
t <sub>SPID<sup>IS</sup></sub>	SPI disable time	see <a href="#">Figure 20</a> and <a href="#">21</a>	0	160	-	160	ns
	slave						
t <sub>SPID<sup>V</sup></sub>	SPI enable to output data valid time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>					
	slave		-	160	-	160	ns
	master		-	111	-	111	ns
t <sub>SPIO<sup>H</sup></sub>	SPI output data hold time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>	0	-	0	-	ns
t <sub>SPIR</sub>	SPI rise time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>					
	SPI outputs (SPICLK, MOSI, MISO)		-	100	-	100	ns
	SPI inputs (SPICLK, MOSI, MISO, $\overline{SS}$ )		-	2000	-	2000	ns
t <sub>SPIF</sub>	SPI fall time	see <a href="#">Figure 18</a> , <a href="#">19</a> , <a href="#">20</a> and <a href="#">21</a>					
	SPI outputs (SPICLK, MOSI, MISO)		-	100	-	100	ns
	SPI inputs (SPICLK, MOSI, MISO, $\overline{SS}$ )		-	2000	-	2000	ns

[1] Parameters are valid over operating temperature range unless otherwise specified.

[2] Parts are tested to 2 MHz, but are guaranteed to operate down to 0 Hz.

11.1 Waveforms

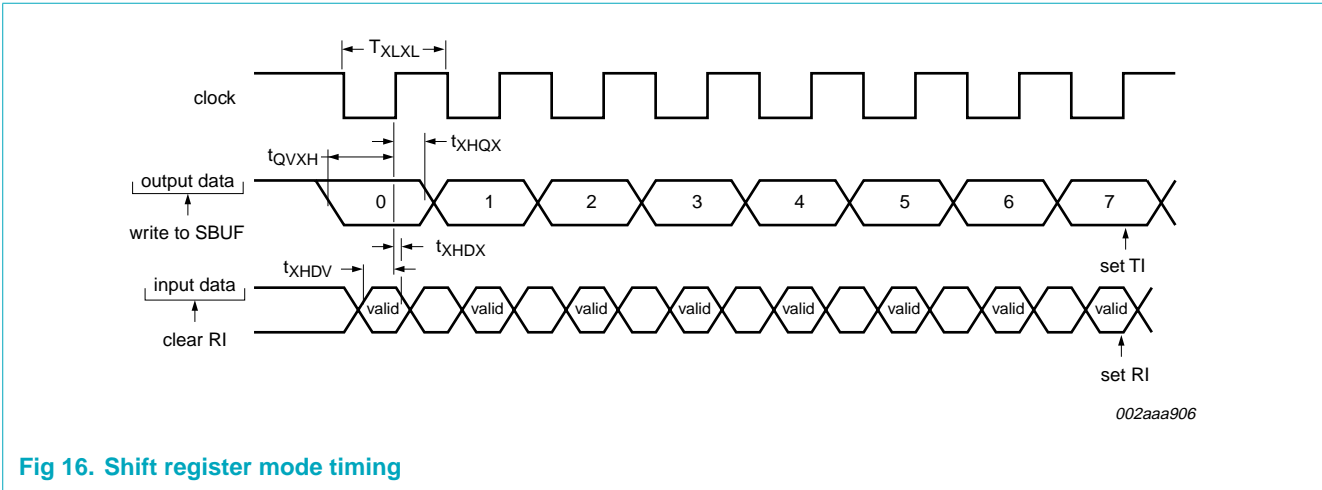


Fig 16. Shift register mode timing

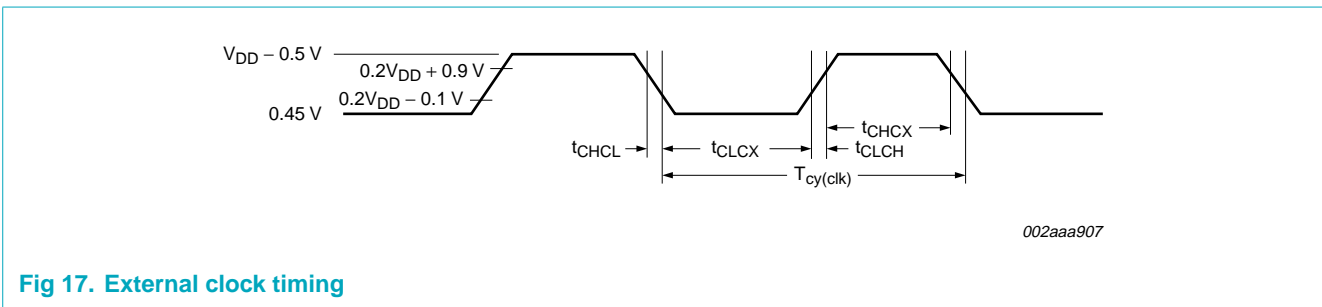


Fig 17. External clock timing

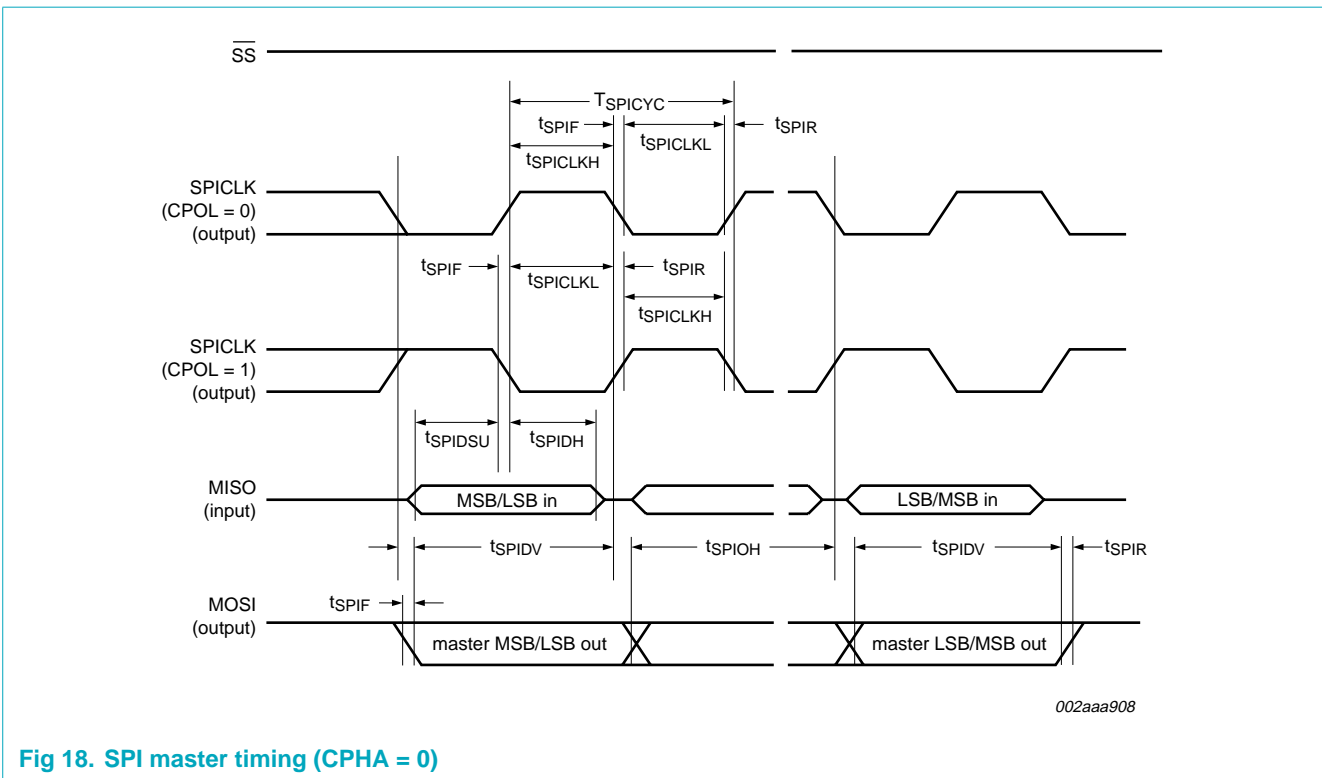
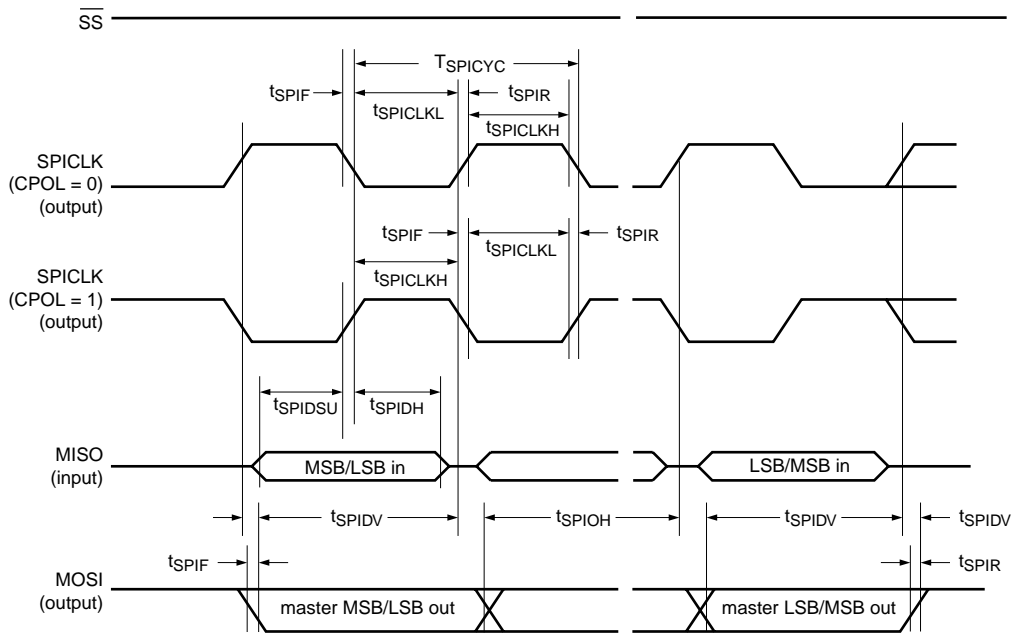
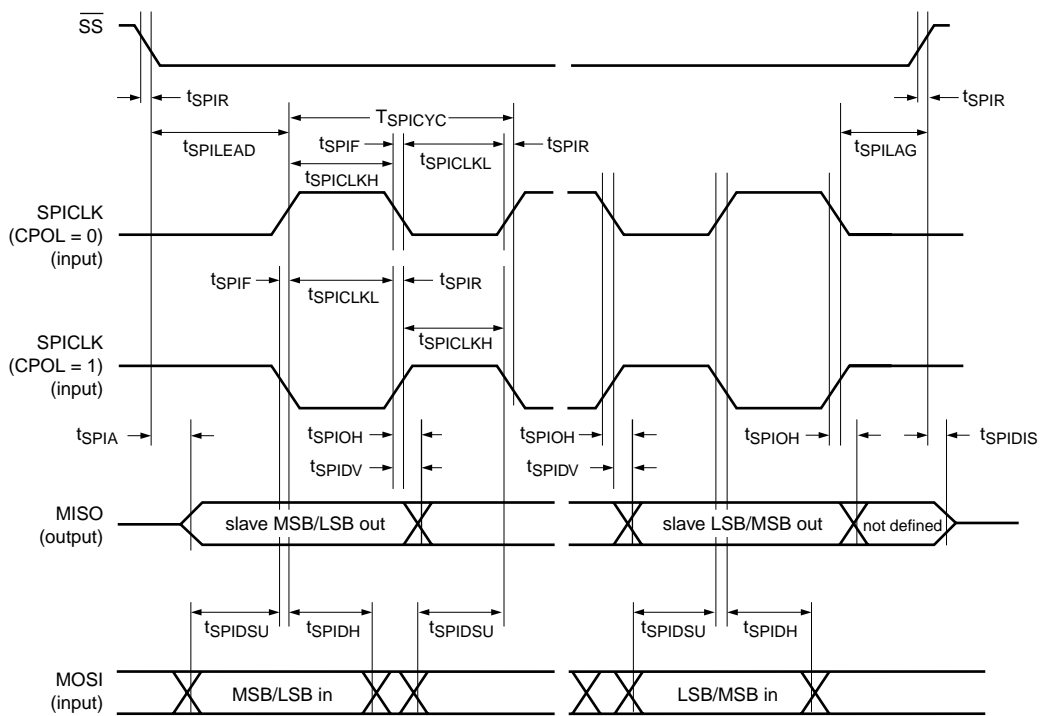


Fig 18. SPI master timing (CPHA = 0)



002aaa909

Fig 19. SPI master timing (CPHA = 1)



002aaa910

Fig 20. SPI slave timing (CPHA = 0)

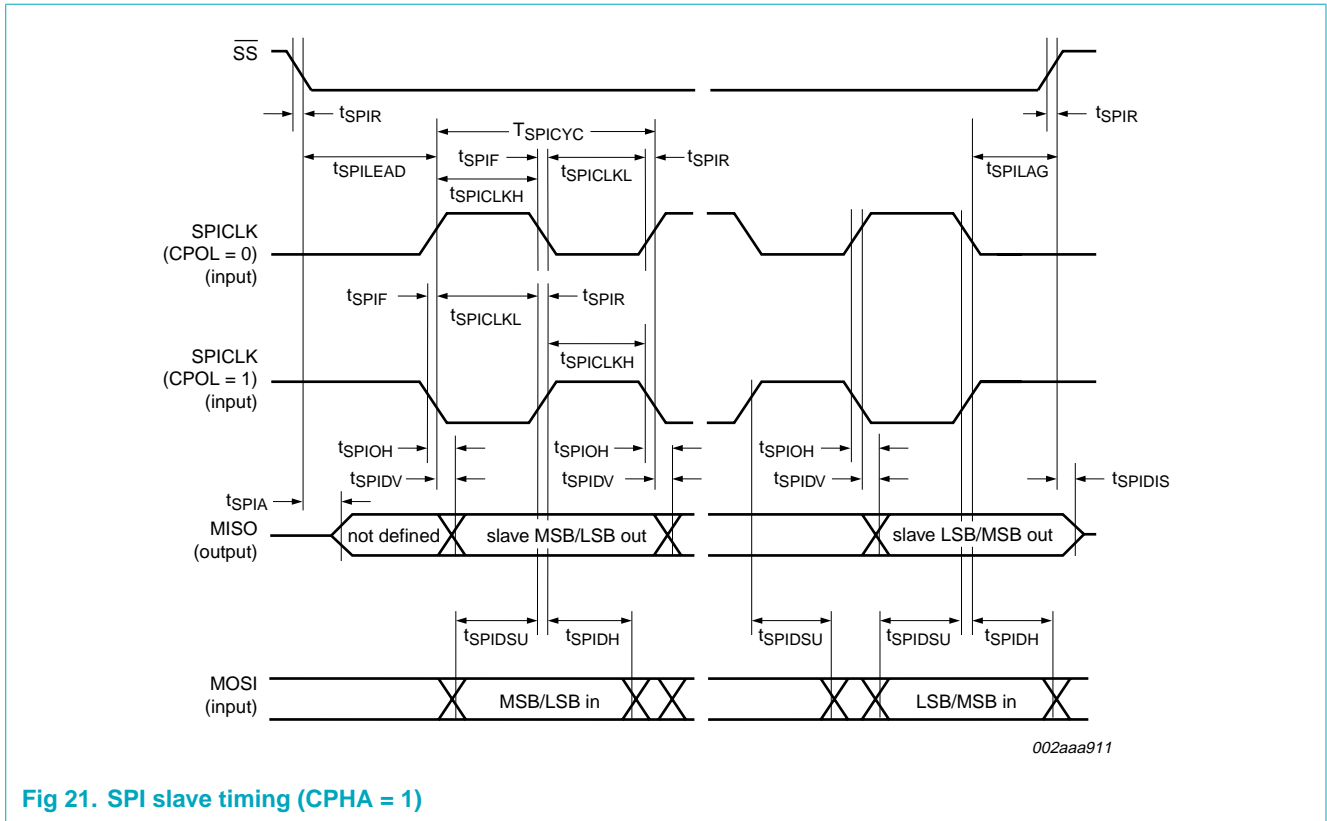


Fig 21. SPI slave timing (CPHA = 1)

### 11.2 ISP entry mode

Table 13. Dynamic characteristics, ISP entry mode

$V_{DD} = 2.4\text{ V to }3.6\text{ V}$ , unless otherwise specified.

$T_{amb} = -40\text{ }^{\circ}\text{C to }+85\text{ }^{\circ}\text{C}$  for industrial applications, unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{VR}$	$V_{DD}$ active to RST active delay time		50	-	-	$\mu\text{s}$
$t_{RH}$	$\overline{\text{RST}}$ HIGH time		1	-	32	$\mu\text{s}$
$t_{RL}$	$\overline{\text{RST}}$ LOW time		1	-	-	$\mu\text{s}$

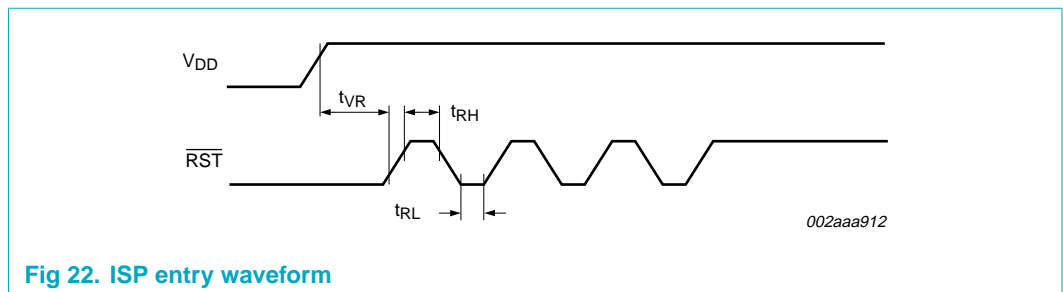


Fig 22. ISP entry waveform

## 12. Other characteristics

### 12.1 Comparator electrical characteristics

**Table 14. Comparator electrical characteristics**

$V_{DD} = 2.4\text{ V to }3.6\text{ V}$ , unless otherwise specified.

$T_{amb} = -40\text{ °C to }+85\text{ °C}$  for industrial applications, unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IO}$	input offset voltage		-	-	$\pm 10$	mV
$V_{IC}$	common-mode input voltage		0	-	$V_{DD} - 0.3$	V
CMRR	common-mode rejection ratio		[1] -	-	-50	dB
$t_{res(tot)}$	total response time		-	250	500	ns
$t_{(CE-OV)}$	chip enable to output valid time		-	-	10	$\mu\text{s}$
$I_{LI}$	input leakage current	$0\text{ V} < V_I < V_{DD}$	-	-	$\pm 10$	$\mu\text{A}$

[1] This parameter is characterized, but not tested in production.

## 12.2 ADC electrical characteristics

**Table 15. ADC electrical characteristics**

$V_{DD} = 2.4\text{ V to }3.6\text{ V}$ , unless otherwise specified.

$T_{amb} = -40\text{ °C to }+85\text{ °C}$  for industrial applications, unless otherwise specified.

All limits valid for an external source impedance of less than  $10\text{ k}\Omega$ .

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IA}$	analog input voltage		$V_{SS} - 0.2$	-	$V_{DD} + 0.2$	V
$C_{ia}$	analog input capacitance		-	-	15	pF
DNL	differential non-linearity		-	-	$\pm 1$	LSB
INL	integral non-linearity		-	-	$\pm 1$	LSB
$E_{offset}$	offset error		-	-	$\pm 2$	LSB
$E_G$	gain error		-	-	$\pm 1$	LSB
$E_{u(tot)}$	total unadjusted error		-	-	$\pm 2$	LSB
$M_{CTC}$	channel-to-channel matching		-	-	$\pm 1$	LSB
$\alpha_{ct(port)}$	crosstalk between port inputs	0 kHz to 100 kHz	-	-	-60	dB
$SR_{in}$	input slew rate		-	-	100	V/ms
$T_{cy(ADC)}$	ADC clock cycle time		111	-	3125	ns
$t_{ADC}$	ADC conversion time	ADC enabled	-	-	$36T_{cy(ADC)}$	$\mu\text{s}$

13. Package outline

TSSOP28: plastic thin shrink small outline package; 28 leads; body width 4.4 mm

SOT361-1

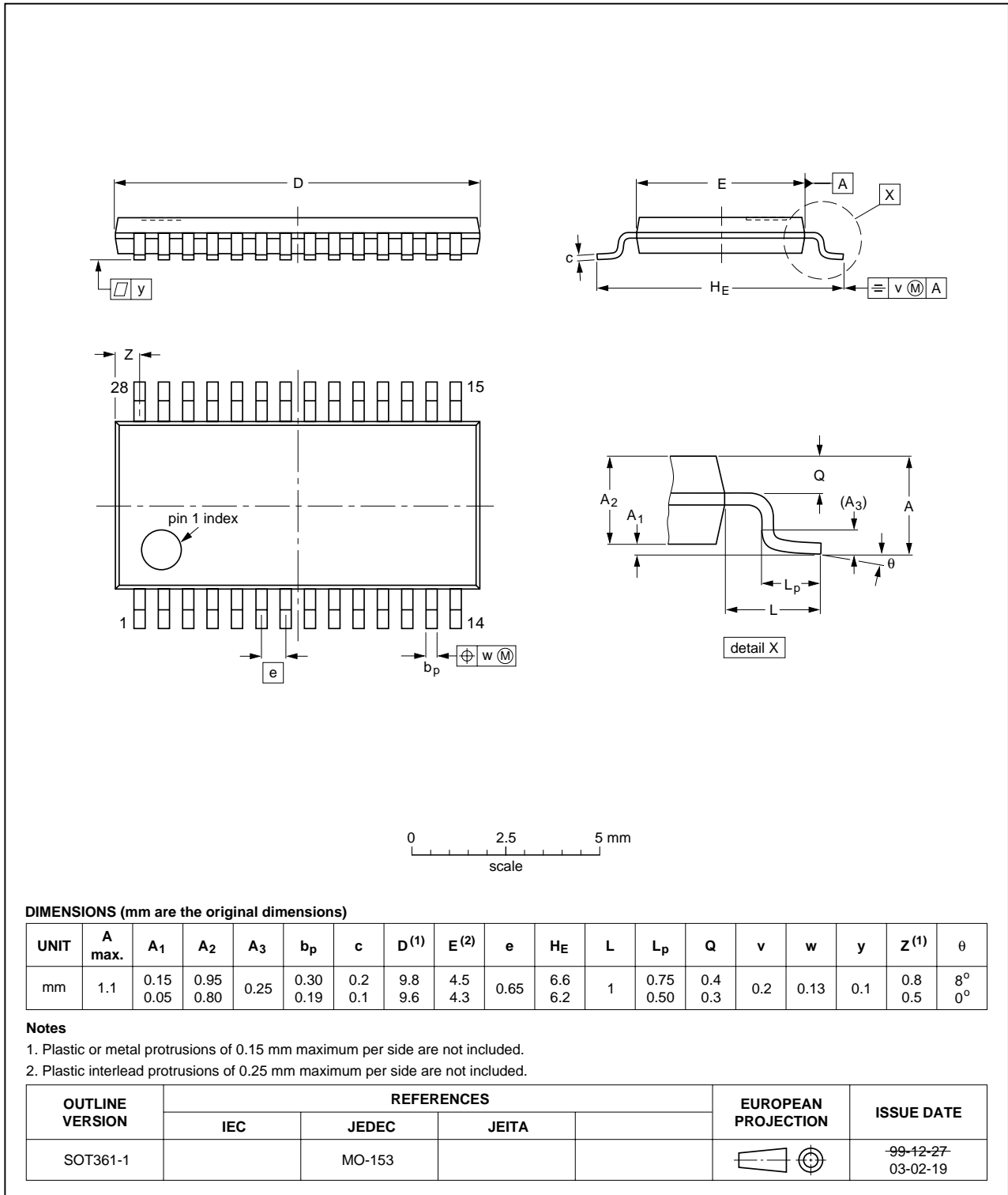


Fig 23. Package outline SOT361-1 (TSSOP28)

## 14. Abbreviations

**Table 16. Abbreviations**

Acronym	Description
ADC	Analog to Digital Converter
CPU	Central Processing Unit
DAC	Digital to Analog Converter
EPROM	Erasable Programmable Read-Only Memory
EMI	Electromagnetic Interference
IAP	In-Application Programming
ICP	In-Circuit Programming
ISP	In-System Programming
LED	Light Emitting Diode
PLL	Phase-Locked Loop
PWM	Pulse Width Modulator
RAM	Random Access Memory
RC	Resistance-Capacitance
RTC	Real-Time Clock
SAR	Successive Approximation Register
SFR	Special Function Register
SPI	Serial Peripheral Interface
UART	Universal Asynchronous Receiver/Transmitter



## 15. Revision history

Table 17. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
P89LPC9381_1	20060908	Product data sheet	-	-

## 16. Legal information

### 16.1 Data sheet status

Document status <sup>[1][2]</sup>	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL <http://www.semiconductors.philips.com>.

### 16.2 Definitions

**Draft** — The document is a draft version only. The content is still under internal review and subject to formal approval, which may result in modifications or additions. Philips Semiconductors does not give any representations or warranties as to the accuracy or completeness of information included herein and shall have no liability for the consequences of use of such information.

**Short data sheet** — A short data sheet is an extract from a full data sheet with the same product type number(s) and title. A short data sheet is intended for quick reference only and should not be relied upon to contain detailed and full information. For detailed and full information see the relevant full data sheet, which is available on request via the local Philips Semiconductors sales office. In case of any inconsistency or conflict with the short data sheet, the full data sheet shall prevail.

### 16.3 Disclaimers

**General** — Information in this document is believed to be accurate and reliable. However, Philips Semiconductors does not give any representations or warranties, expressed or implied, as to the accuracy or completeness of such information and shall have no liability for the consequences of use of such information.

**Right to make changes** — Philips Semiconductors reserves the right to make changes to information published in this document, including without limitation specifications and product descriptions, at any time and without notice. This document supersedes and replaces all information supplied prior to the publication hereof.

**Suitability for use** — Philips Semiconductors products are not designed, authorized or warranted to be suitable for use in medical, military, aircraft, space or life support equipment, nor in applications where failure or malfunction of a Philips Semiconductors product can reasonably be expected

to result in personal injury, death or severe property or environmental damage. Philips Semiconductors accepts no liability for inclusion and/or use of Philips Semiconductors products in such equipment or applications and therefore such inclusion and/or use is at the customer's own risk.

**Applications** — Applications that are described herein for any of these products are for illustrative purposes only. Philips Semiconductors makes no representation or warranty that such applications will be suitable for the specified use without further testing or modification.

**Limiting values** — Stress above one or more limiting values (as defined in the Absolute Maximum Ratings System of IEC 60134) may cause permanent damage to the device. Limiting values are stress ratings only and operation of the device at these or any other conditions above those given in the Characteristics sections of this document is not implied. Exposure to limiting values for extended periods may affect device reliability.

**Terms and conditions of sale** — Philips Semiconductors products are sold subject to the general terms and conditions of commercial sale, as published at <http://www.semiconductors.philips.com/profile/terms>, including those pertaining to warranty, intellectual property rights infringement and limitation of liability, unless explicitly otherwise agreed to in writing by Philips Semiconductors. In case of any inconsistency or conflict between information in this document and such terms and conditions, the latter will prevail.

**No offer to sell or license** — Nothing in this document may be interpreted or construed as an offer to sell products that is open for acceptance or the grant, conveyance or implication of any license under any copyrights, patents or other industrial or intellectual property rights.

### 16.4 Trademarks

Notice: All referenced brands, product names, service names and trademarks are the property of their respective owners.

**IC-bus** — logo is a trademark of Koninklijke Philips Electronics N.V.

## 17. Contact information

For additional information, please visit: <http://www.semiconductors.philips.com>

For sales office addresses, send an email to: [sales.addresses@www.semiconductors.philips.com](mailto:sales.addresses@www.semiconductors.philips.com)

**18. Contents**

<b>1</b>	<b>General description</b> . . . . .	<b>1</b>	7.16.1	Reset vector . . . . .	25
<b>2</b>	<b>Features</b> . . . . .	<b>1</b>	7.17	Timers/counters 0 and 1 . . . . .	25
2.1	Principal features . . . . .	1	7.17.1	Mode 0 . . . . .	26
2.2	Additional features . . . . .	2	7.17.2	Mode 1 . . . . .	26
<b>3</b>	<b>Ordering information</b> . . . . .	<b>3</b>	7.17.3	Mode 2 . . . . .	26
3.1	Ordering options . . . . .	3	7.17.4	Mode 3 . . . . .	26
<b>4</b>	<b>Block diagram</b> . . . . .	<b>4</b>	7.17.5	Mode 6 . . . . .	26
<b>5</b>	<b>Functional diagram</b> . . . . .	<b>5</b>	7.17.6	Timer overflow toggle output . . . . .	26
<b>6</b>	<b>Pinning information</b> . . . . .	<b>6</b>	7.18	RTC/system timer . . . . .	26
6.1	Pinning . . . . .	6	7.19	UART . . . . .	26
6.2	Pin description . . . . .	6	7.19.1	Mode 0 . . . . .	27
<b>7</b>	<b>Functional description</b> . . . . .	<b>10</b>	7.19.2	Mode 1 . . . . .	27
7.1	Special function registers . . . . .	10	7.19.3	Mode 2 . . . . .	27
7.2	Enhanced CPU . . . . .	17	7.19.4	Mode 3 . . . . .	27
7.3	Clocks . . . . .	17	7.19.5	Baud rate generator and selection . . . . .	27
7.3.1	Clock definitions . . . . .	17	7.19.6	Framing error . . . . .	28
7.3.2	CPU clock (OSCCLK) . . . . .	17	7.19.7	Break detect . . . . .	28
7.3.3	Low speed oscillator option . . . . .	17	7.19.8	Double buffering . . . . .	28
7.3.4	Medium speed oscillator option . . . . .	17	7.19.9	Transmit interrupts with double buffering enabled (modes 1, 2 and 3) . . . . .	28
7.3.5	High speed oscillator option . . . . .	17	7.19.10	The 9 <sup>th</sup> bit (bit 8) in double buffering (modes 1, 2 and 3) . . . . .	28
7.3.6	Clock output . . . . .	17	7.20	I <sup>2</sup> C-bus serial interface . . . . .	29
7.4	On-chip RC oscillator option . . . . .	18	7.21	SPI . . . . .	31
7.5	Watchdog oscillator option . . . . .	18	7.21.1	Typical SPI configurations . . . . .	32
7.6	External clock input option . . . . .	18	7.22	Analog comparators . . . . .	34
7.7	CCLK wake-up delay . . . . .	19	7.22.1	Internal reference voltage . . . . .	34
7.8	CCLK modification: DIVM register . . . . .	19	7.22.2	Comparator interrupt . . . . .	34
7.9	Low power select . . . . .	19	7.22.3	Comparators and power reduction modes . . . . .	34
7.10	Memory organization . . . . .	19	7.23	KBI . . . . .	35
7.11	Data RAM arrangement . . . . .	20	7.24	Watchdog timer . . . . .	36
7.12	Interrupts . . . . .	20	7.25	Additional features . . . . .	36
7.12.1	External interrupt inputs . . . . .	20	7.25.1	Software reset . . . . .	36
7.13	I/O ports . . . . .	22	7.25.2	Dual data pointers . . . . .	36
7.13.1	Port configurations . . . . .	22	7.26	Flash program memory . . . . .	37
7.13.1.1	Quasi-bidirectional output configuration . . . . .	22	7.26.1	General description . . . . .	37
7.13.1.2	Open-drain output configuration . . . . .	22	7.26.2	Features . . . . .	37
7.13.1.3	Input-only configuration . . . . .	23	7.26.3	Flash organization . . . . .	37
7.13.1.4	Push-pull output configuration . . . . .	23	7.26.4	Using flash as data storage . . . . .	37
7.13.2	Port 0 analog functions . . . . .	23	7.26.5	Flash programming and erasing . . . . .	37
7.13.3	Additional port features . . . . .	23	7.26.6	ICP . . . . .	38
7.14	Power monitoring functions . . . . .	23	7.26.7	IAP . . . . .	38
7.14.1	Brownout detection . . . . .	23	7.26.8	ISP . . . . .	38
7.14.2	Power-on detection . . . . .	24	7.26.9	Power-on reset code execution . . . . .	39
7.15	Power reduction modes . . . . .	24	7.26.10	Hardware activation of the bootloader . . . . .	39
7.15.1	Idle mode . . . . .	24	7.27	User configuration bytes . . . . .	39
7.15.2	Power-down mode . . . . .	24	7.28	User sector security bytes . . . . .	39
7.15.3	Total Power-down mode . . . . .	24	<b>8</b>	<b>ADC</b> . . . . .	<b>39</b>
7.16	Reset . . . . .	25			

continued >>

8.1	General description . . . . .	39
8.2	Features . . . . .	40
8.3	Block diagram . . . . .	40
8.4	ADC operating modes . . . . .	41
8.4.1	Fixed channel, single conversion mode . . . . .	41
8.4.2	Fixed channel, continuous conversion mode . . . . .	41
8.4.3	Auto scan, single conversion mode . . . . .	41
8.4.4	Auto scan, continuous conversion mode . . . . .	41
8.4.5	Dual channel, continuous conversion mode . . . . .	41
8.4.6	Single step mode . . . . .	42
8.5	Conversion start modes . . . . .	42
8.5.1	Timer triggered start . . . . .	42
8.5.2	Start immediately . . . . .	42
8.5.3	Edge triggered . . . . .	42
8.6	Boundary limits interrupt . . . . .	42
8.7	Clock divider . . . . .	42
8.8	Power-down and Idle mode . . . . .	42
<b>9</b>	<b>Limiting values . . . . .</b>	<b>43</b>
<b>10</b>	<b>Static characteristics . . . . .</b>	<b>44</b>
<b>11</b>	<b>Dynamic characteristics . . . . .</b>	<b>46</b>
11.1	Waveforms . . . . .	50
11.2	ISP entry mode . . . . .	52
<b>12</b>	<b>Other characteristics . . . . .</b>	<b>53</b>
12.1	Comparator electrical characteristics . . . . .	53
12.2	ADC electrical characteristics . . . . .	54
<b>13</b>	<b>Package outline . . . . .</b>	<b>55</b>
<b>14</b>	<b>Abbreviations . . . . .</b>	<b>56</b>
<b>15</b>	<b>Revision history . . . . .</b>	<b>57</b>
<b>16</b>	<b>Legal information . . . . .</b>	<b>58</b>
16.1	Data sheet status . . . . .	58
16.2	Definitions . . . . .	58
16.3	Disclaimers . . . . .	58
16.4	Trademarks . . . . .	58
<b>17</b>	<b>Contact information . . . . .</b>	<b>58</b>
<b>18</b>	<b>Contents . . . . .</b>	<b>59</b>



Please be aware that important notices concerning this document and the product(s) described herein, have been included in section 'Legal information'.

© Koninklijke Philips Electronics N.V. 2006. All rights reserved.

For more information, please visit: <http://www.semiconductors.philips.com>.

For sales office addresses, email to: [sales.addresses@www.semiconductors.philips.com](mailto:sales.addresses@www.semiconductors.philips.com).

Date of release: 8 September 2006

Document identifier: P89LPC9381\_1